

# Structured Electronic Design

## Frequency stability of feedback amplifiers

# Introduction

# Introduction

1. A system is stable if all its poles have a negative real part

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain
  - c. The poles of the servo function must have a negative real part

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain
  - c. The poles of the servo function must have a negative real part
  - d. The poles of the direct transfer must have a negative real part



# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain
  - c. The poles of the servo function must have a negative real part
  - d. The poles of the direct transfer must have a negative real part
3. We will assume a, b and d to be true, and present techniques to evaluate c:

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain
  - c. The poles of the servo function must have a negative real part
  - d. The poles of the direct transfer must have a negative real part
3. We will assume a, b and d to be true, and present techniques to evaluate c:
  - a. Routh-Hurwitz criterion

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain
  - c. The poles of the servo function must have a negative real part
  - d. The poles of the direct transfer must have a negative real part
3. We will assume a, b and d to be true, and present techniques to evaluate c:
  - a. Routh-Hurwitz criterion
  - b. Nyquist criterion

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain
  - c. The poles of the servo function must have a negative real part
  - d. The poles of the direct transfer must have a negative real part
3. We will assume a, b and d to be true, and present techniques to evaluate c:
  - a. Routh-Hurwitz criterion
  - b. Nyquist criterion
  - c. Root locus technique

# Introduction

1. A system is stable if all its poles have a negative real part
2. For a negative feedback system it can be formulated as:
  - a. All poles of the ideal gain must have a negative real part
  - b. The loop gain reference has been selected such that the asymptotic gain equals the ideal gain
  - c. The poles of the servo function must have a negative real part
  - d. The poles of the direct transfer must have a negative real part
3. We will assume a, b and d to be true, and present techniques to evaluate c:
  - a. Routh-Hurwitz criterion
  - b. Nyquist criterion
  - c. Root locus technique

# Routh-Hurwitz criterion

# Routh-Hurwitz criterion

1. Routh test: 1876, Edward John Routh, 1895 Adolf Hurwitz

# Routh-Hurwitz criterion

1. Routh test: 1876, Edward John Routh, 1895 Adolf Hurwitz
2. Coefficients of the characteristic polynomial taken as input for the Routh Array



# Routh-Hurwitz criterion

1. Routh test: 1876, Edward John Routh, 1895 Adolf Hurwitz
2. Coefficients of the characteristic polynomial taken as input for the Routh Array
3. The number of sign changes in the first column of this array is equal to the number of solutions of this polynomial that have a positive real part

# Routh-Hurwitz criterion

1. Routh test: 1876, Edward John Routh, 1895 Adolf Hurwitz
2. Coefficients of the characteristic polynomial taken as input for the Routh Array
3. The number of sign changes in the first column of this array is equal to the number of solutions of this polynomial that have a positive real part
4. For detailed description, refer to: [Control Theory](#)

# Routh-Hurwitz criterion

1. Routh test: 1876, Edward John Routh, 1895 Adolf Hurwitz
2. Coefficients of the characteristic polynomial taken as input for the Routh Array
3. The number of sign changes in the first column of this array is equal to the number of solutions of this polynomial that have a positive real part
4. For detailed description, refer to: Control Theory
5. Implemented in SLiCAP

# Routh-Hurwitz criterion

1. Routh test: 1876, Edward John Routh, 1895 Adolf Hurwitz
2. Coefficients of the characteristic polynomial taken as input for the Routh Array
3. The number of sign changes in the first column of this array is equal to the number of solutions of this polynomial that have a positive real part
4. For detailed description, refer to: Control Theory
5. Implemented in SLiCAP

# Nyquist criterion

# Nyquist criterion

## 1. Nyquist: 1932

# Nyquist criterion

1. Nyquist: 1932
2. Consider a contour plot of the loop gain ( $-L$ )

# Nyquist criterion

1. Nyquist: 1932
2. Consider a contour plot of the loop gain  $(-L)$
3. The number of poles with a positive real part equals  
The difference between the number of clockwise encirclements  
of the point  $(-1,0)$  by the contour, and the number of poles of  $L$   
that have a positive real part (the latter one is usually zero)



# Nyquist criterion

1. Nyquist: 1932
2. Consider a contour plot of the loop gain  $(-L)$
3. The number of poles with a positive real part equals  
The difference between the number of clockwise encirclements of the point  $(-1,0)$  by the contour, and the number of poles of  $L$  that have a positive real part (the latter one is usually zero)
4. For detailed description, refer to: [Control Theory](#)

# Nyquist criterion

1. Nyquist: 1932
2. Consider a contour plot of the loop gain  $(-L)$
3. The number of poles with a positive real part equals  
The difference between the number of clockwise encirclements of the point  $(-1,0)$  by the contour, and the number of poles of  $L$  that have a positive real part (the latter one is usually zero)
4. For detailed description, refer to: Control Theory
5. Phase margin and amplitude margin are often taken as measure for stability

# Nyquist criterion

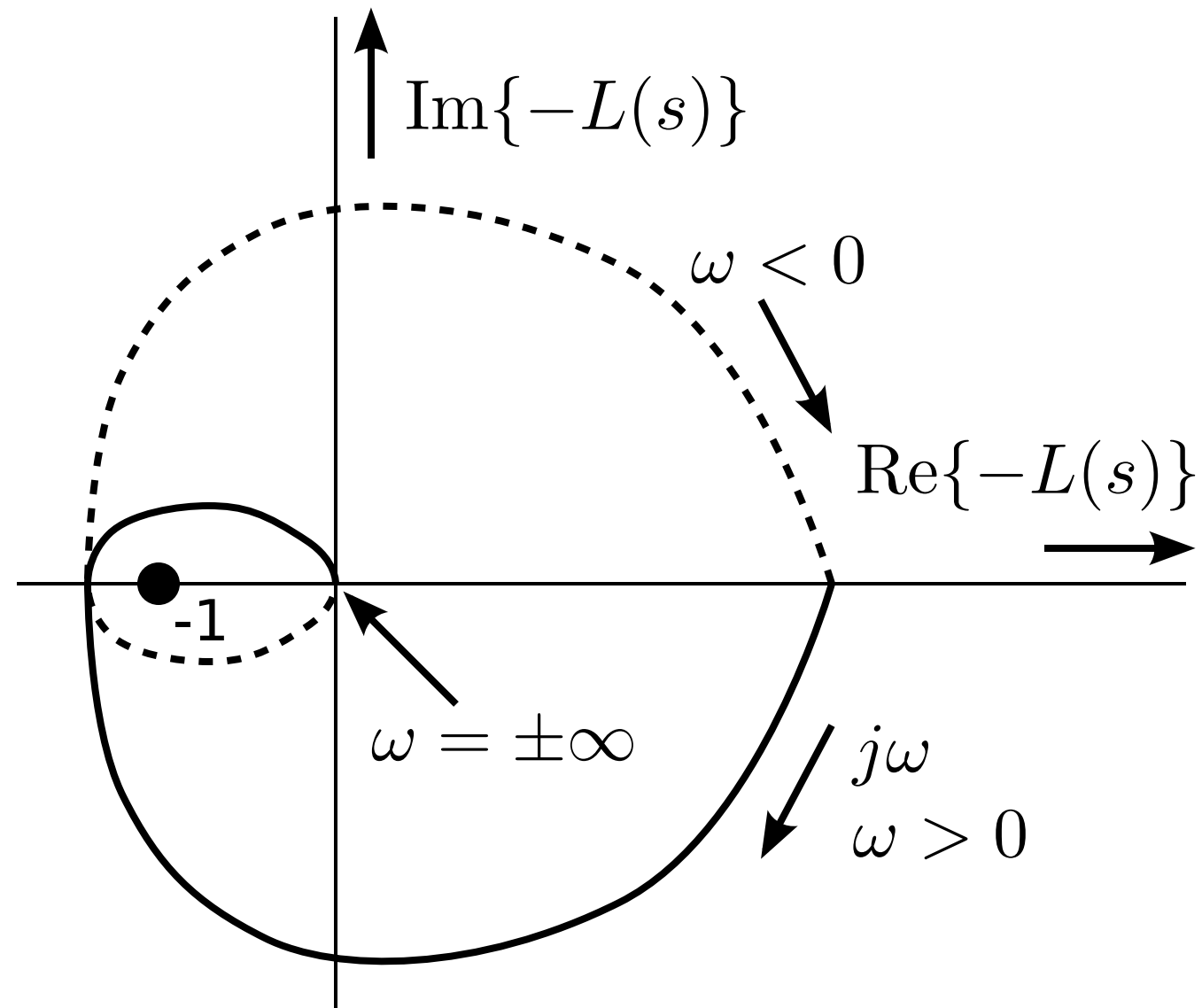
1. Nyquist: 1932
2. Consider a contour plot of the loop gain  $(-L)$
3. The number of poles with a positive real part equals  
The difference between the number of clockwise encirclements of the point  $(-1,0)$  by the contour, and the number of poles of  $L$  that have a positive real part (the latter one is usually zero)
4. For detailed description, refer to: Control Theory
5. Phase margin and amplitude margin are often taken as measure for stability
6. Implemented in SLiCAP

# Nyquist criterion

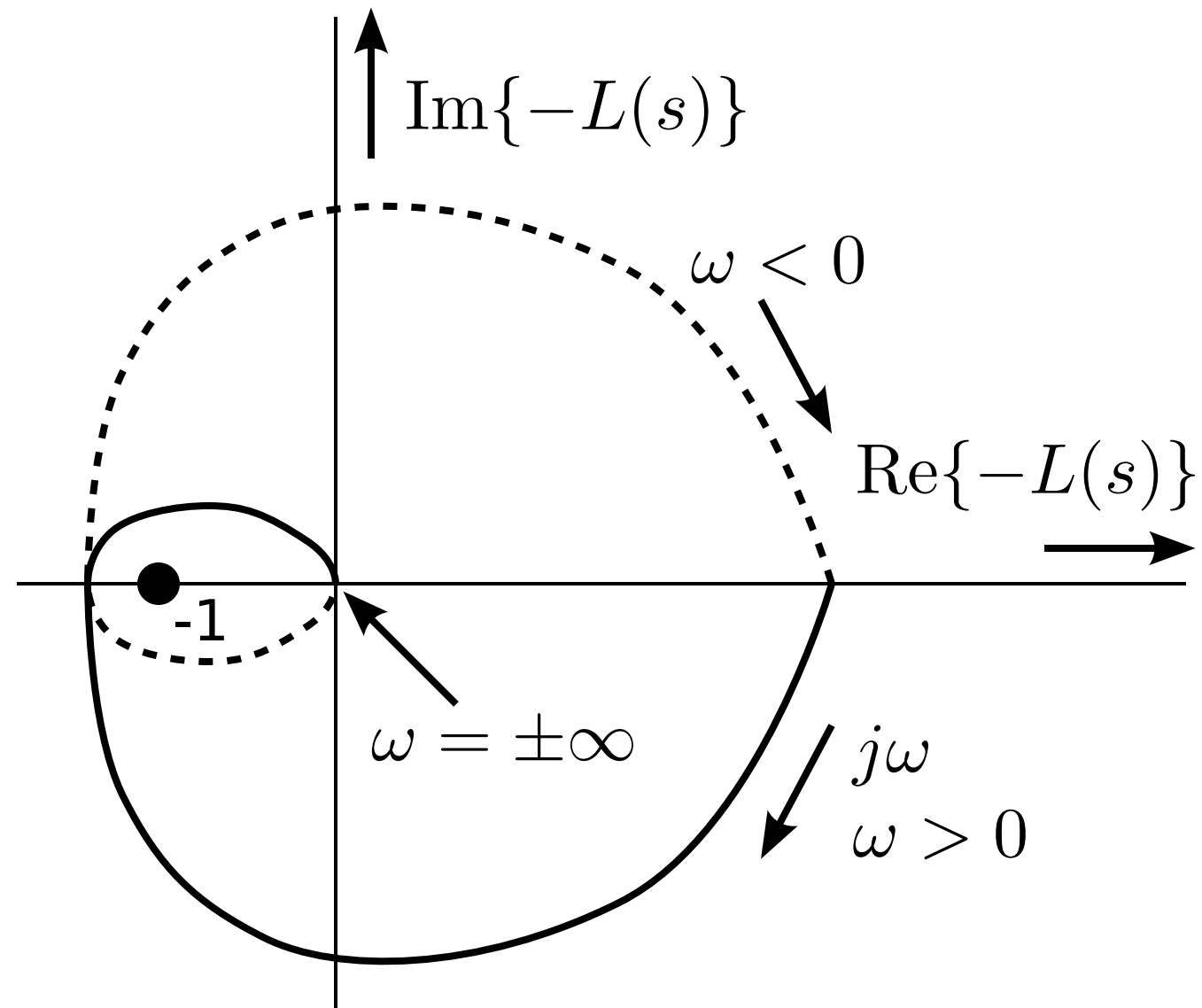
1. Nyquist: 1932
2. Consider a contour plot of the loop gain  $(-L)$
3. The number of poles with a positive real part equals  
The difference between the number of clockwise encirclements of the point  $(-1,0)$  by the contour, and the number of poles of  $L$  that have a positive real part (the latter one is usually zero)
4. For detailed description, refer to: Control Theory
5. Phase margin and amplitude margin are often taken as measure for stability
6. Implemented in SLiCAP

# Nyquist plots

# Nyquist plots

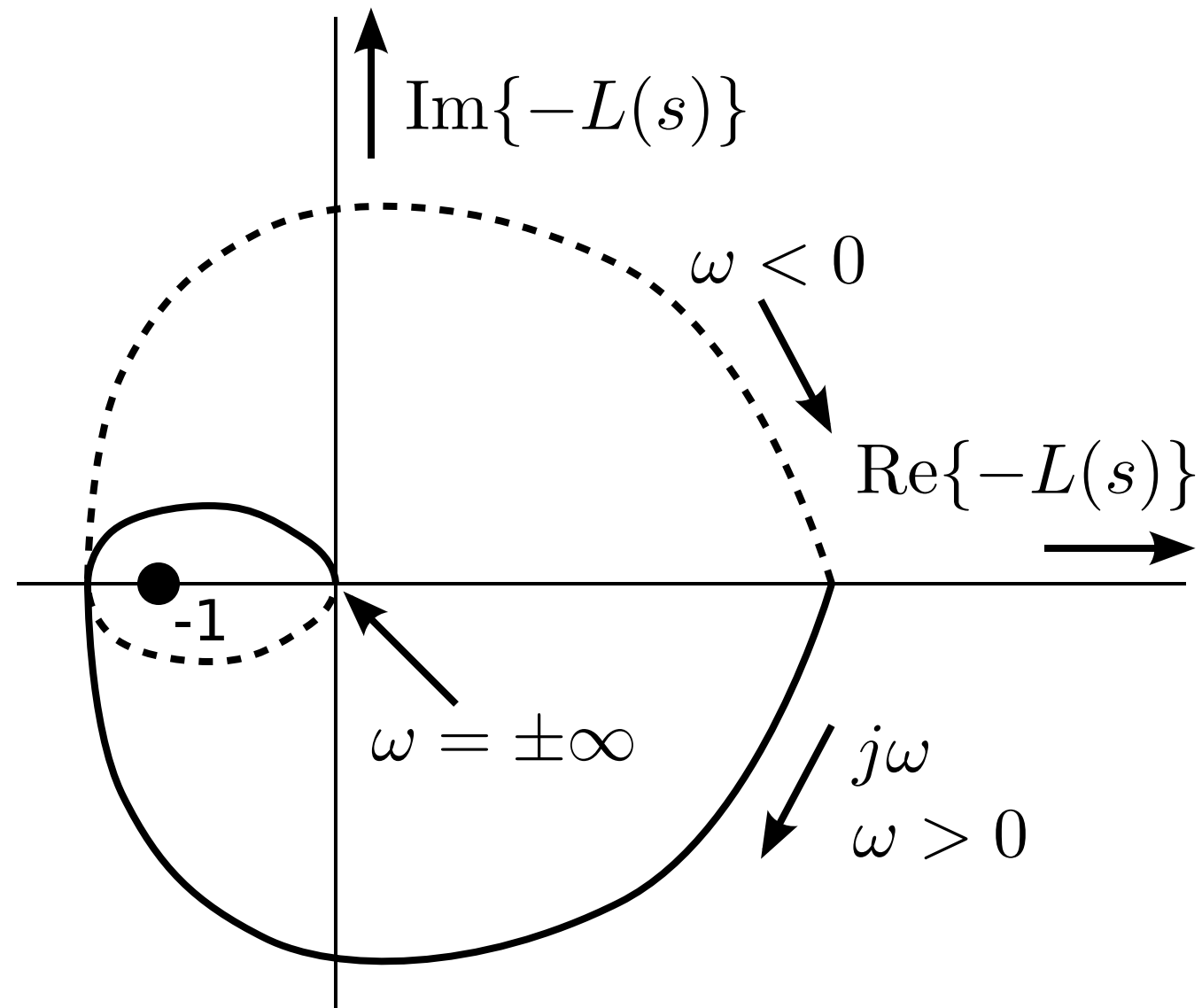


# Nyquist plots

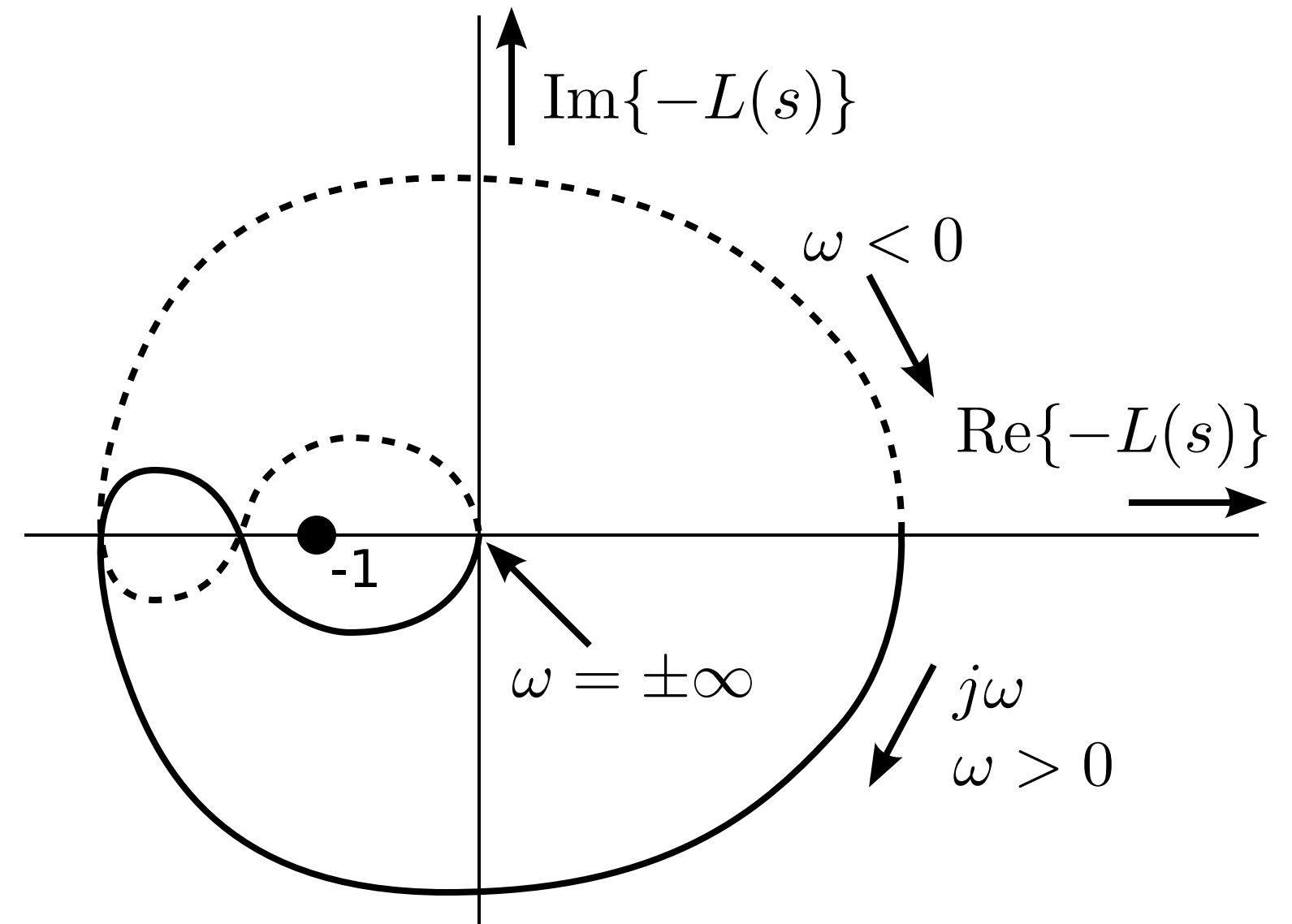


One clockwise encirclement

# Nyquist plots

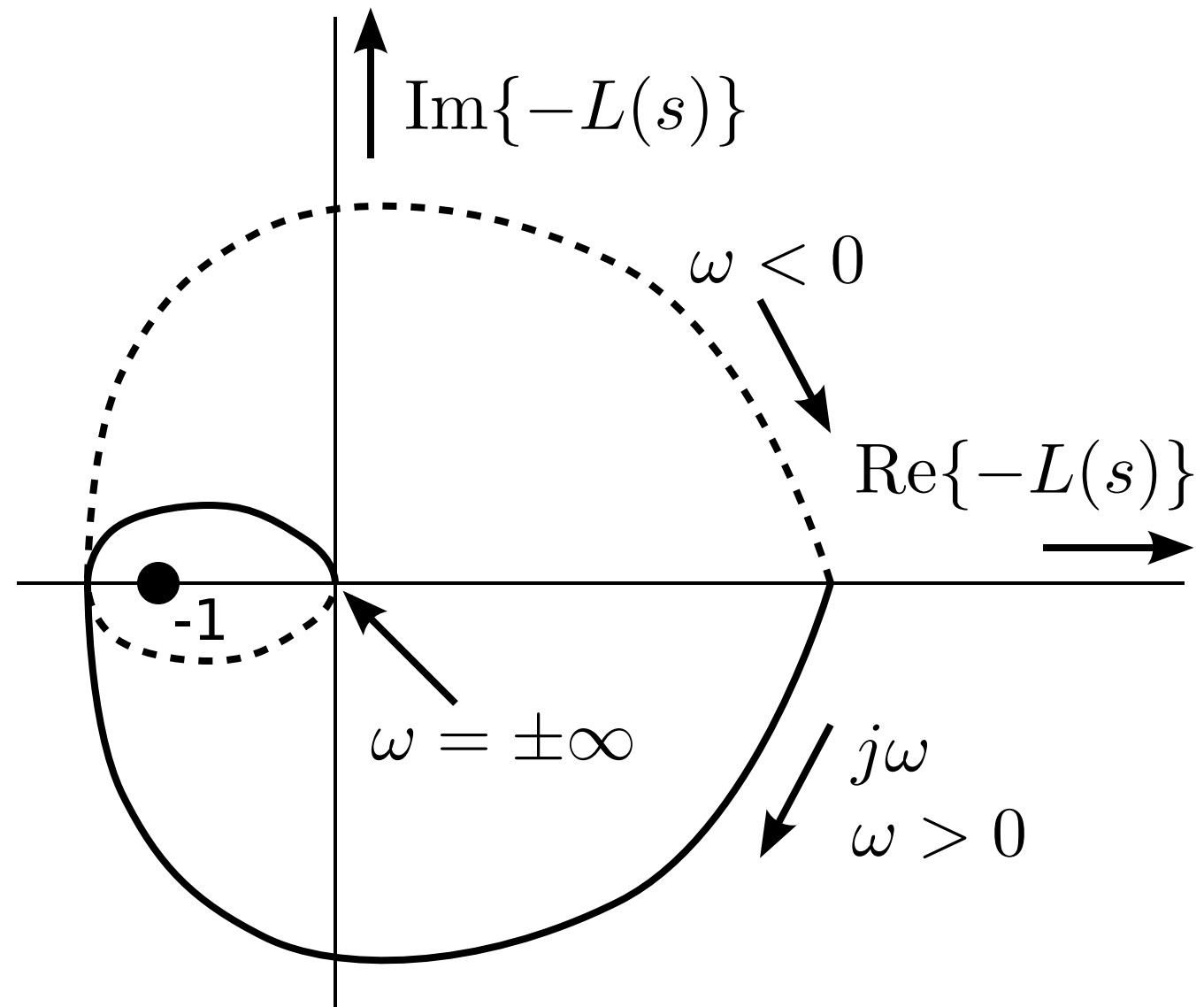


One clockwise encirclement

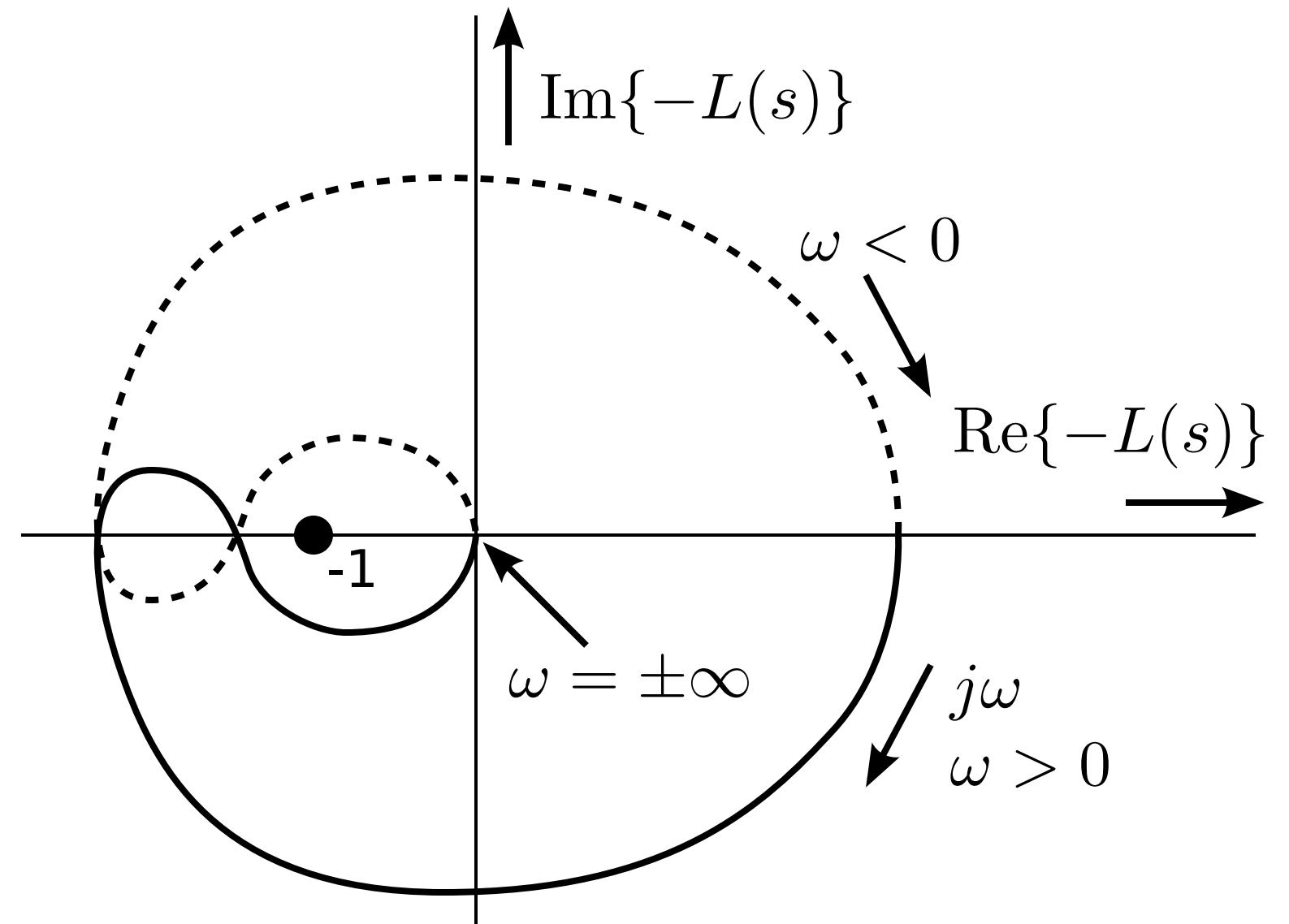




# Nyquist plots

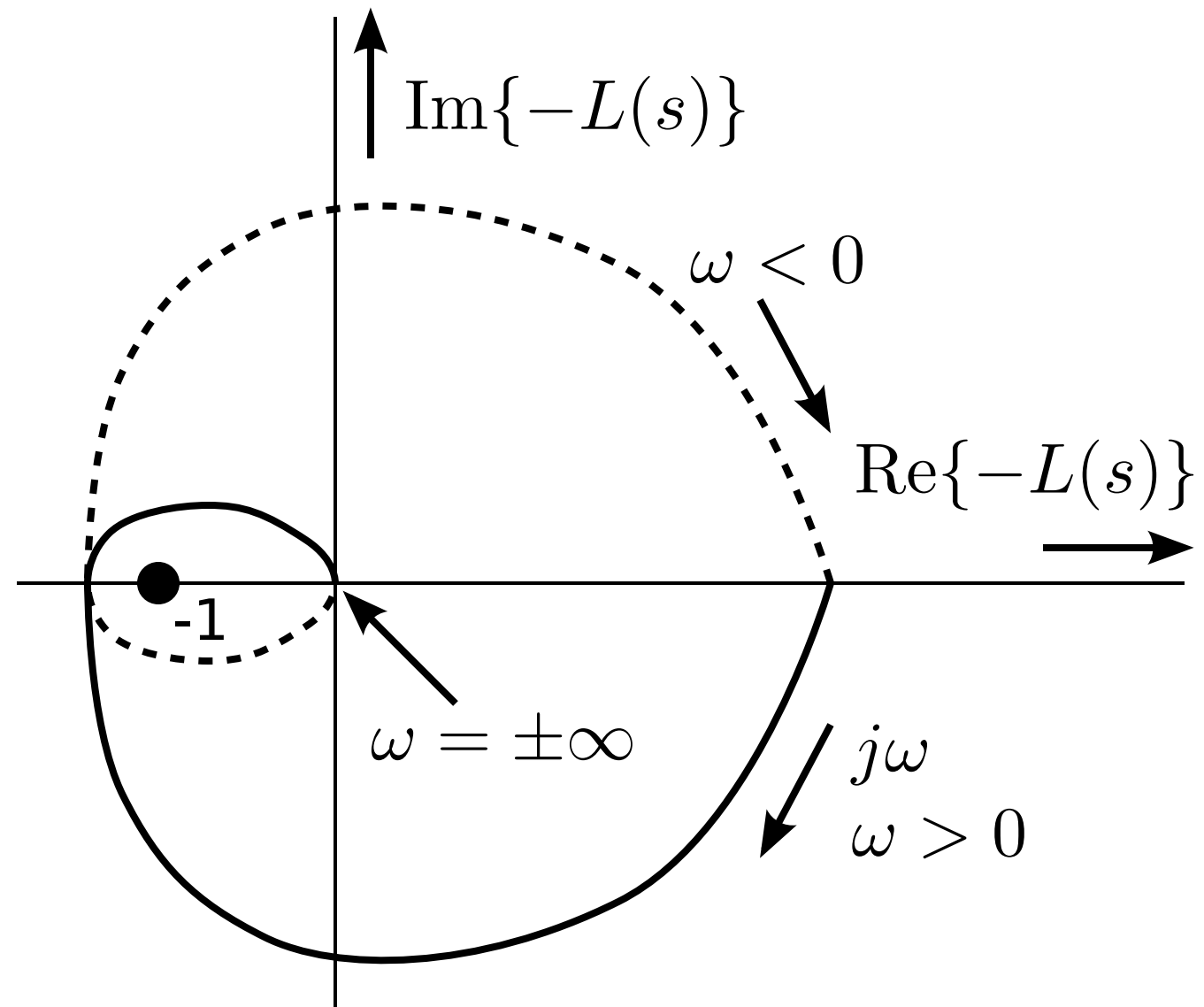


One clockwise encirclement

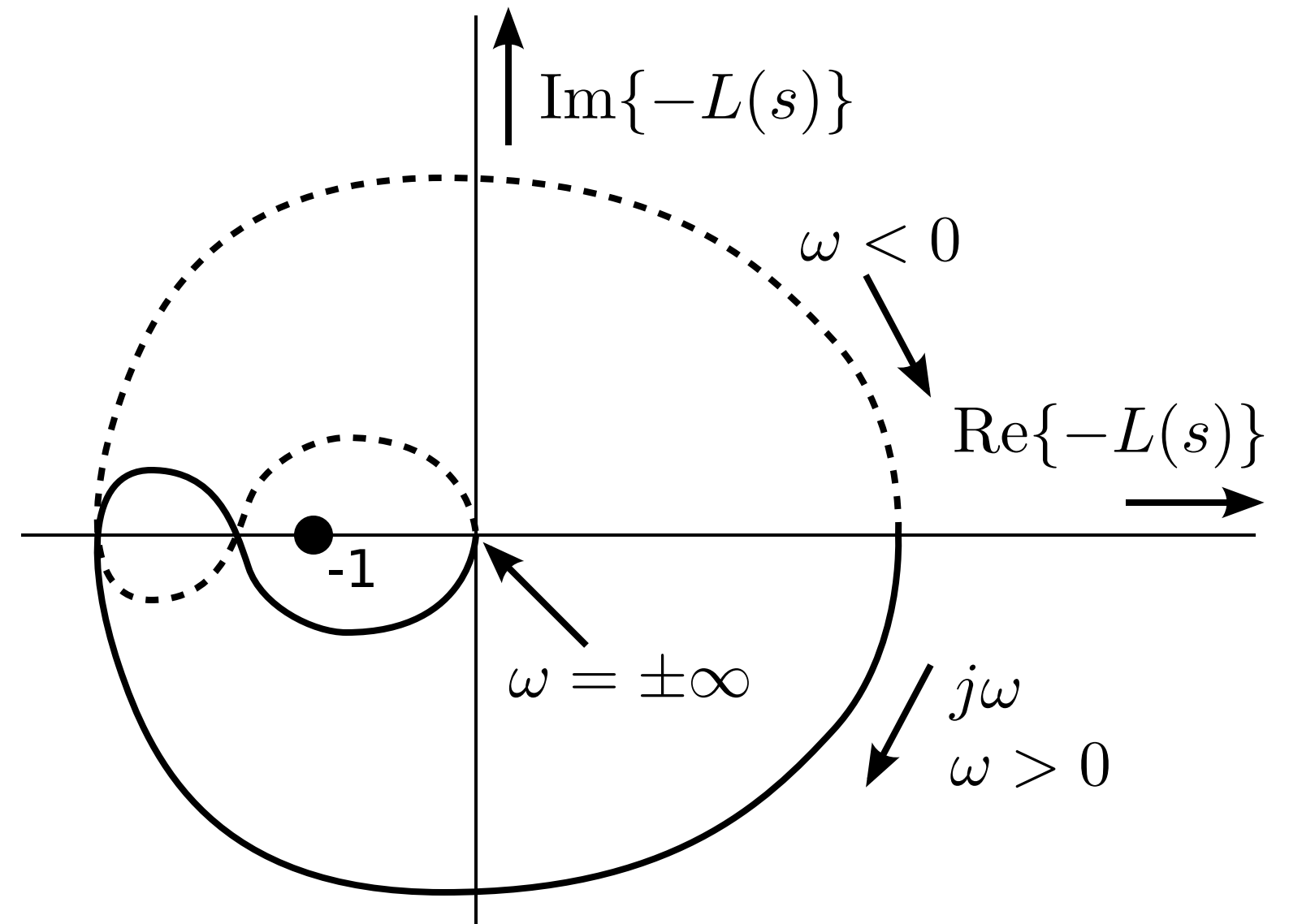


Zero clockwise encirclements

# Nyquist plots



One clockwise encirclement

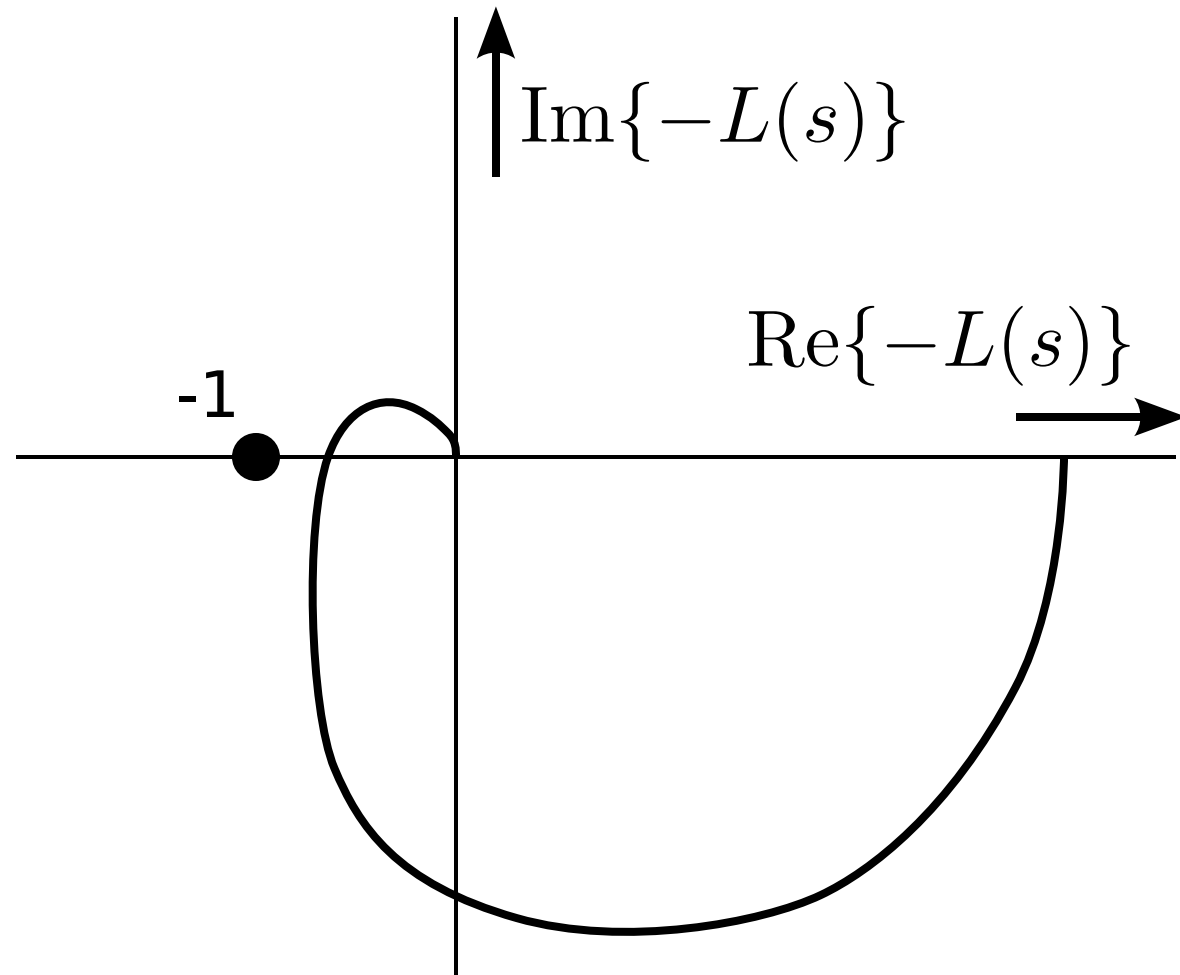


Zero clockwise encirclements

# Gain margin and phase margin

# Gain margin and phase margin

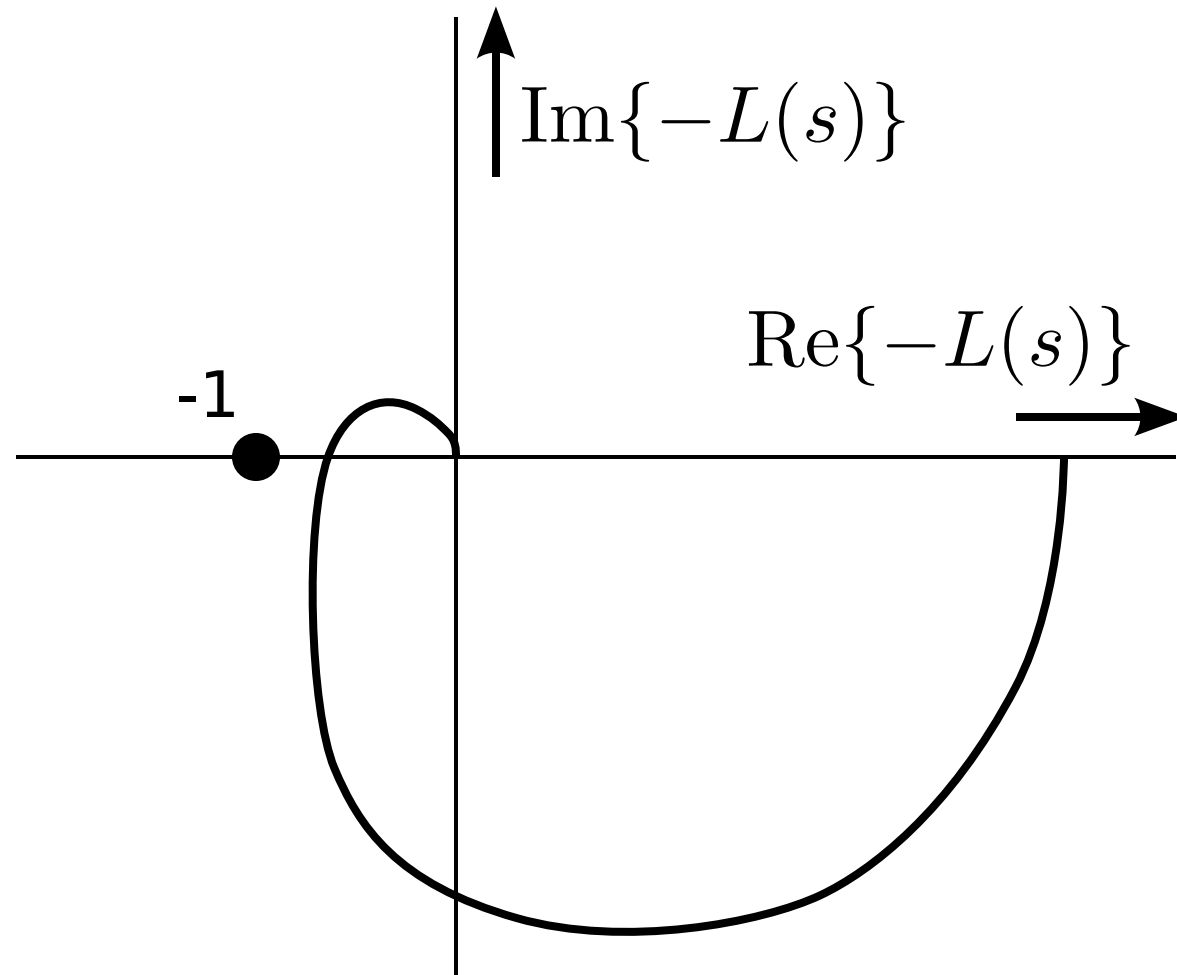
Indicative measure for stability



# Gain margin and phase margin

Indicative measure for stability

How far is the system from instability ...

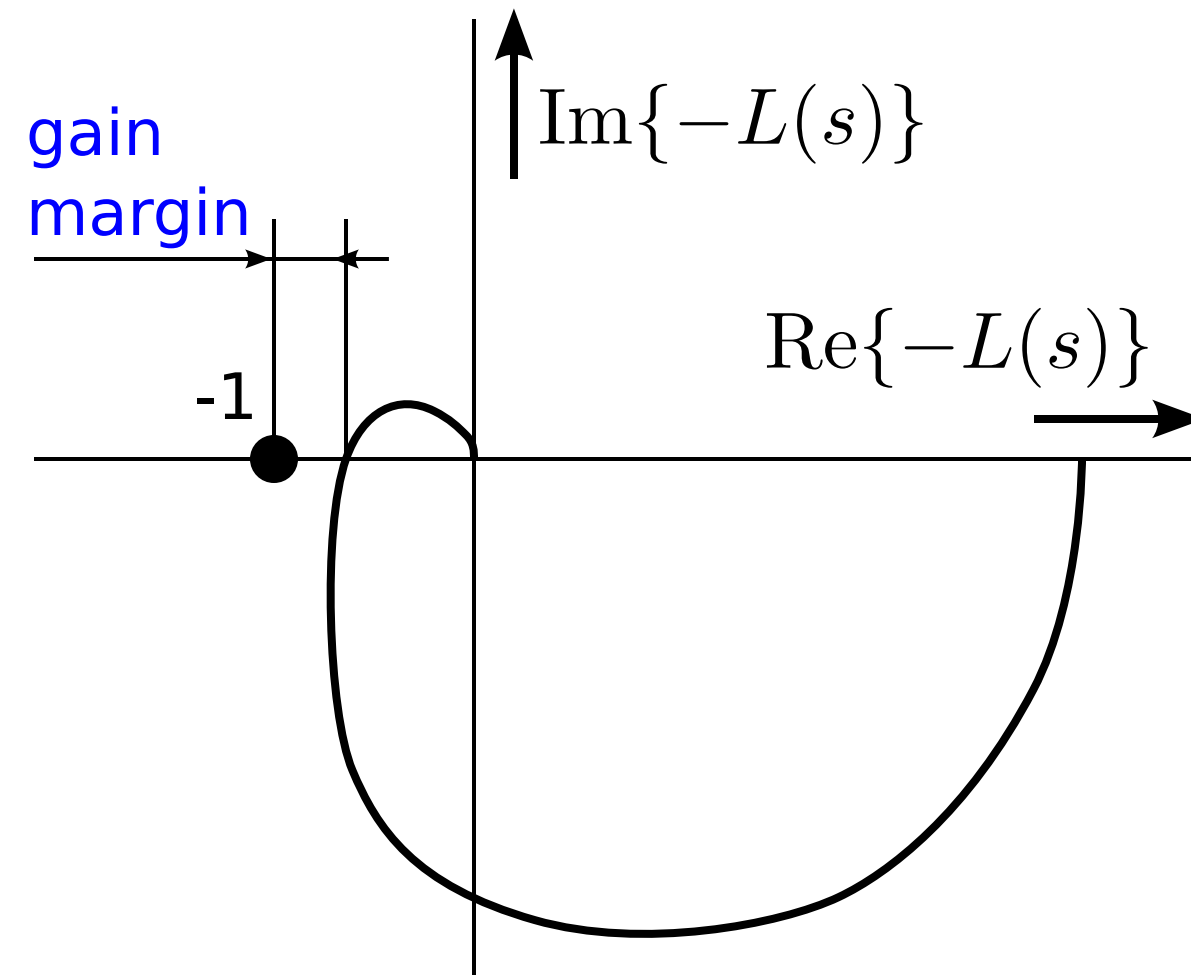


# Gain margin and phase margin

Indicative measure for stability

How far is the system from instability ...

Gain margin: how much in dB the loop gain must be increased to cause instability



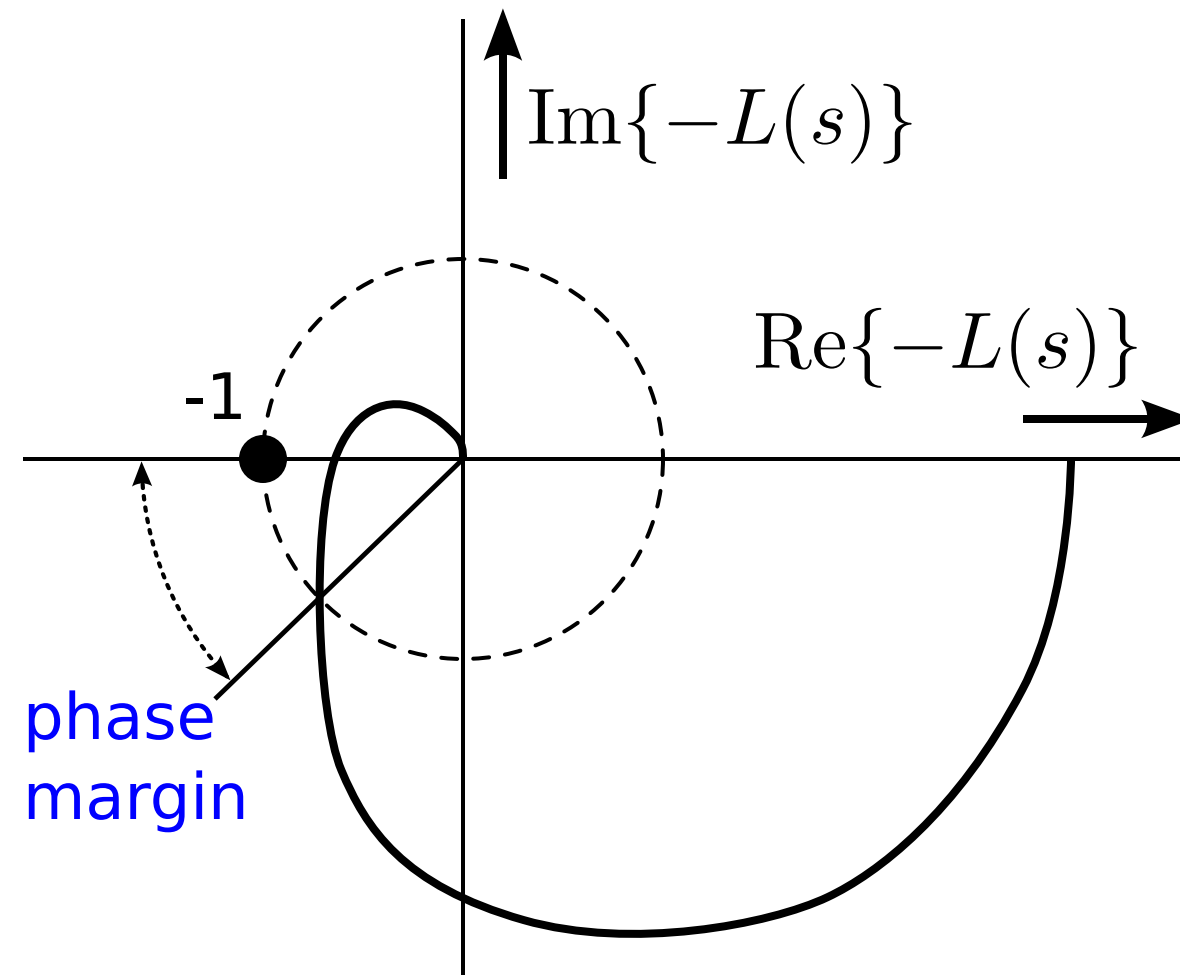
# Gain margin and phase margin

Indicative measure for stability

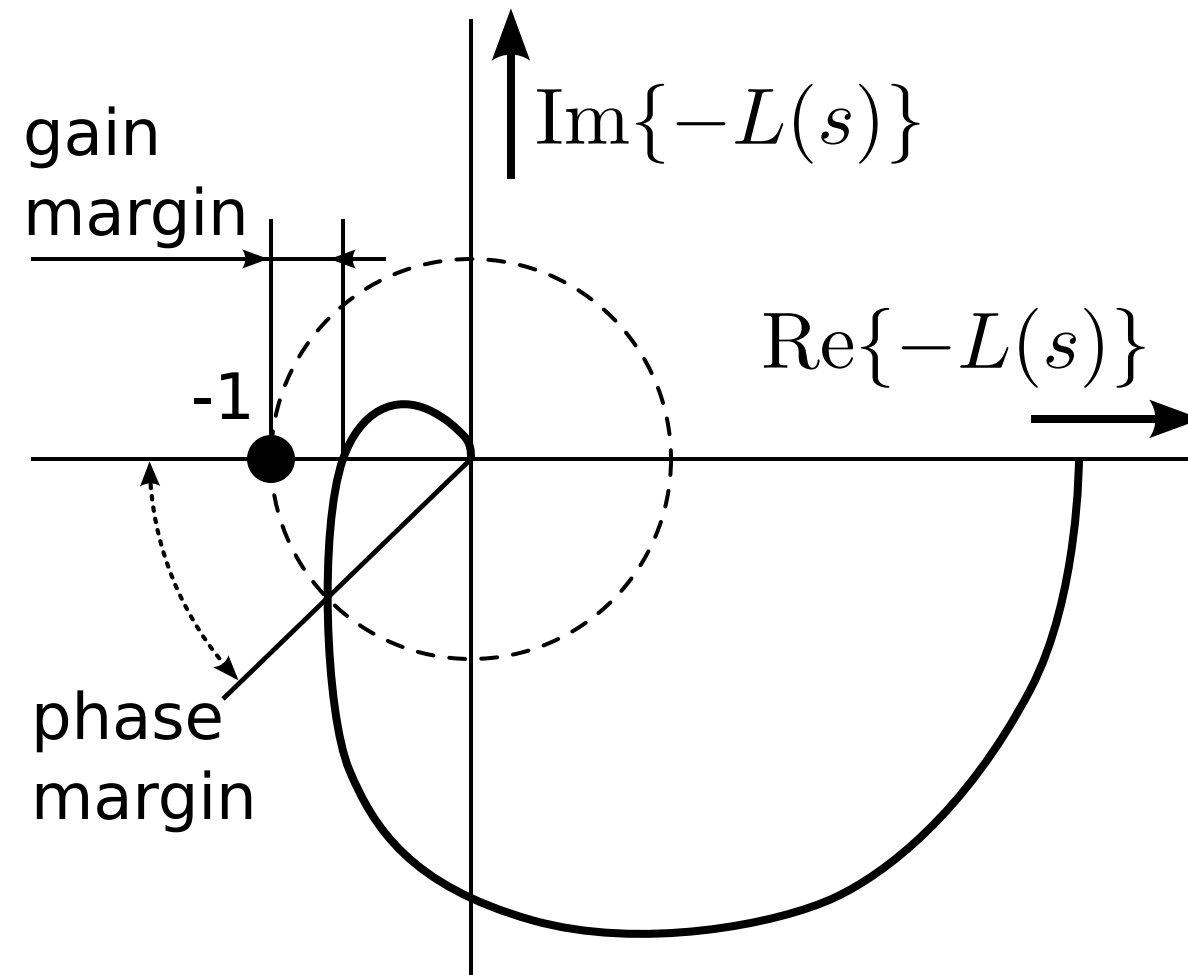
How far is the system from instability ...

Gain margin: how much in dB the loop gain must be increased to cause instability

Phase margin: how much in degrees the phase lag of the loop gain must be increased to cause instability



# Gain margin and phase margin



Indicative measure for stability

How far is the system from instability ...

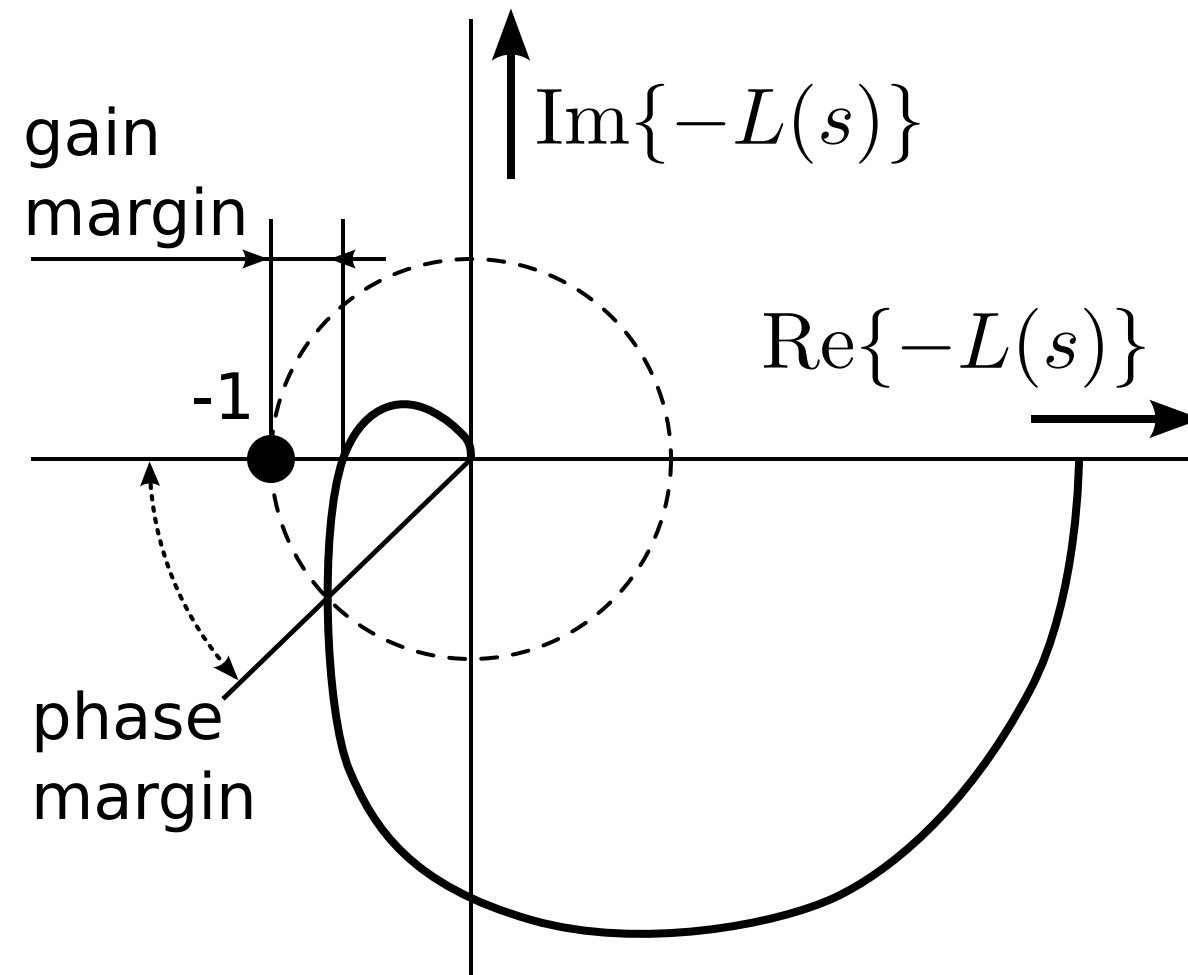
Gain margin: how much in dB the loop gain must be increased to cause instability

Phase margin: how much in degrees the phase lag of the loop gain must be increased to cause instability

Rules of thumb:



# Gain margin and phase margin



Indicative measure for stability

How far is the system from instability ...

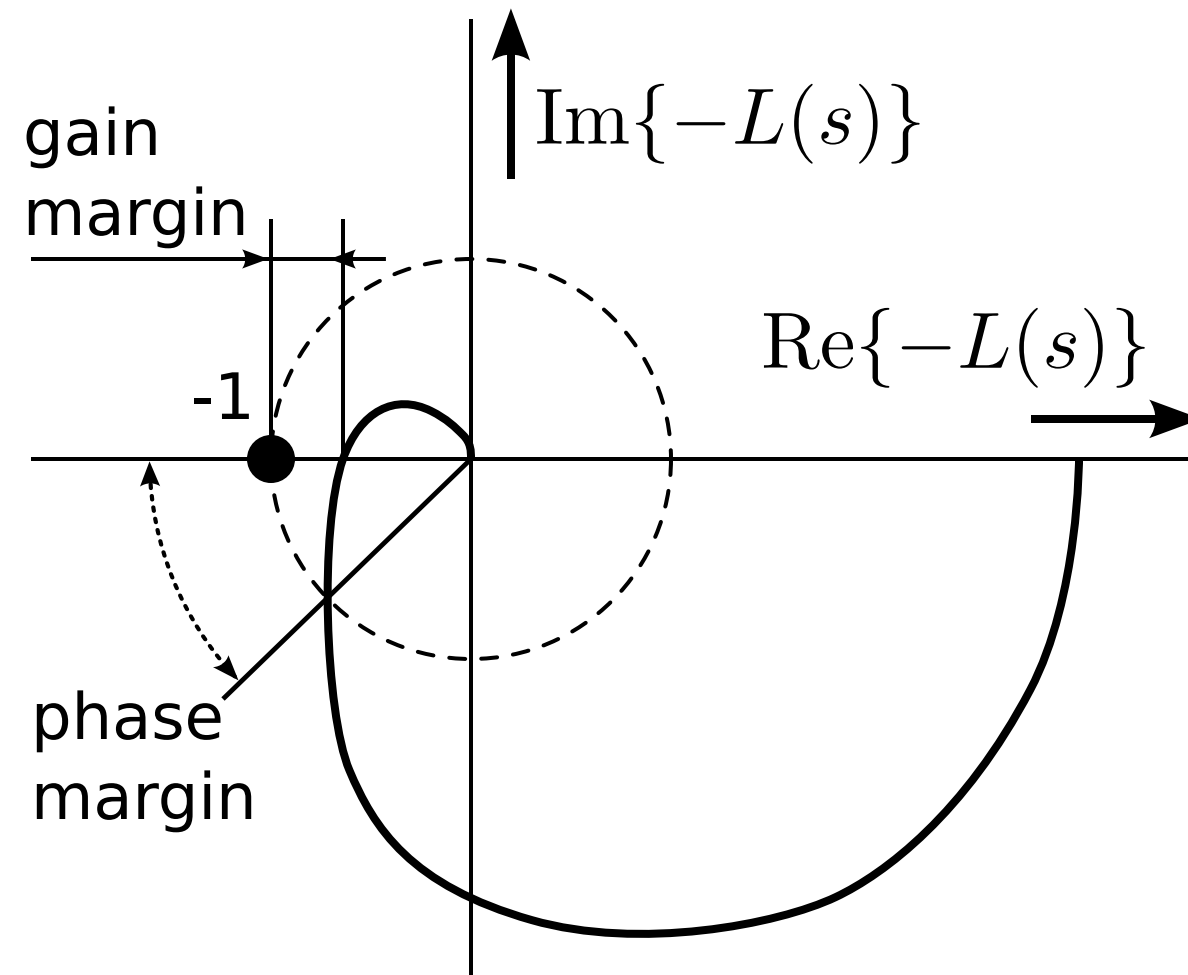
Gain margin: how much in dB the loop gain must be increased to cause instability

Phase margin: how much in degrees the phase lag of the loop gain must be increased to cause instability

Rules of thumb:

1. Gain margin better than 6 dB

# Gain margin and phase margin



Indicative measure for stability

How far is the system from instability ...

Gain margin: how much in dB the loop gain must be increased to cause instability

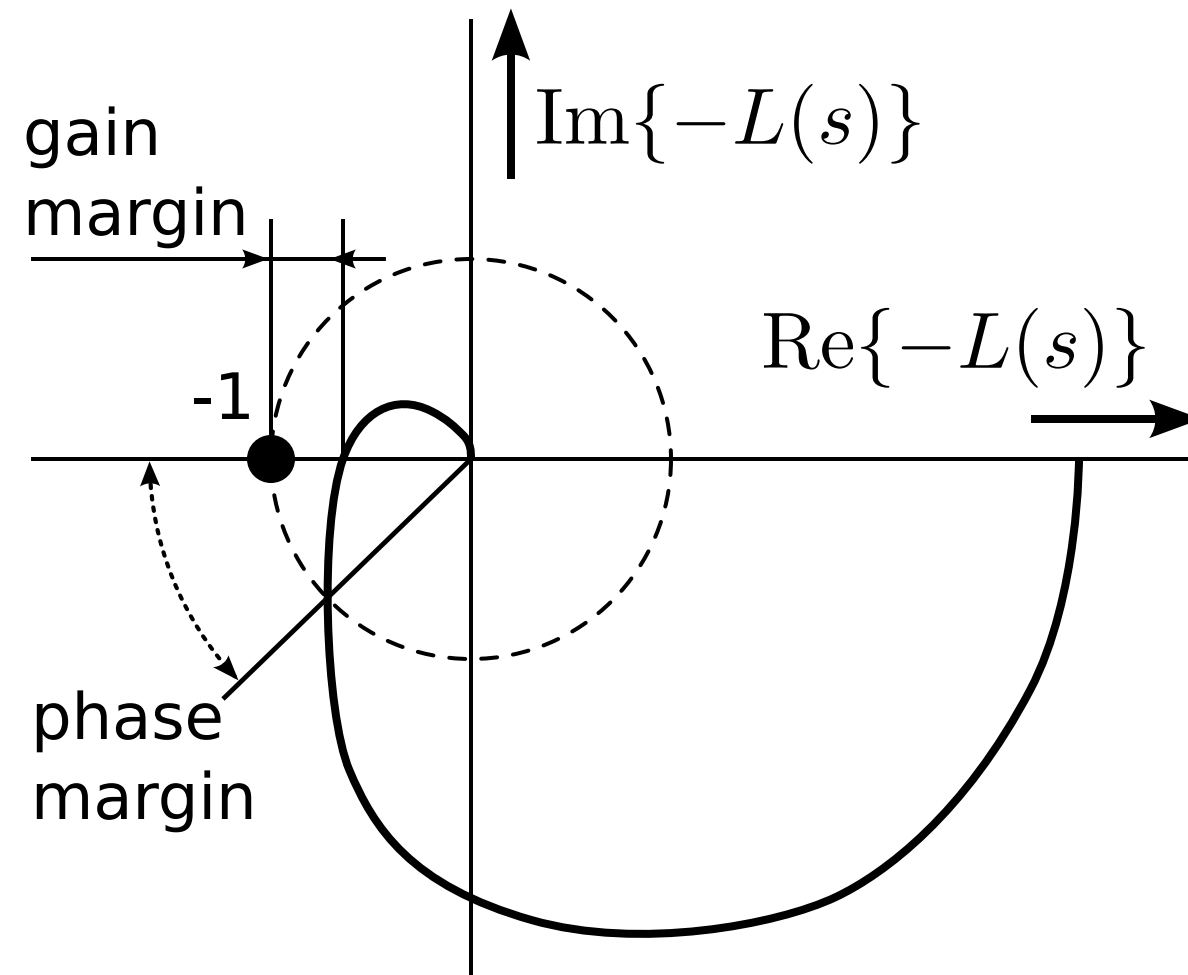
Phase margin: how much in degrees the phase lag of the loop gain must be increased to cause instability

Rules of thumb:

1. Gain margin better than 6 dB
2. Phase margin better than 30 degrees

Warnings:

# Gain margin and phase margin



Indicative measure for stability

How far is the system from instability ...

Gain margin: how much in dB the loop gain must be increased to cause instability

Phase margin: how much in degrees the phase lag of the loop gain must be increased to cause instability

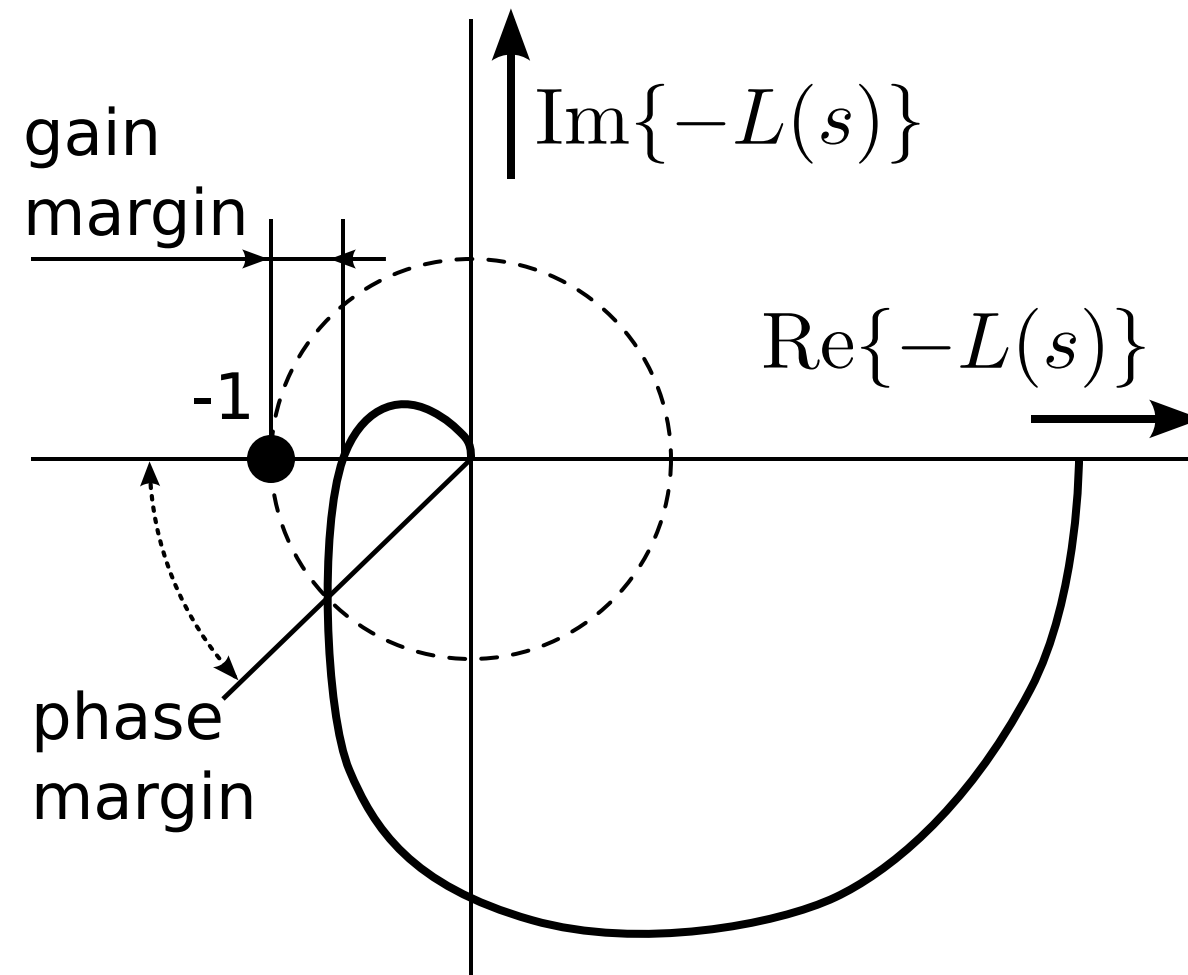
Rules of thumb:

1. Gain margin better than 6 dB
2. Phase margin better than 30 degrees

Warnings:

1. No one-to-one mapping on frequency response

# Gain margin and phase margin



Indicative measure for stability

How far is the system from instability ...

Gain margin: how much in dB the loop gain must be increased to cause instability

Phase margin: how much in degrees the phase lag of the loop gain must be increased to cause instability

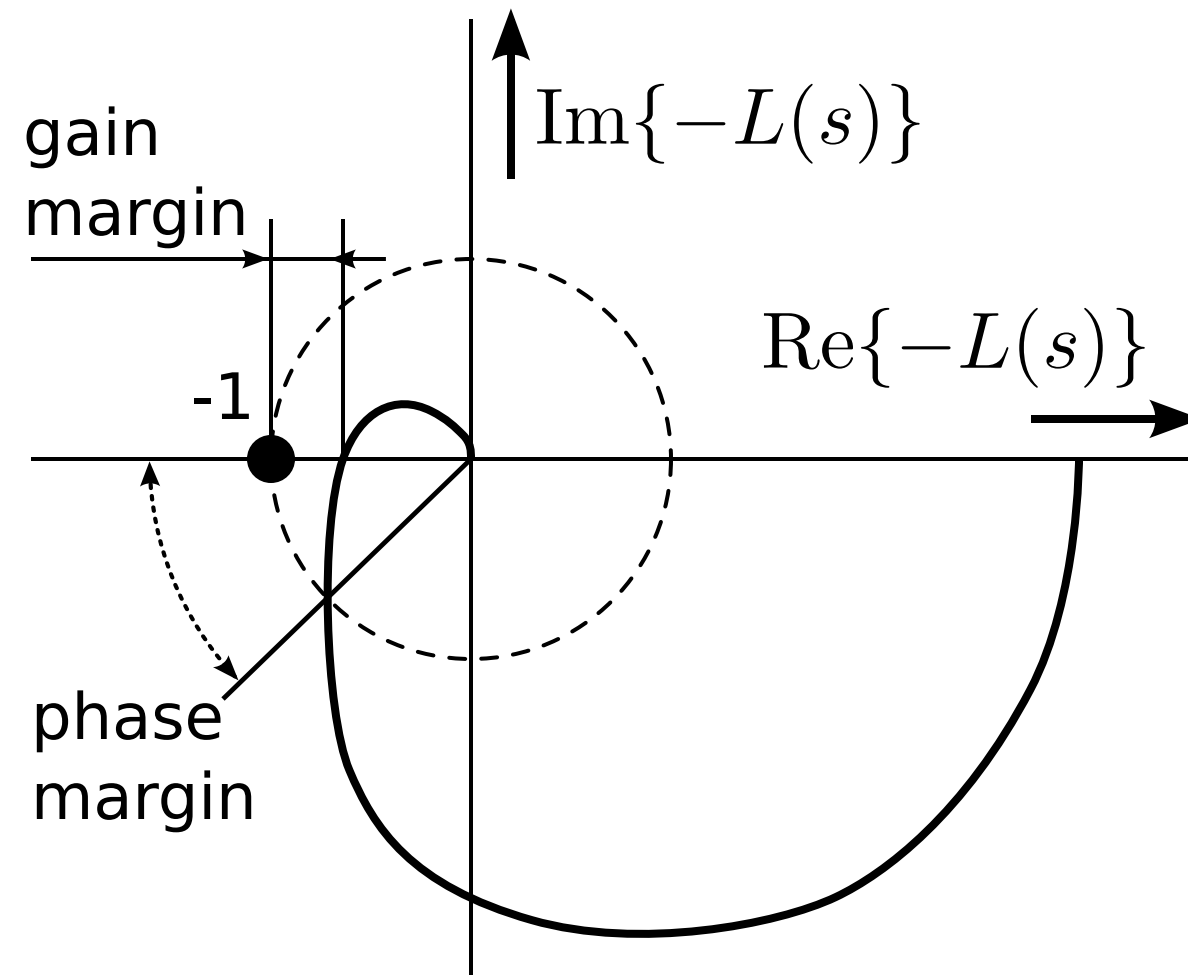
Rules of thumb:

1. Gain margin better than 6 dB
2. Phase margin better than 30 degrees

Warnings:

1. No one-to-one mapping on frequency response
2. Property of loop gain

# Gain margin and phase margin



Indicative measure for stability

How far is the system from instability ...

Gain margin: how much in dB the loop gain must be increased to cause instability

Phase margin: how much in degrees the phase lag of the loop gain must be increased to cause instability

Rules of thumb:

1. Gain margin better than 6 dB
2. Phase margin better than 30 degrees

Warnings:

1. No one-to-one mapping on frequency response
2. Property of loop gain

# Root locus technique

# Root locus technique

1. Introduced by Evans in 1948

# Root locus technique

1. Introduced by Evans in 1948
2. Graphical method to find poles from the servo function from the poles of the loop gain



# Root locus technique

1. Introduced by Evans in 1948
2. Graphical method to find poles from the servo function from the poles of the loop gain
3. 10 construction rules (see book)

# Root locus technique

1. Introduced by Evans in 1948
2. Graphical method to find poles from the servo function from the poles of the loop gain
3. 10 construction rules (see book)
4. For proof refer to control theory

# Root locus technique

1. Introduced by Evans in 1948
2. Graphical method to find poles from the servo function from the poles of the loop gain
3. 10 construction rules (see book)
4. For proof refer to control theory
5. Here: examples with finite, nonzero DC loop gain

# Root locus technique

1. Introduced by Evans in 1948
2. Graphical method to find poles from the servo function from the poles of the loop gain
3. 10 construction rules (see book)
4. For proof refer to control theory
5. Here: examples with finite, nonzero DC loop gain
6. Implemented in SLiCAP

# Root locus technique

1. Introduced by Evans in 1948
2. Graphical method to find poles from the servo function from the poles of the loop gain
3. 10 construction rules (see book)
4. For proof refer to control theory
5. Here: examples with finite, nonzero DC loop gain
6. Implemented in SLiCAP

# Root locus technique

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain



# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)}$$

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

**Root locus:**

Paths traced out by the poles of the servo function while varying the DC loop gain

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:



# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

**Root locus:**

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{\text{DC}}$

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{\text{DC}}$

1. Number of branches equals number of poles of the loop gain

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{\text{DC}}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis



# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{\text{DC}}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain

# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity

# Root locus technique

$$L(s) = L_{\text{DC}} \frac{N(s)}{D(s)}$$

$L_{\text{DC}}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{\text{DC}}N(s)}{D(s)-L_{\text{DC}}N(s)}$$

$D(s) - L_{\text{DC}}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{\text{DC}}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch

# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6.  $n$  poles and  $m$  zeros, then  $n - m$  asymptotes

# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6. n poles and m zeros, then n - m asymptotes
7. Real axis intersection point asymptotes

# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6. n poles and m zeros, then n - m asymptotes
7. Real axis intersection point asymptotes

$$\sigma = \frac{\sum_{k=1}^n p_k - \sum_{i=1}^m z_i}{n-m}$$



# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6. n poles and m zeros, then n - m asymptotes
7. Real axis intersection point asymptotes

$$\sigma = \frac{\sum_{k=1}^n p_k - \sum_{i=1}^m z_i}{n-m}$$

8. Angle asymptotes equally spaced

# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6. n poles and m zeros, then n - m asymptotes
7. Real axis intersection point asymptotes

$$\sigma = \frac{\sum_{k=1}^n p_k - \sum_{i=1}^m z_i}{n-m}$$

8. Angle asymptotes equally spaced
9. Break away (and arrival) points



# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6. n poles and m zeros, then n - m asymptotes
7. Real axis intersection point asymptotes

$$\sigma = \frac{\sum_{k=1}^n p_k - \sum_{i=1}^m z_i}{n-m}$$

8. Angle asymptotes equally spaced
9. Break away (and arrival) points

$$\frac{d}{ds} L(s) = 0$$

10. Break away angles equally spaced

# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6. n poles and m zeros, then n - m asymptotes
7. Real axis intersection point asymptotes

$$\sigma = \frac{\sum_{k=1}^n p_k - \sum_{i=1}^m z_i}{n-m}$$

8. Angle asymptotes equally spaced
9. Break away (and arrival) points

$$\frac{d}{ds} L(s) = 0$$

10. Break away angles equally spaced

# Root locus technique

$$L(s) = L_{DC} \frac{N(s)}{D(s)}$$

$L_{DC}$  = DC value of the loop gain

$N(s) = 0 \implies$  Zeros of the loop gain

$D(s) = 0 \implies$  Poles of the loop gain

$$S(s) = \frac{-L(s)}{1-L(s)} = \frac{-L_{DC}N(s)}{D(s)-L_{DC}N(s)}$$

$D(s) - L_{DC}N(s) = 0 \implies$  Poles of servo function

## Root locus:

Paths traced out by the poles of the servo function while varying the DC loop gain

Paths depend on:

1. Poles of the loop gain
2. Zeros of the loop gain
3. DC loop gain

Starting point: DC loop gain equals zero

End point: DC loop gain equals infinity

Actual servo poles: DC loop gain =  $L_{DC}$

1. Number of branches equals number of poles of the loop gain
2. Symmetrical with respect to the real axis
3. Branches start at poles of the loop gain
4. Branches end at zeros of the loop gain or at infinity
5. Parts of the real axis left from odd number of poles + zeros belong to a branch
6. n poles and m zeros, then n - m asymptotes
7. Real axis intersection point asymptotes

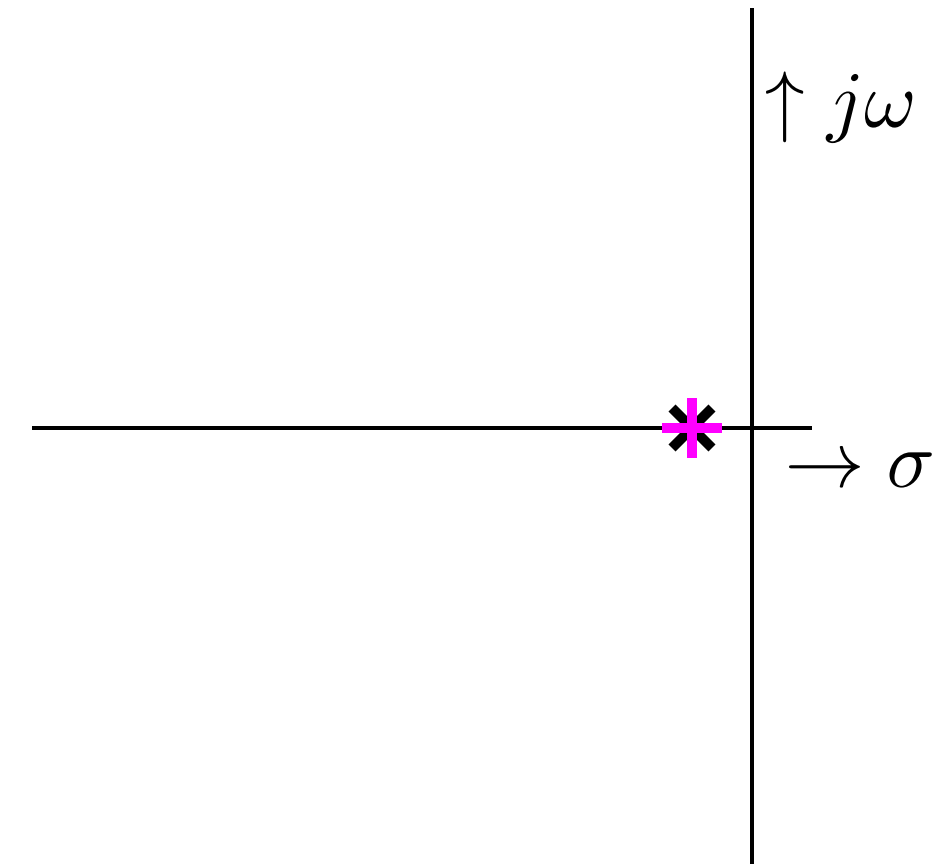
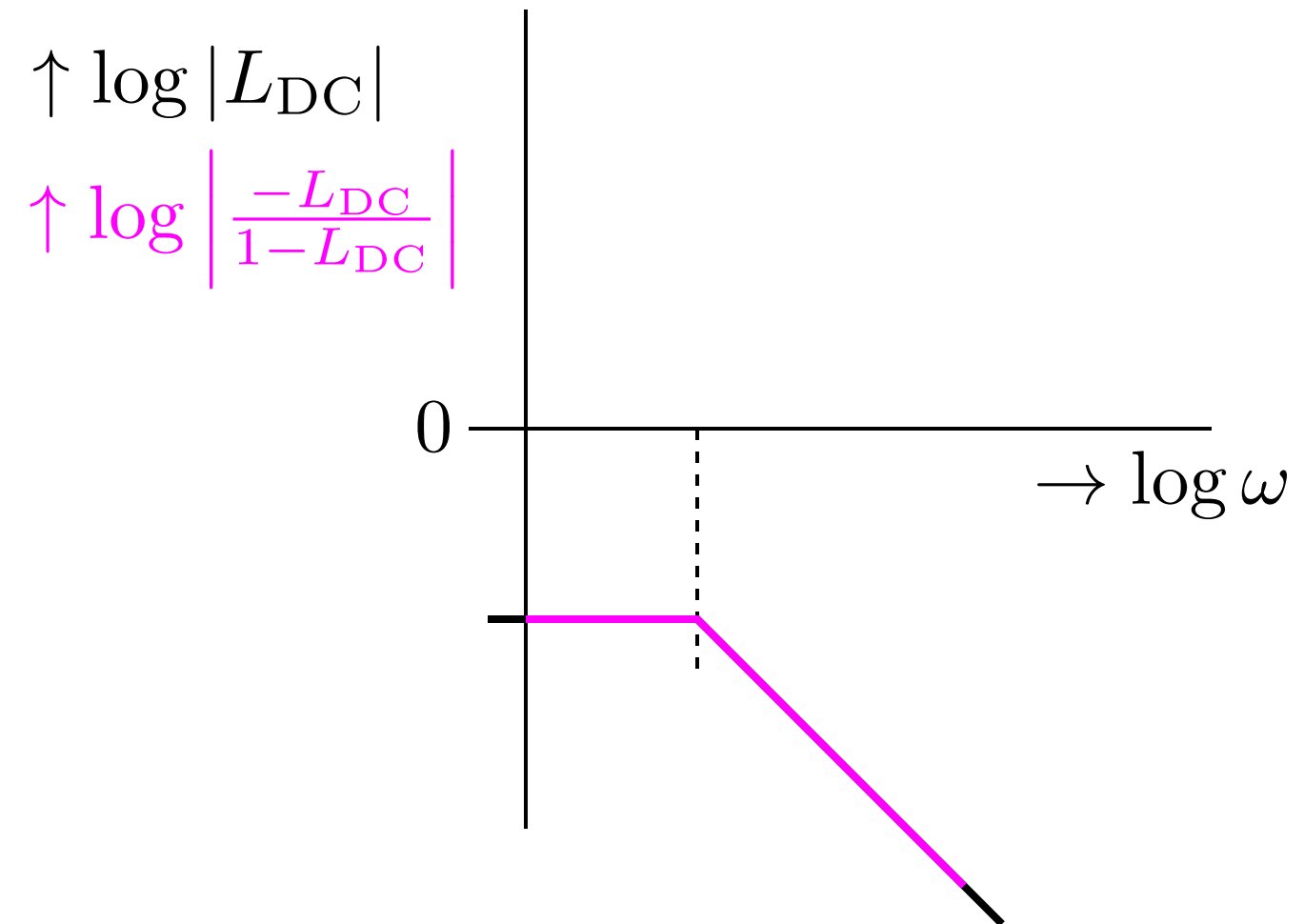
$$\sigma = \frac{\sum_{k=1}^n p_k - \sum_{i=1}^m z_i}{n-m}$$

8. Angle asymptotes equally spaced
9. Break away (and arrival) points

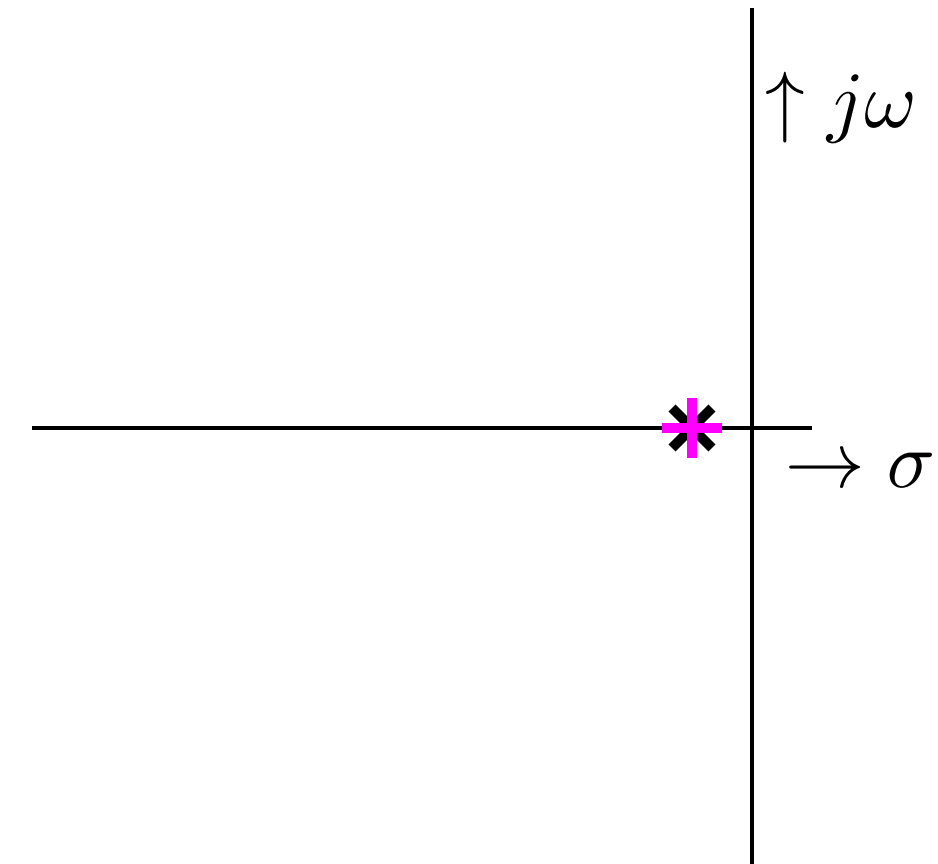
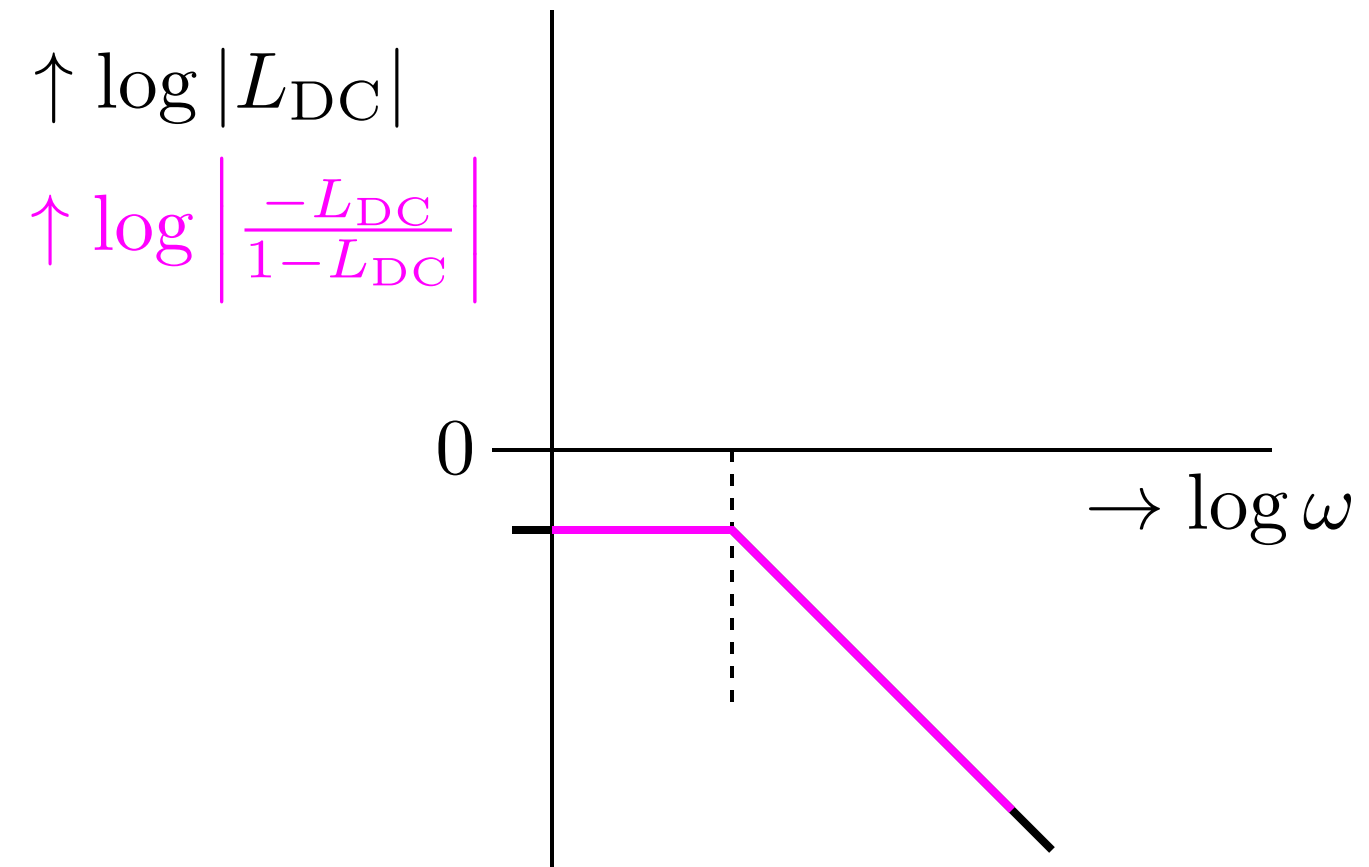
$$\frac{d}{ds} L(s) = 0$$

10. Break away angles equally spaced

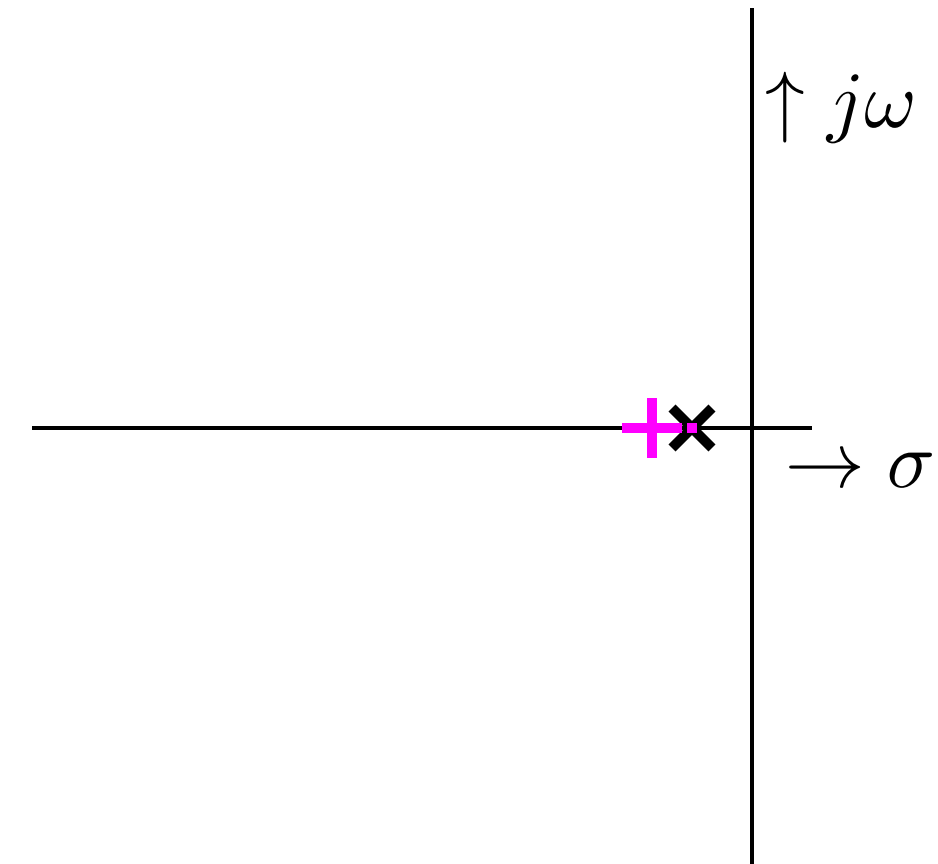
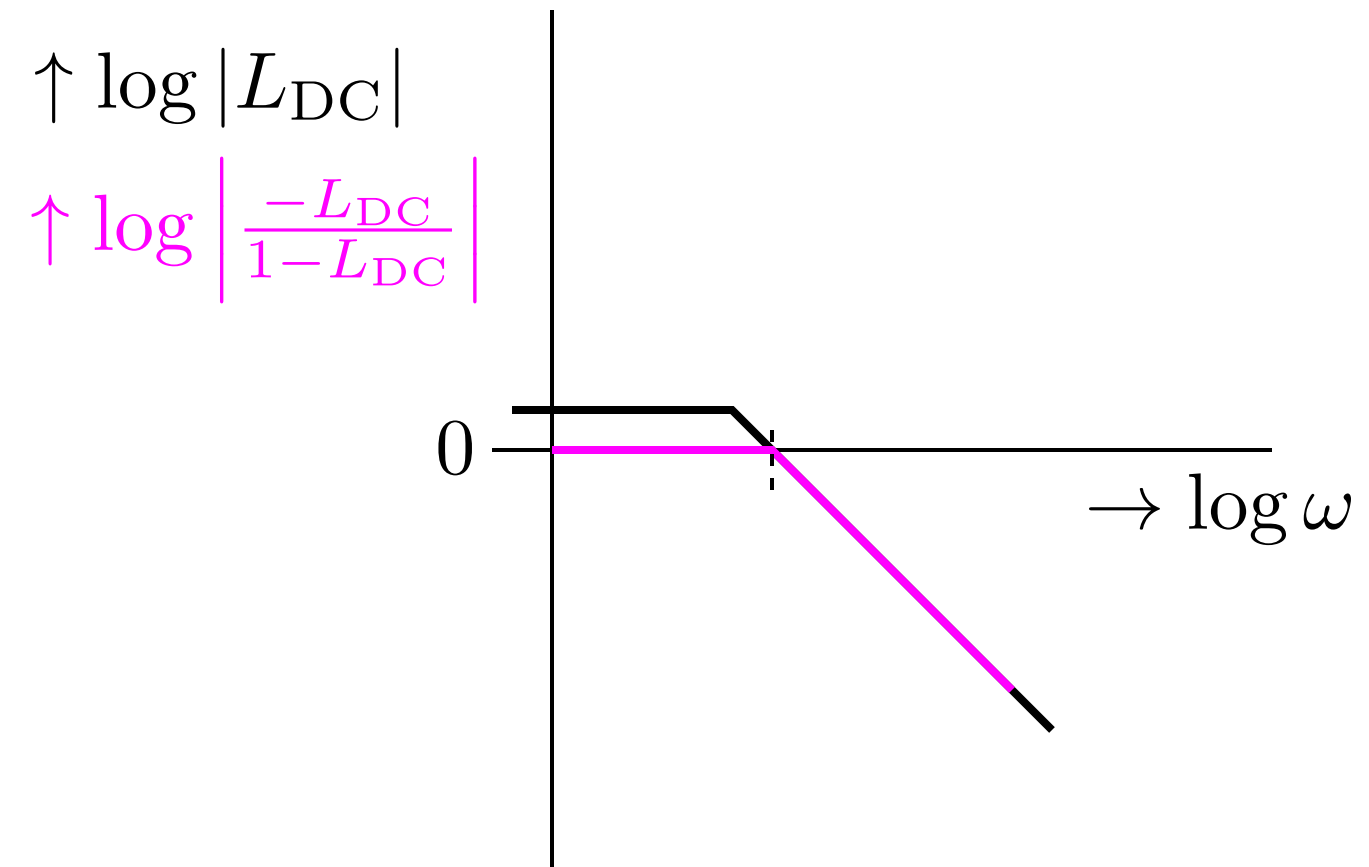
# Root locus first order



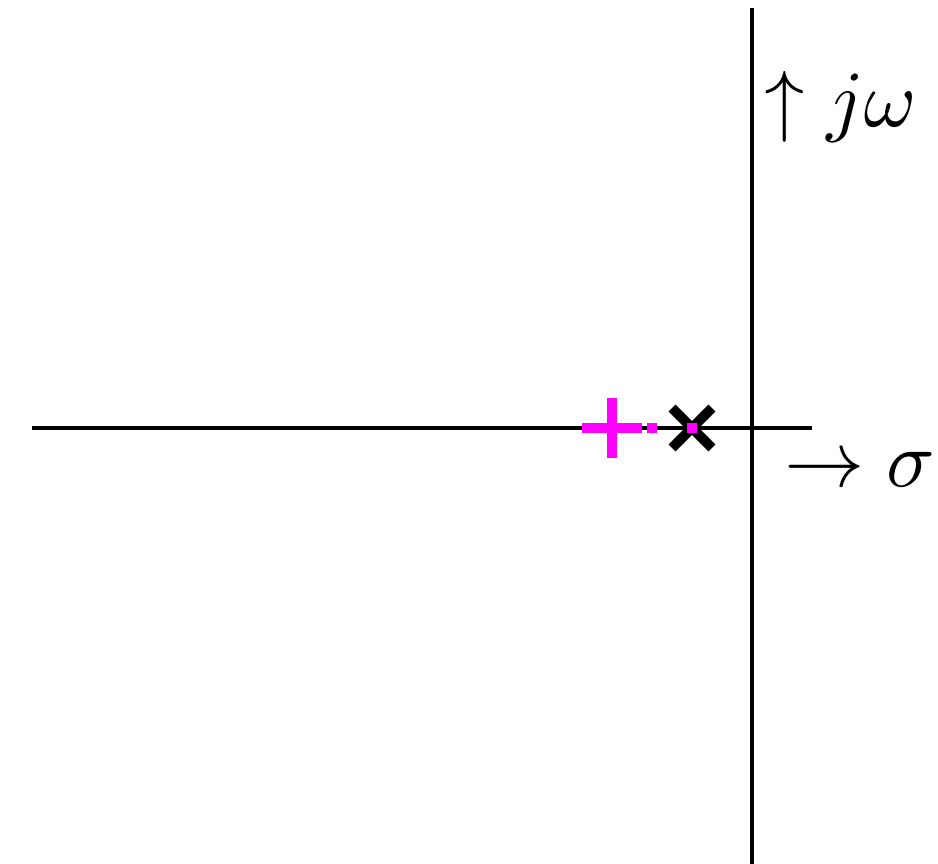
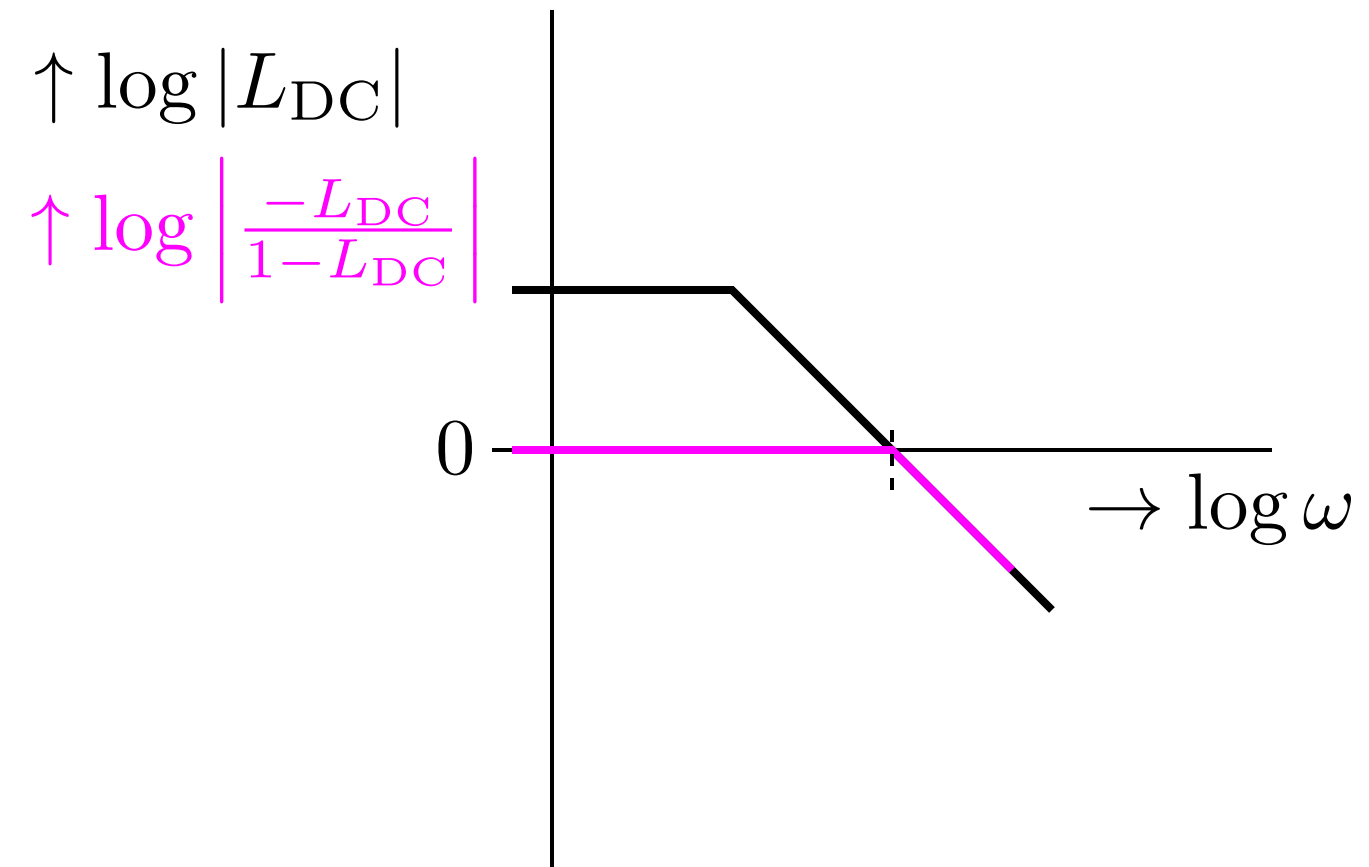
# Root locus first order



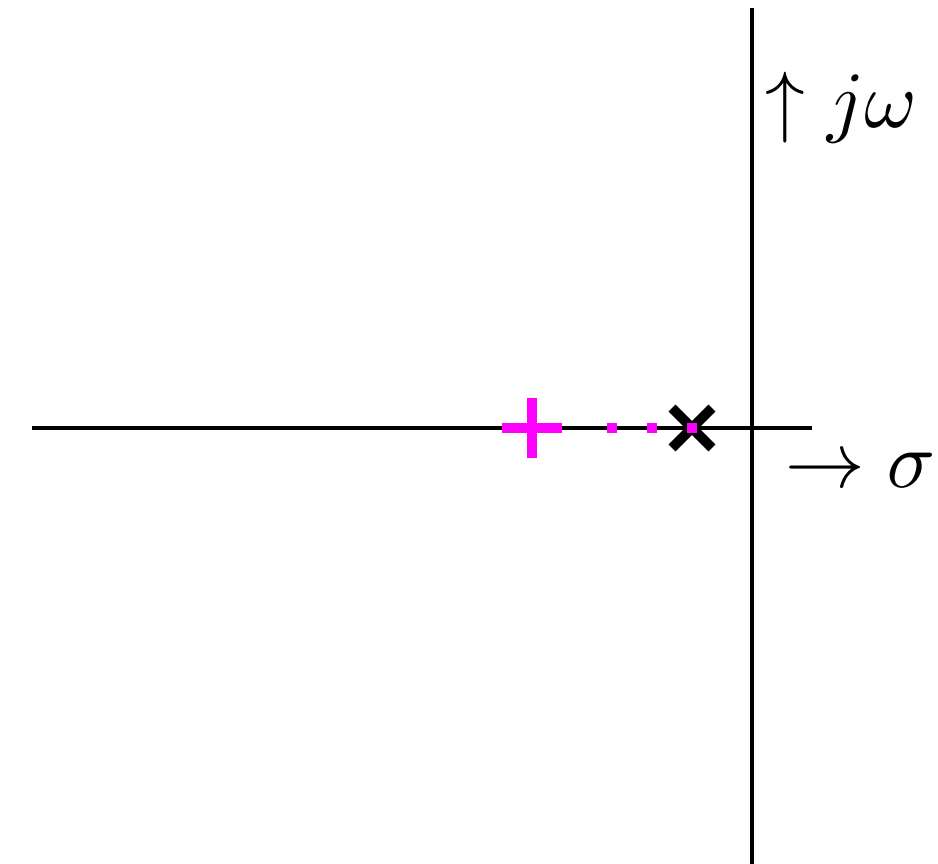
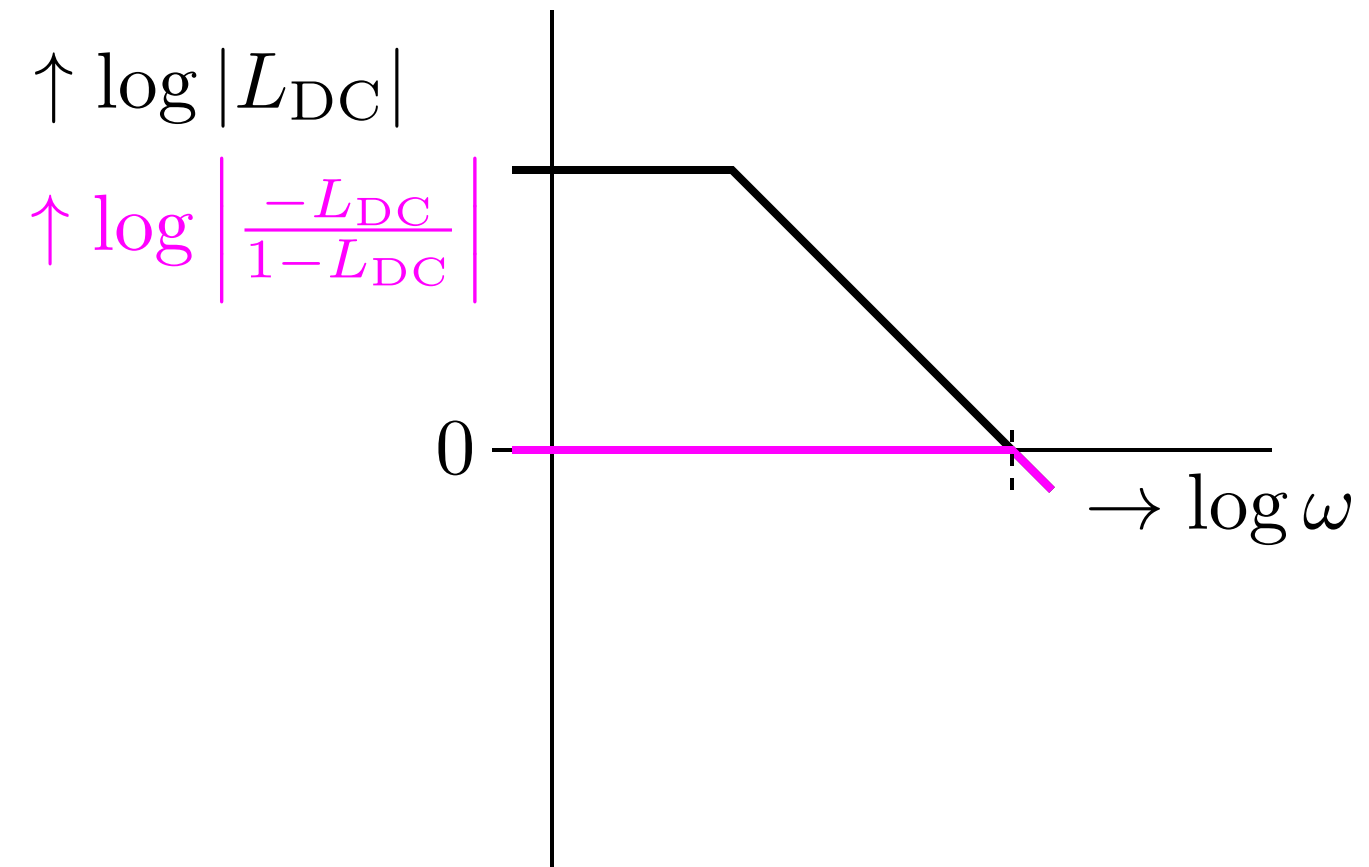
# Root locus first order



# Root locus first order

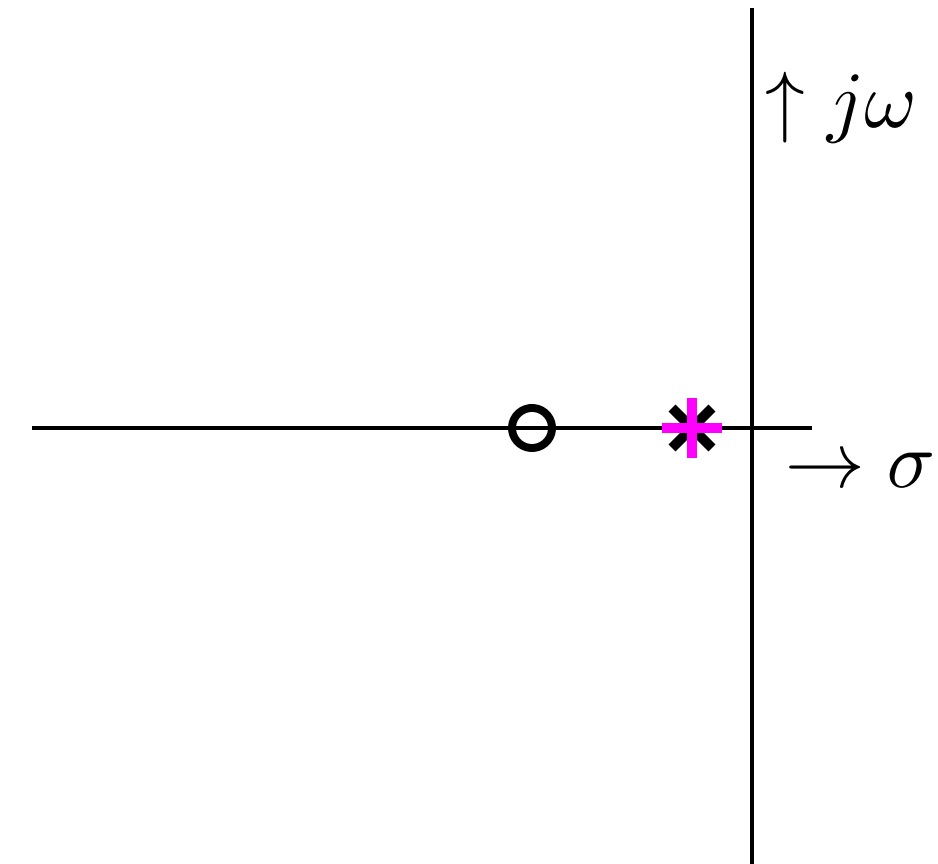
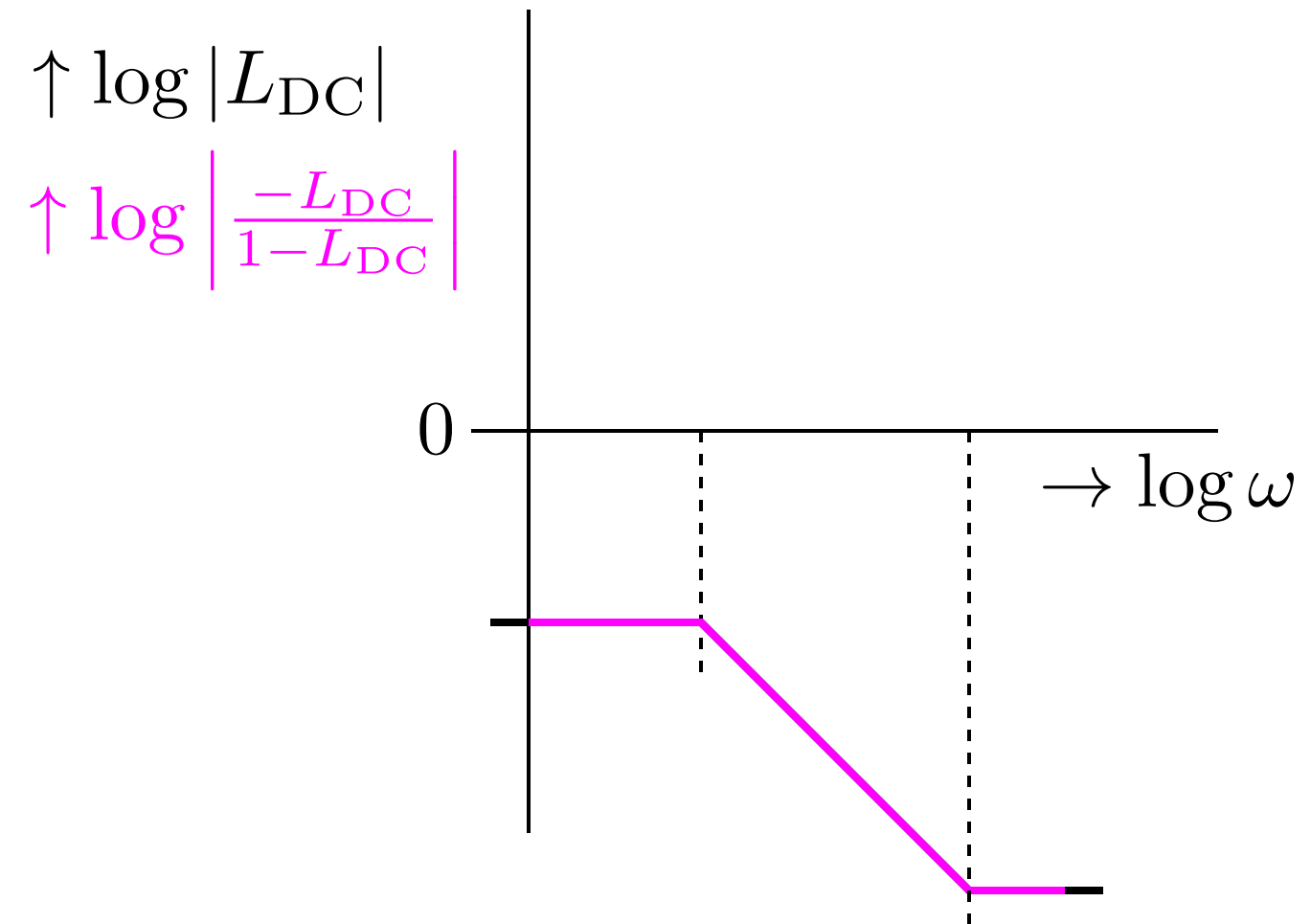


# Root locus first order

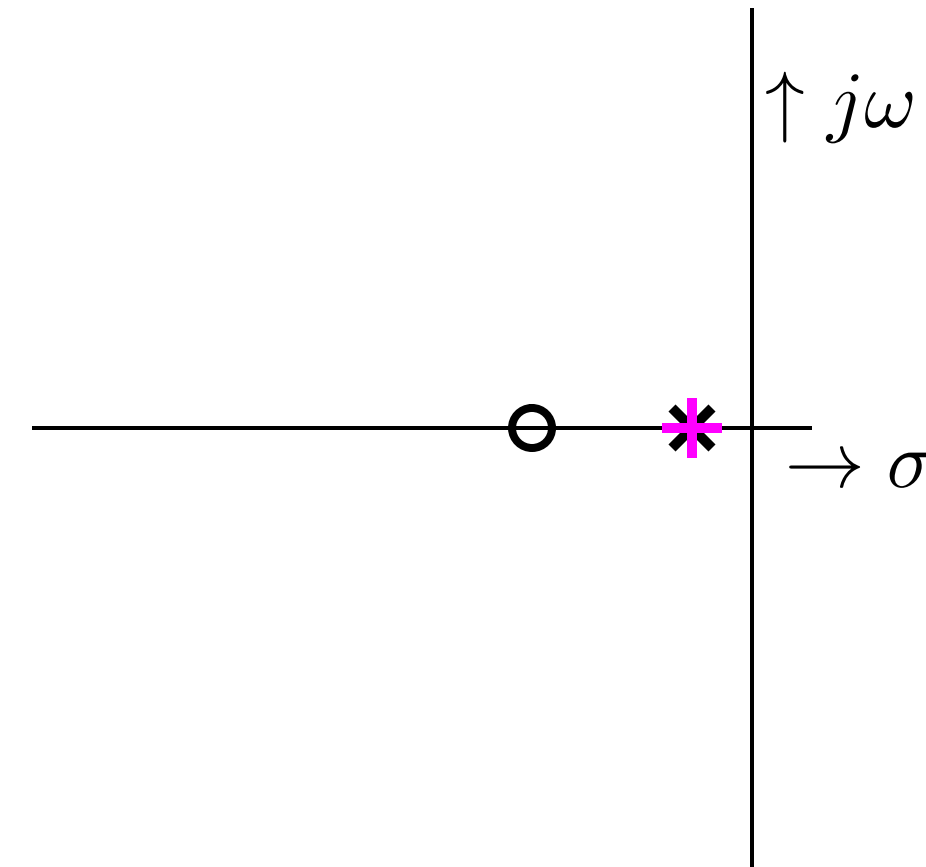
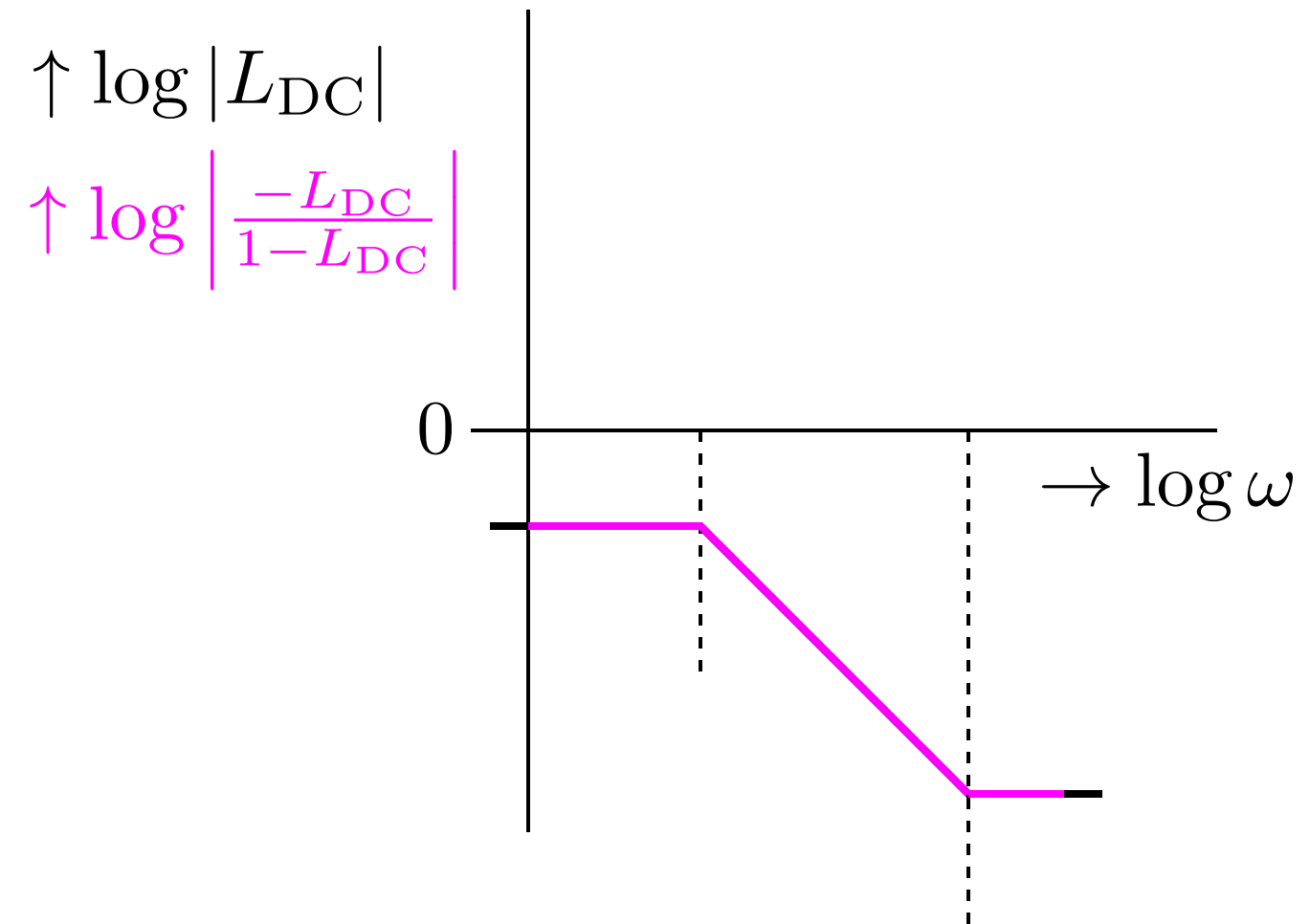




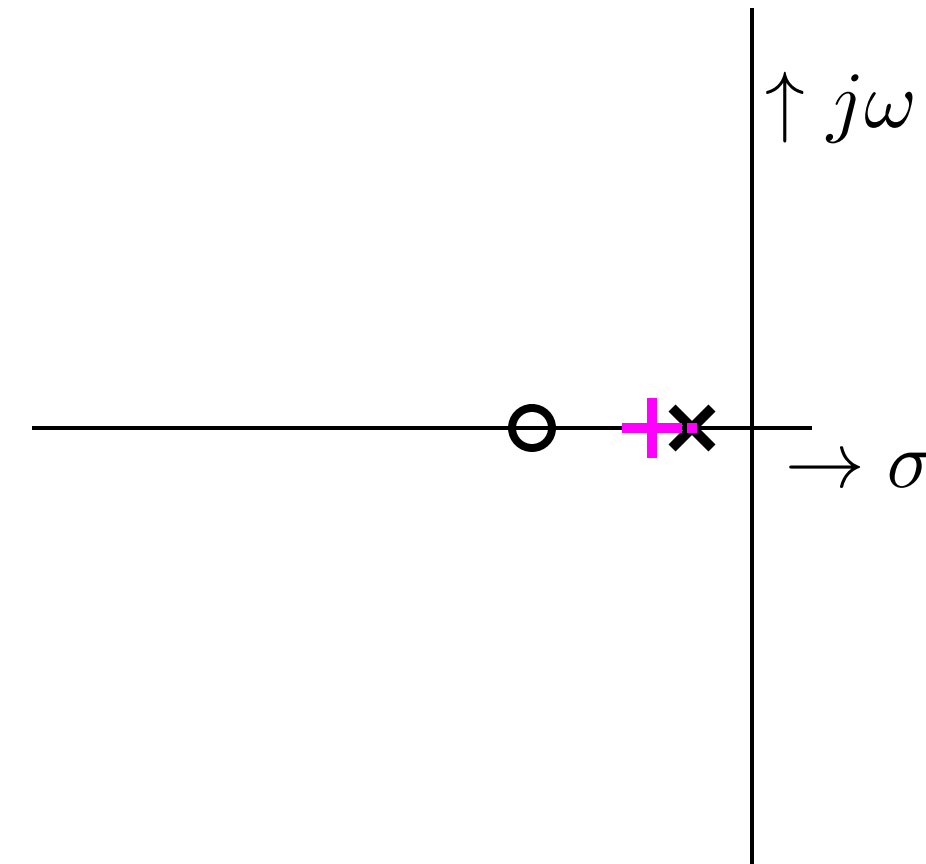
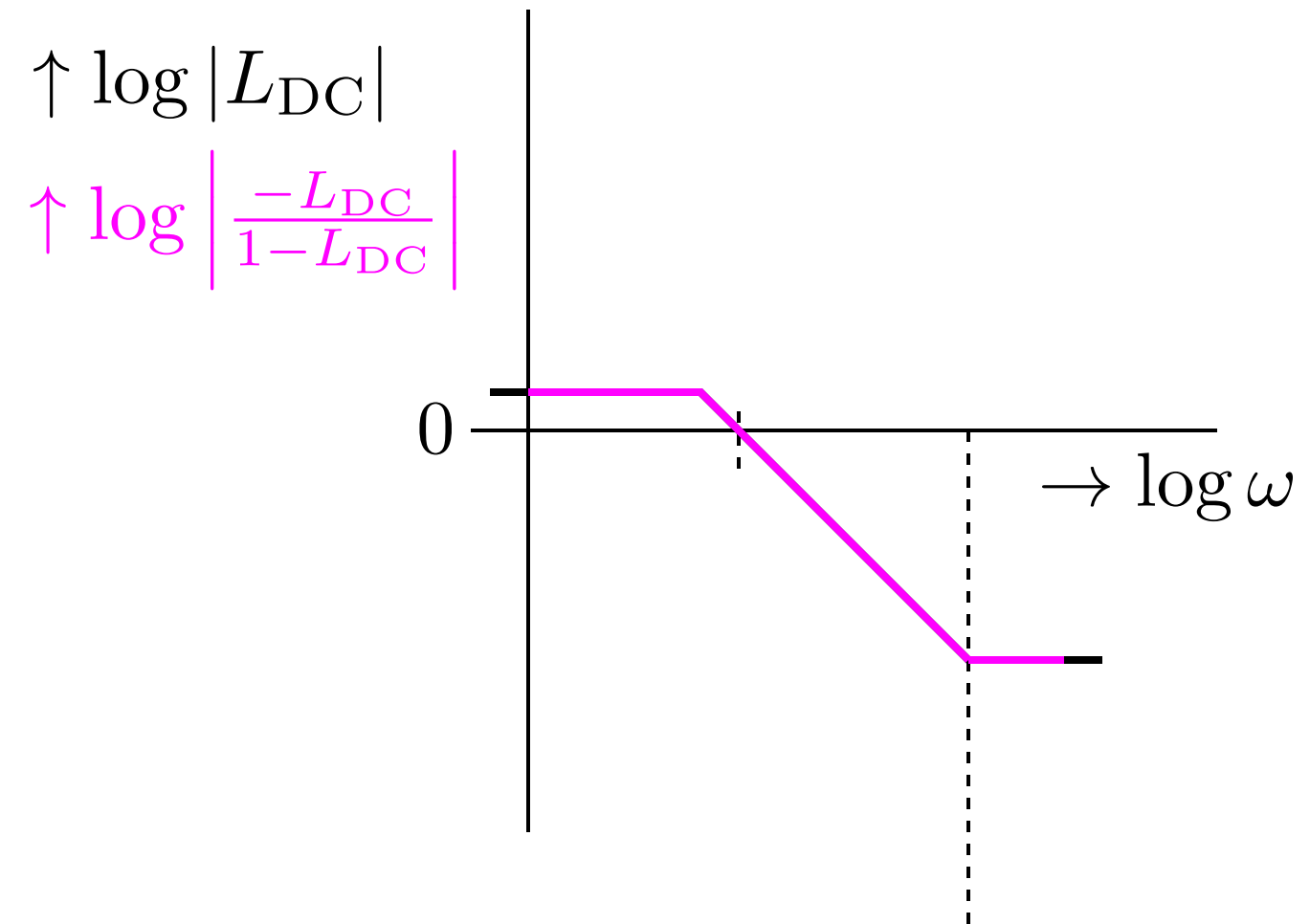
# Root locus first order with zero left



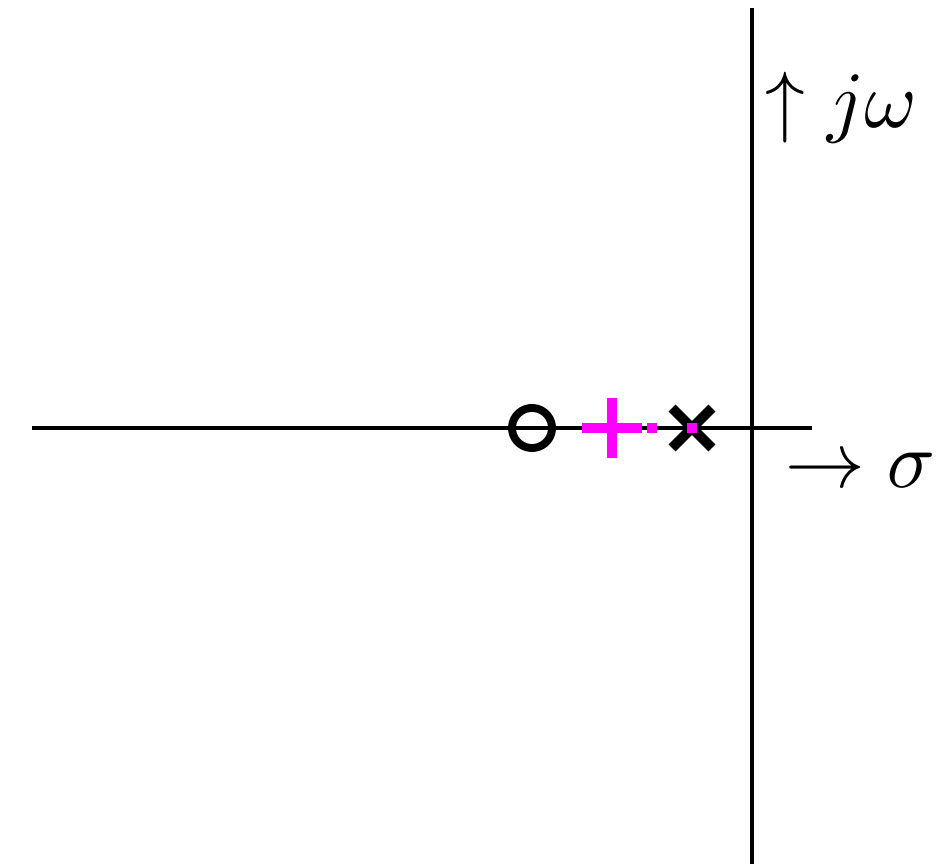
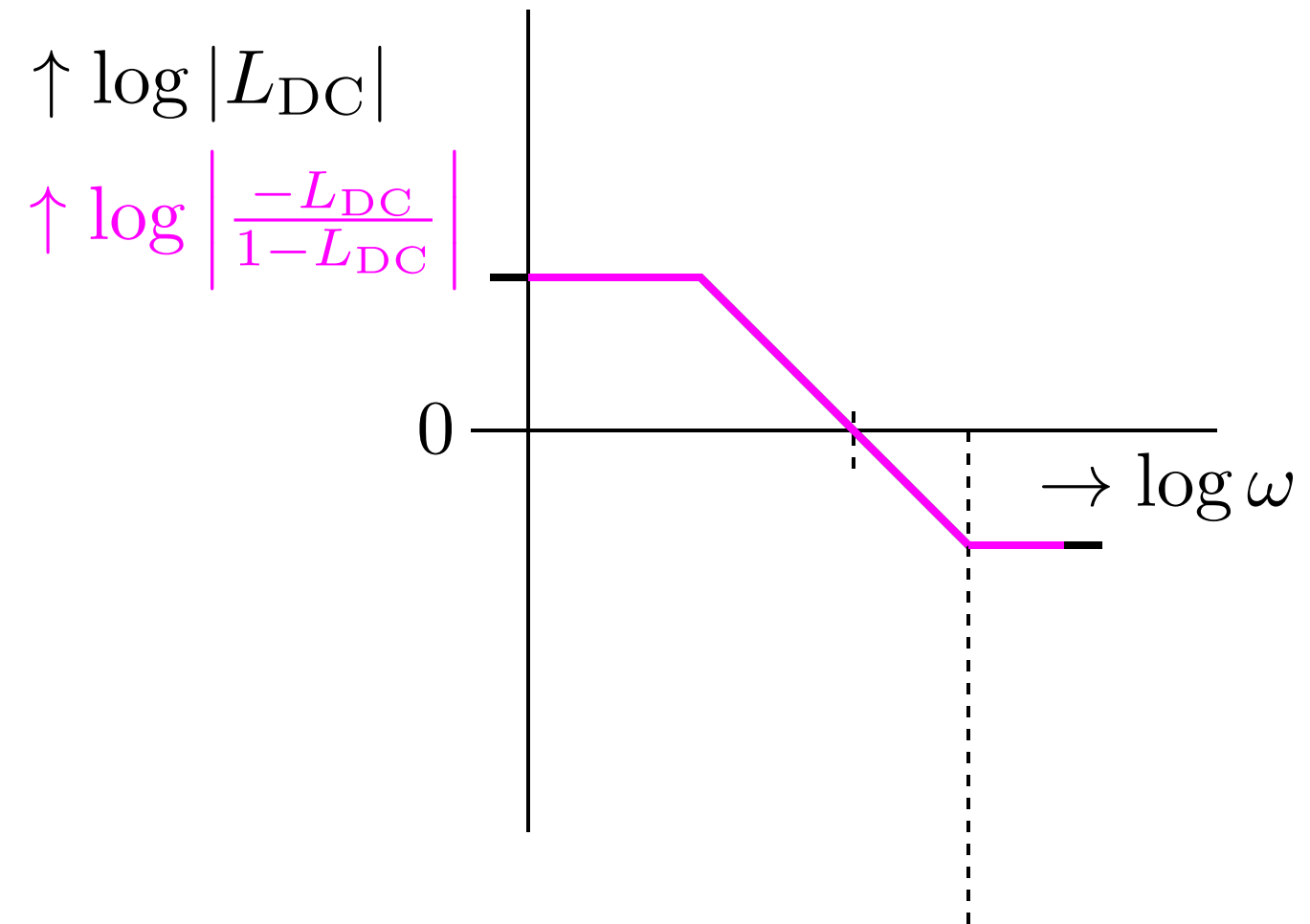
# Root locus first order with zero left



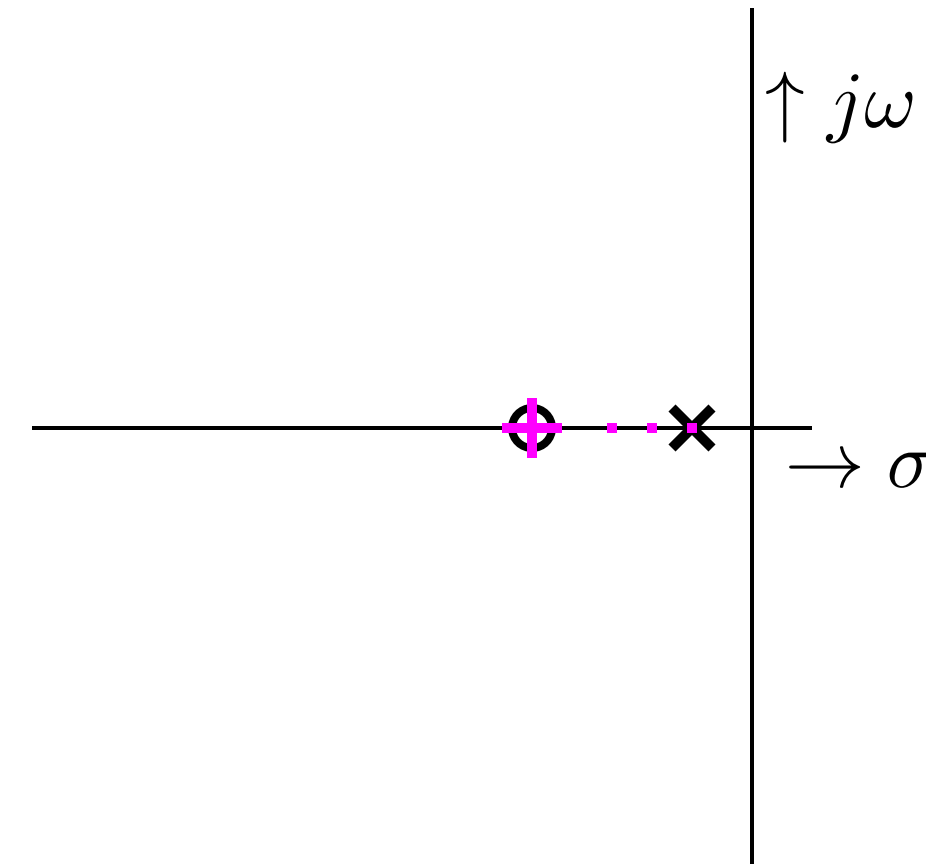
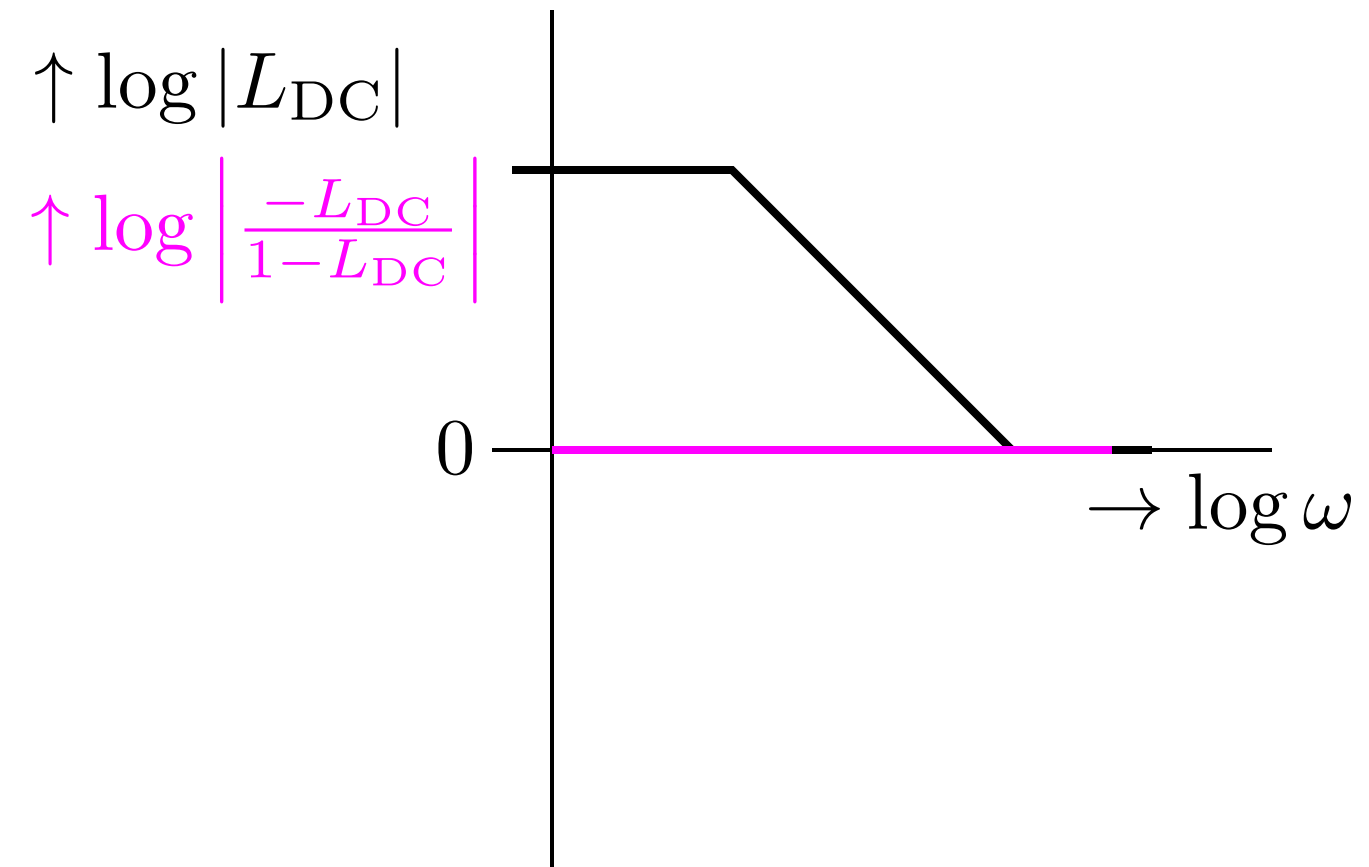
# Root locus first order with zero left



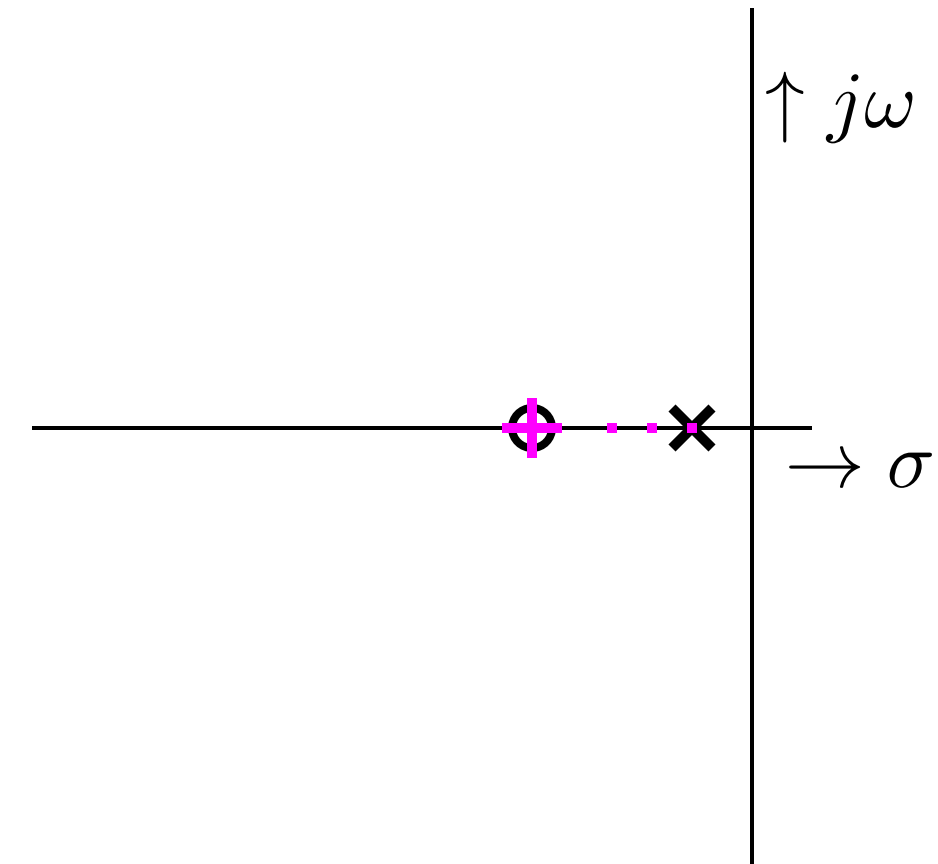
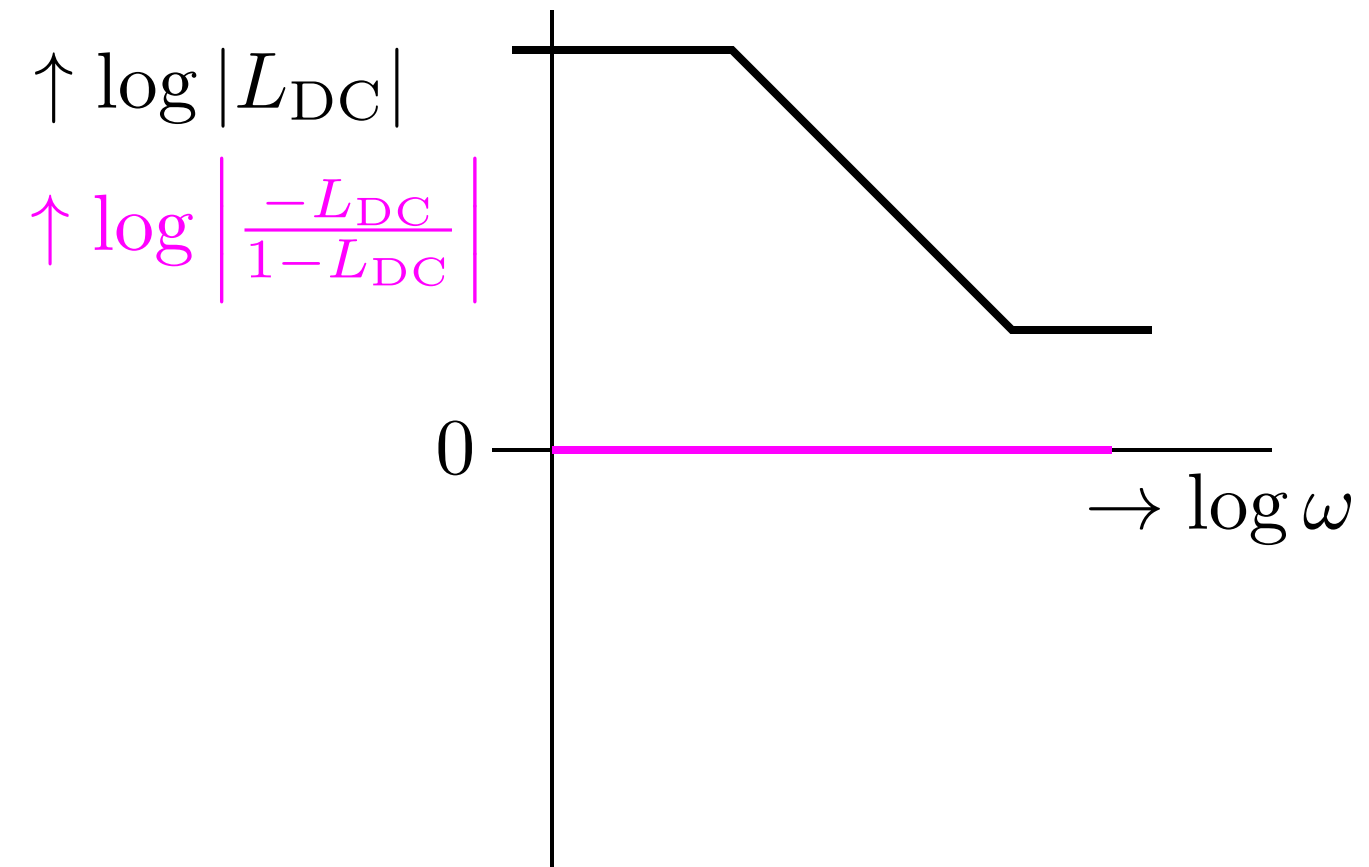
# Root locus first order with zero left



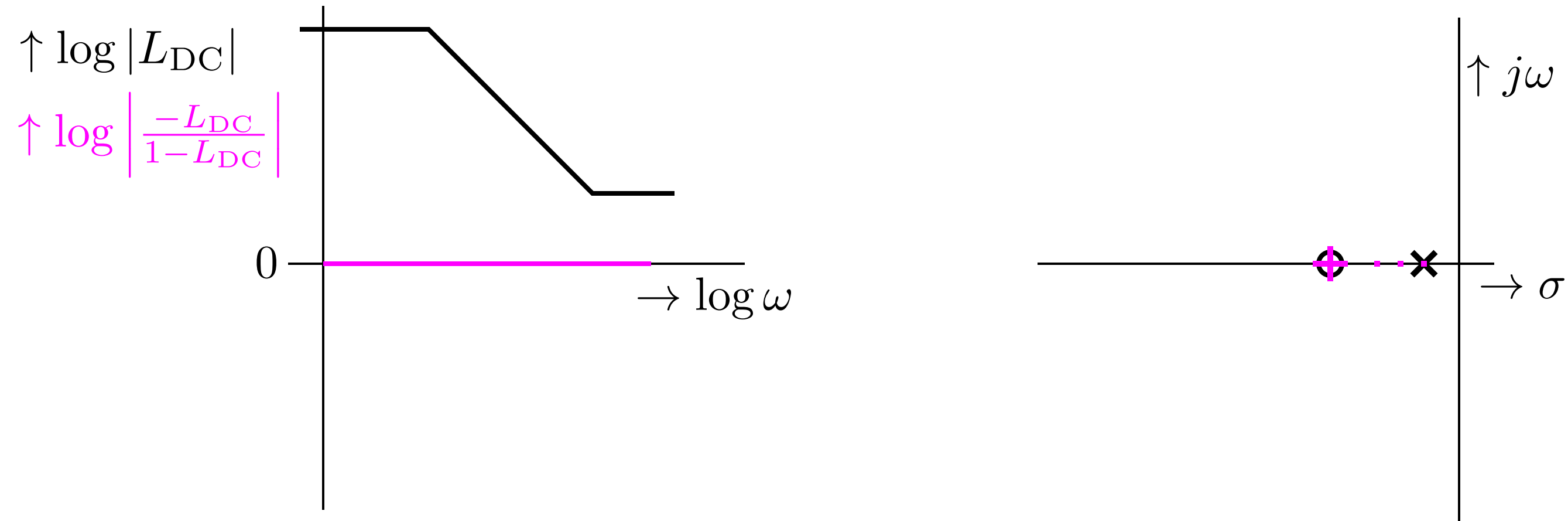
# Root locus first order with zero left



# Root locus first order with zero left

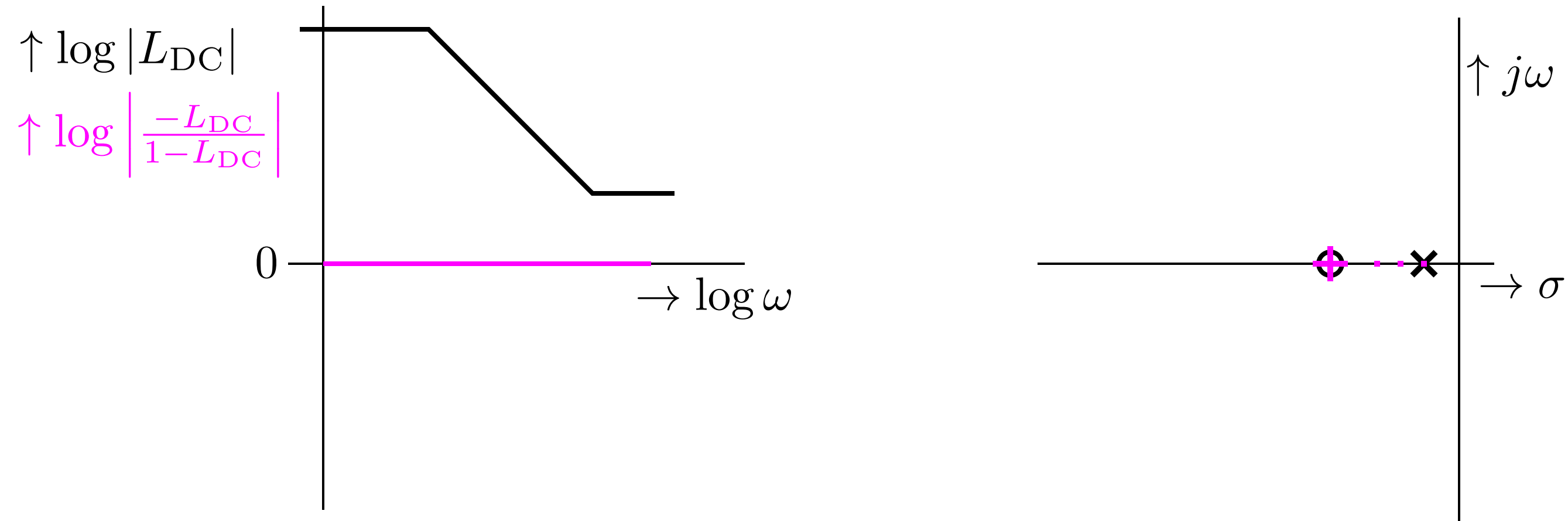


# Root locus first order with zero left



Note: pole only drops on the zero if DC loop gain is infinite!

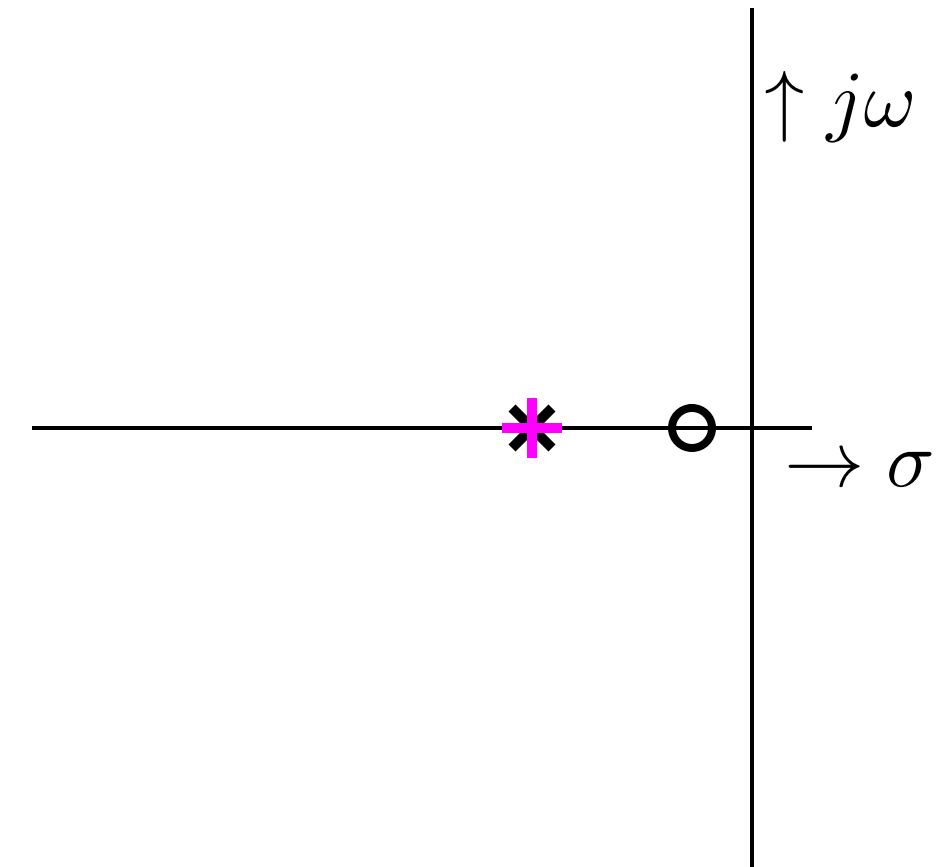
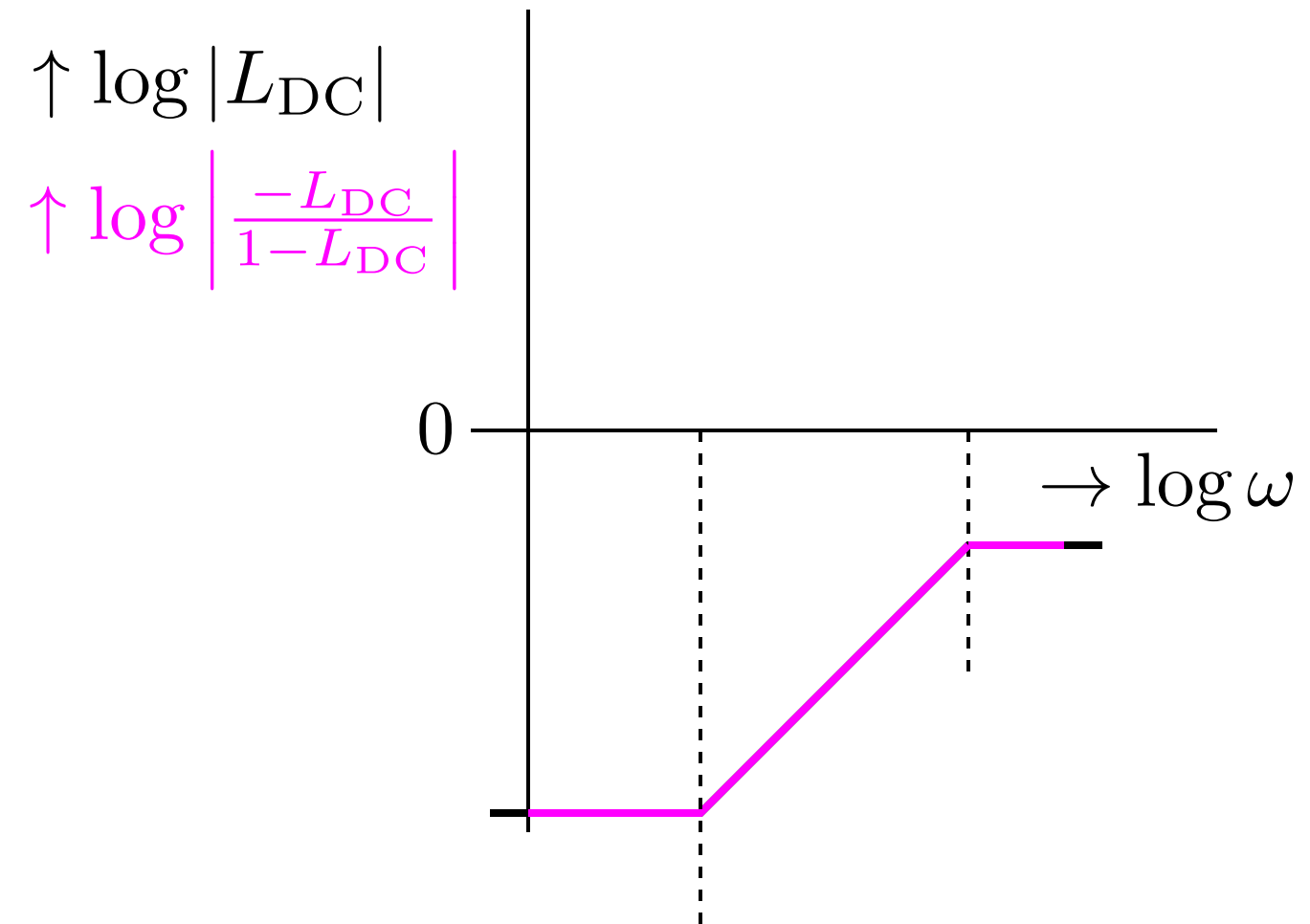
# Root locus first order with zero left



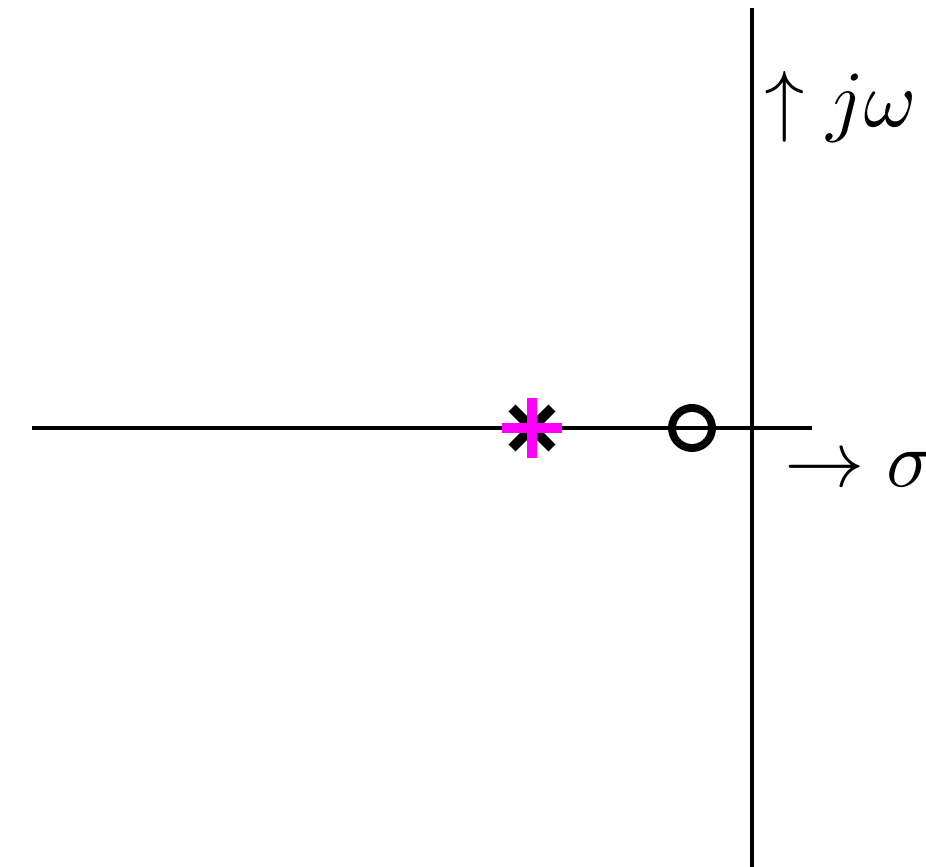
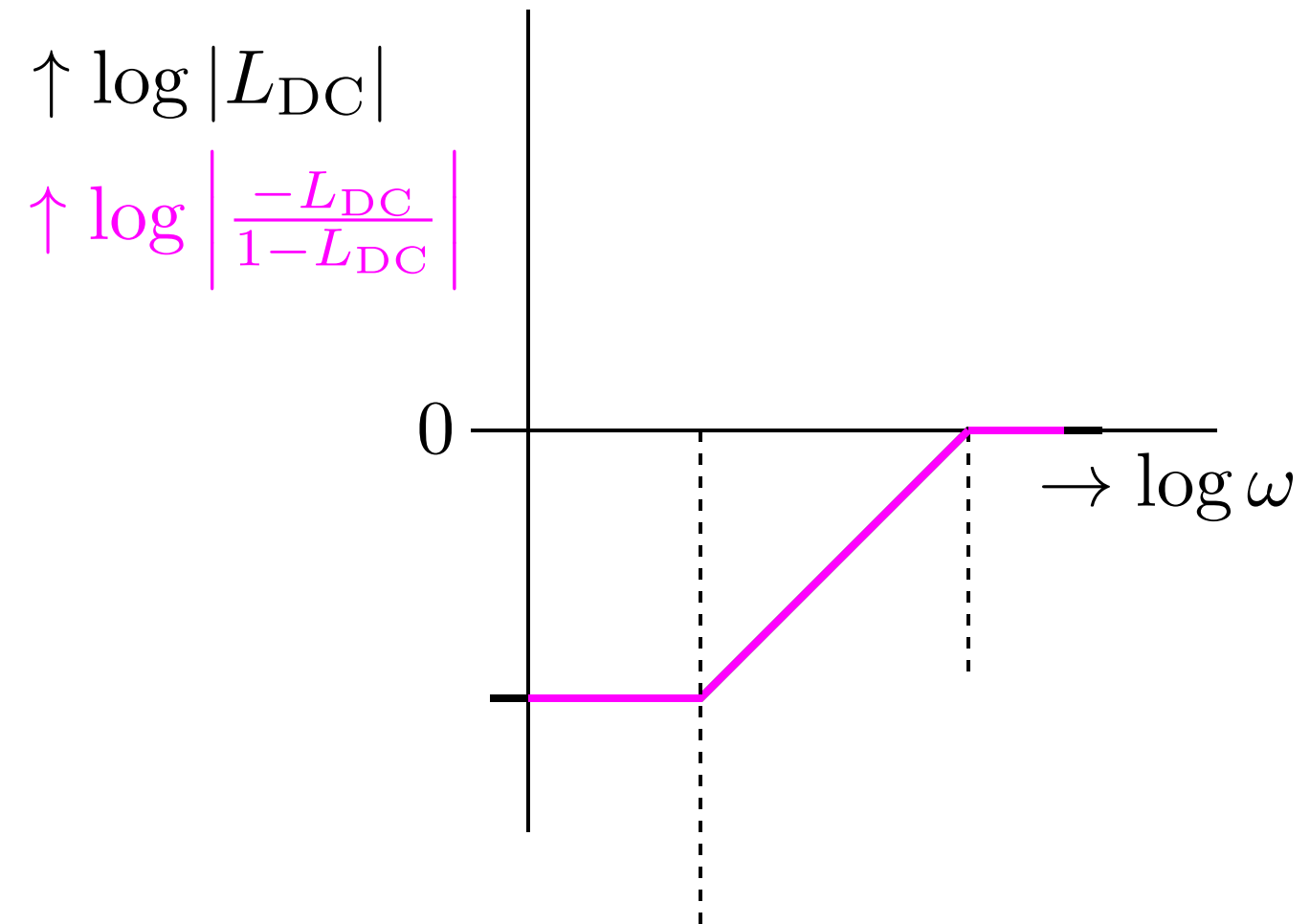
Note: pole only drops on the zero if DC loop gain is infinite!



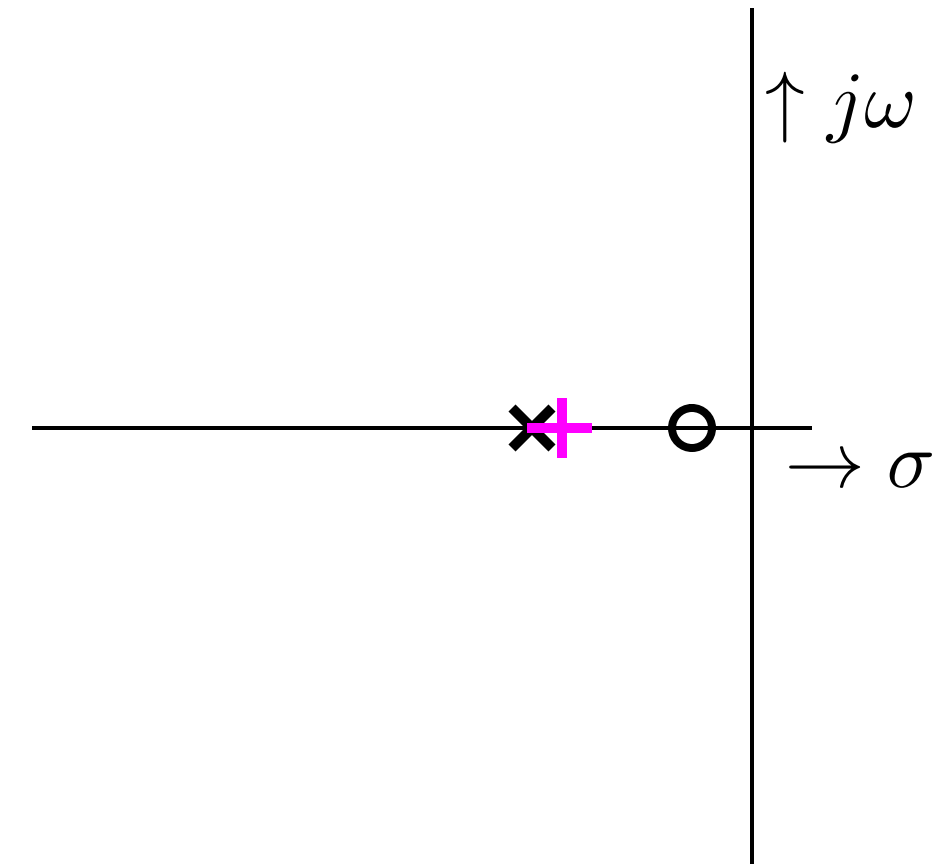
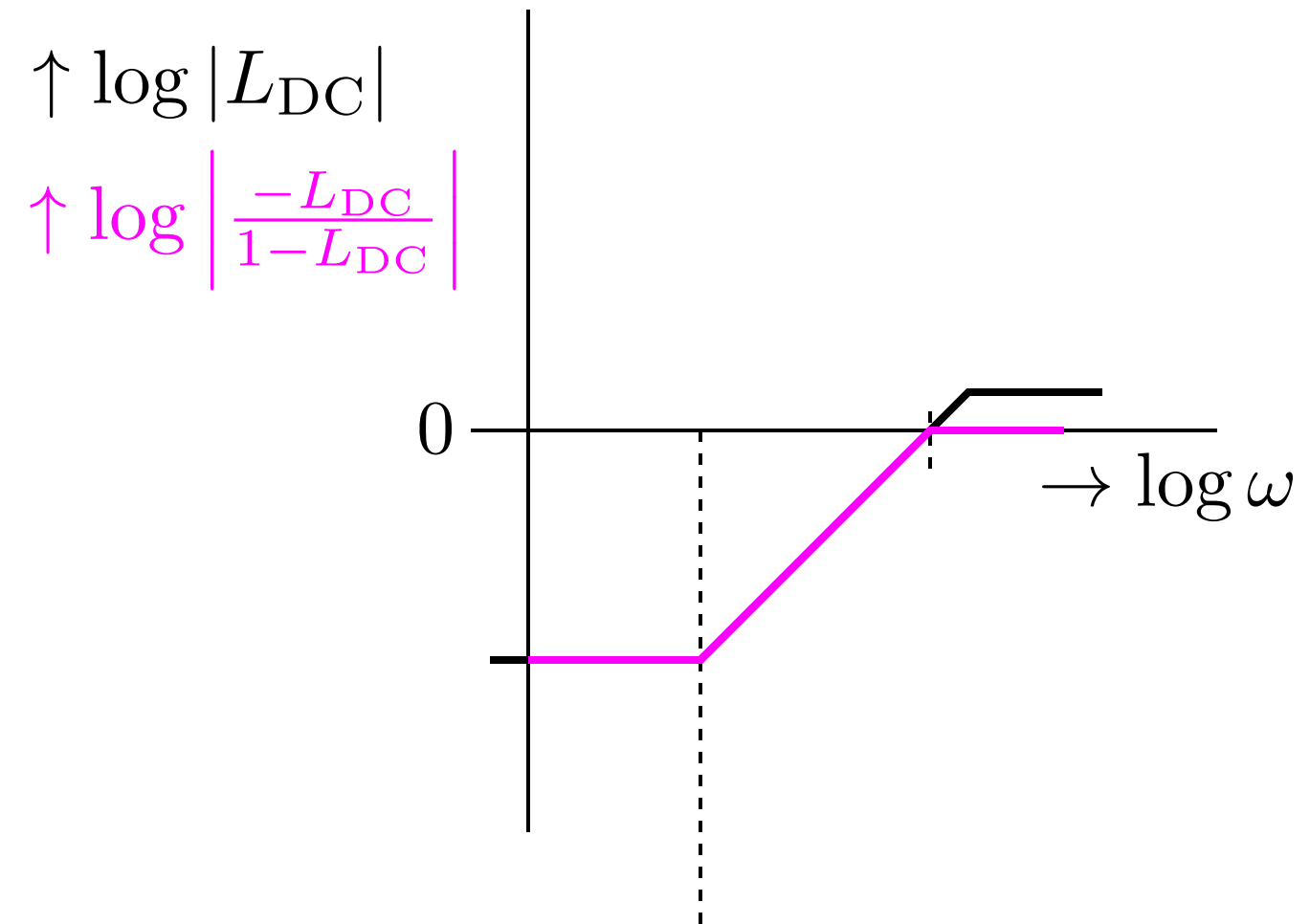
# Root locus first order with zero right



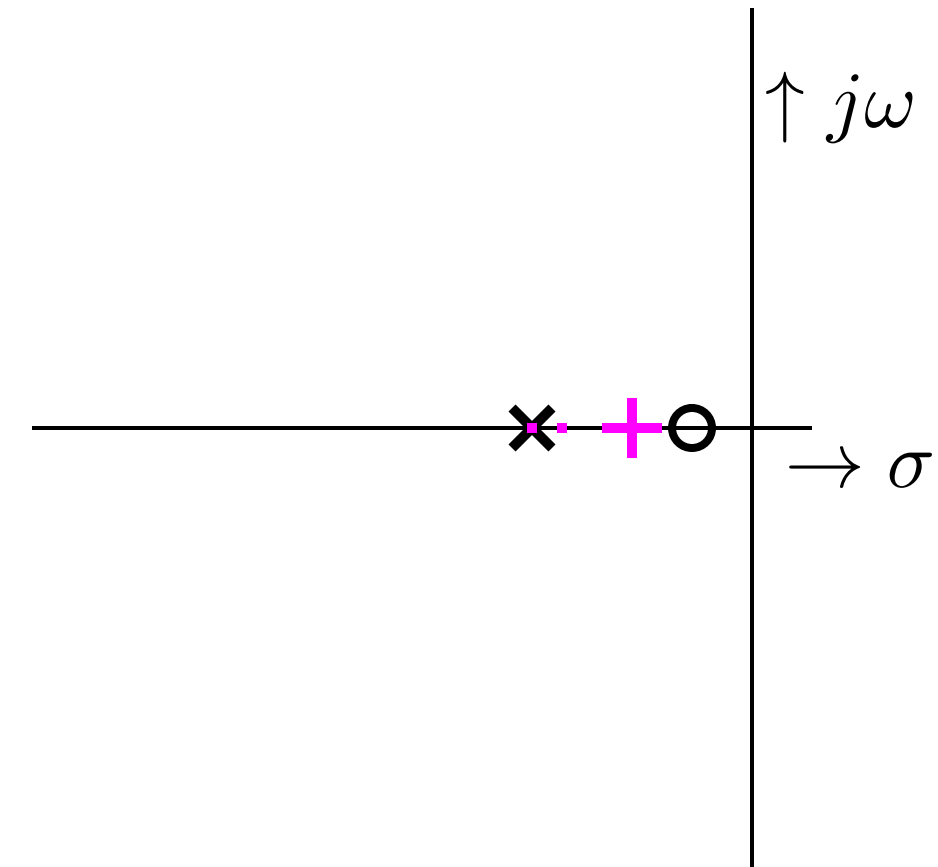
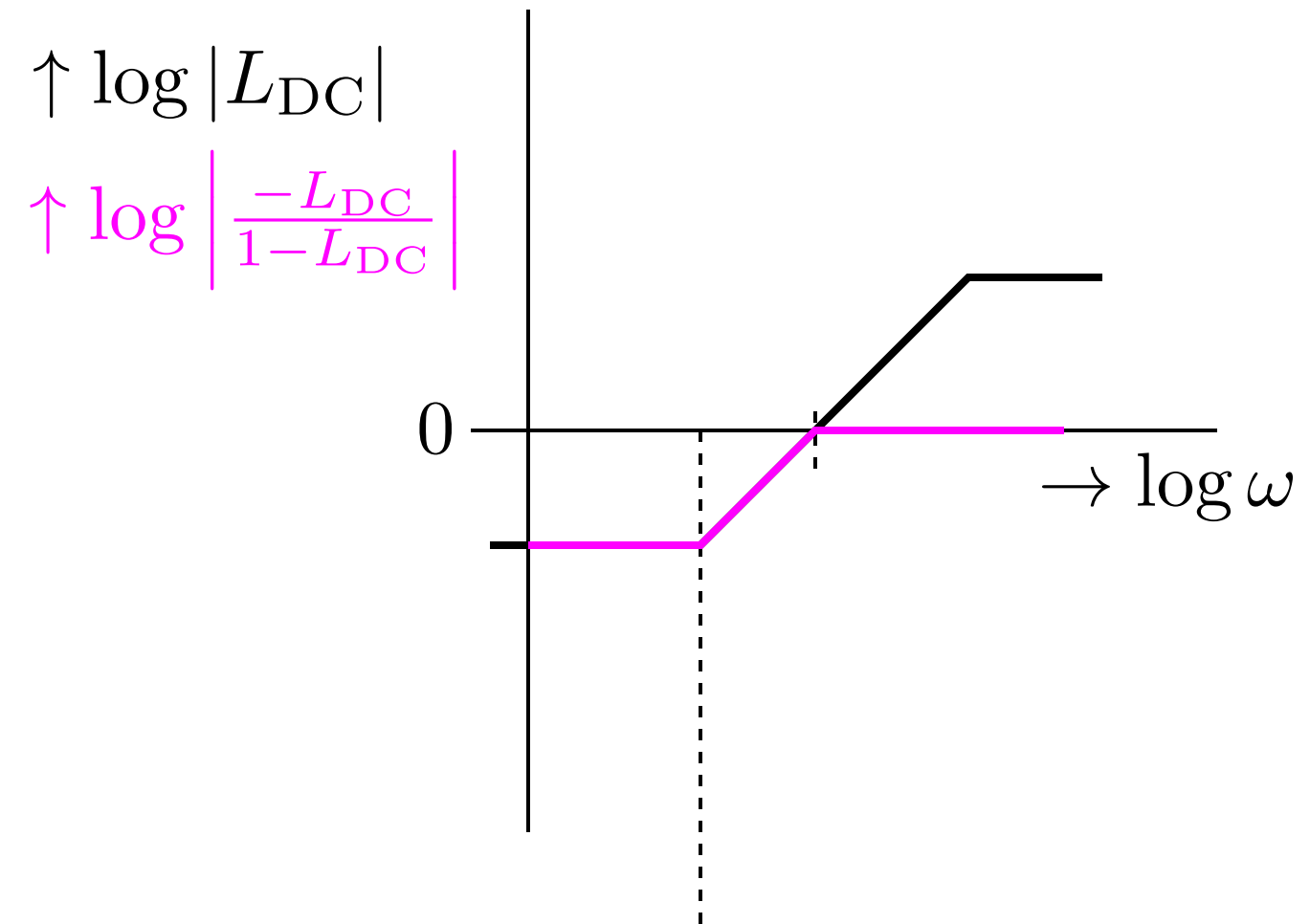
# Root locus first order with zero right



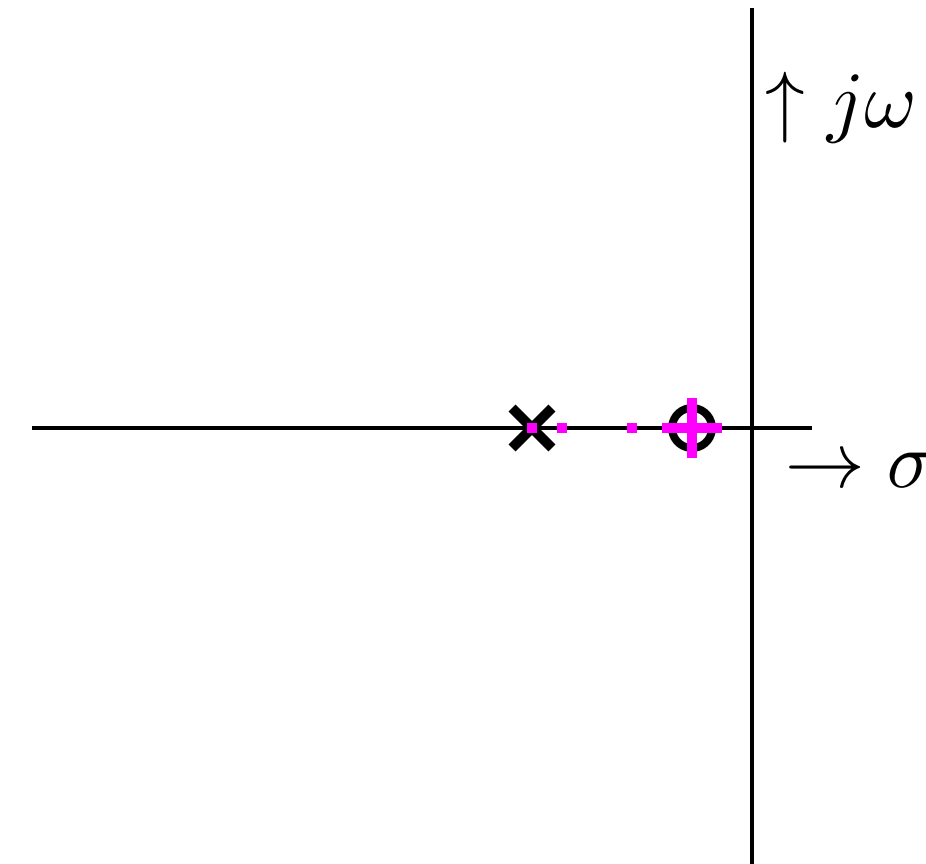
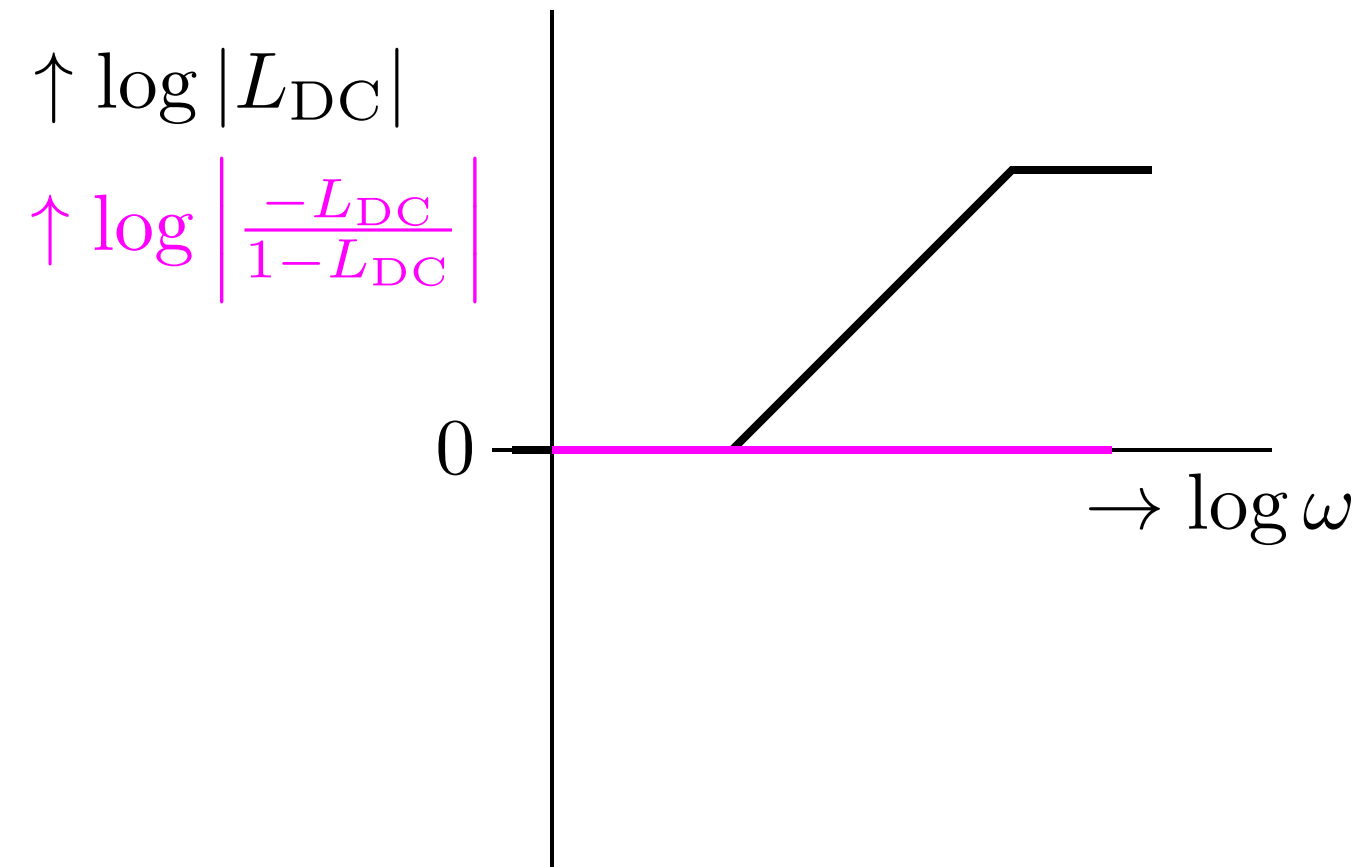
# Root locus first order with zero right



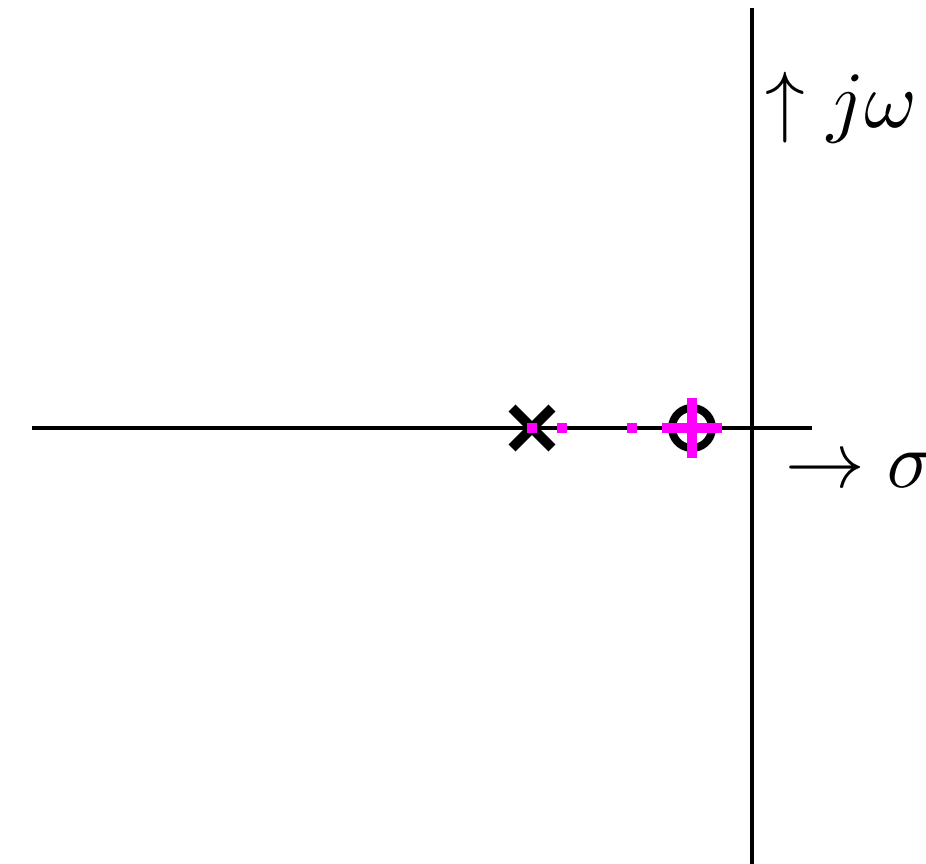
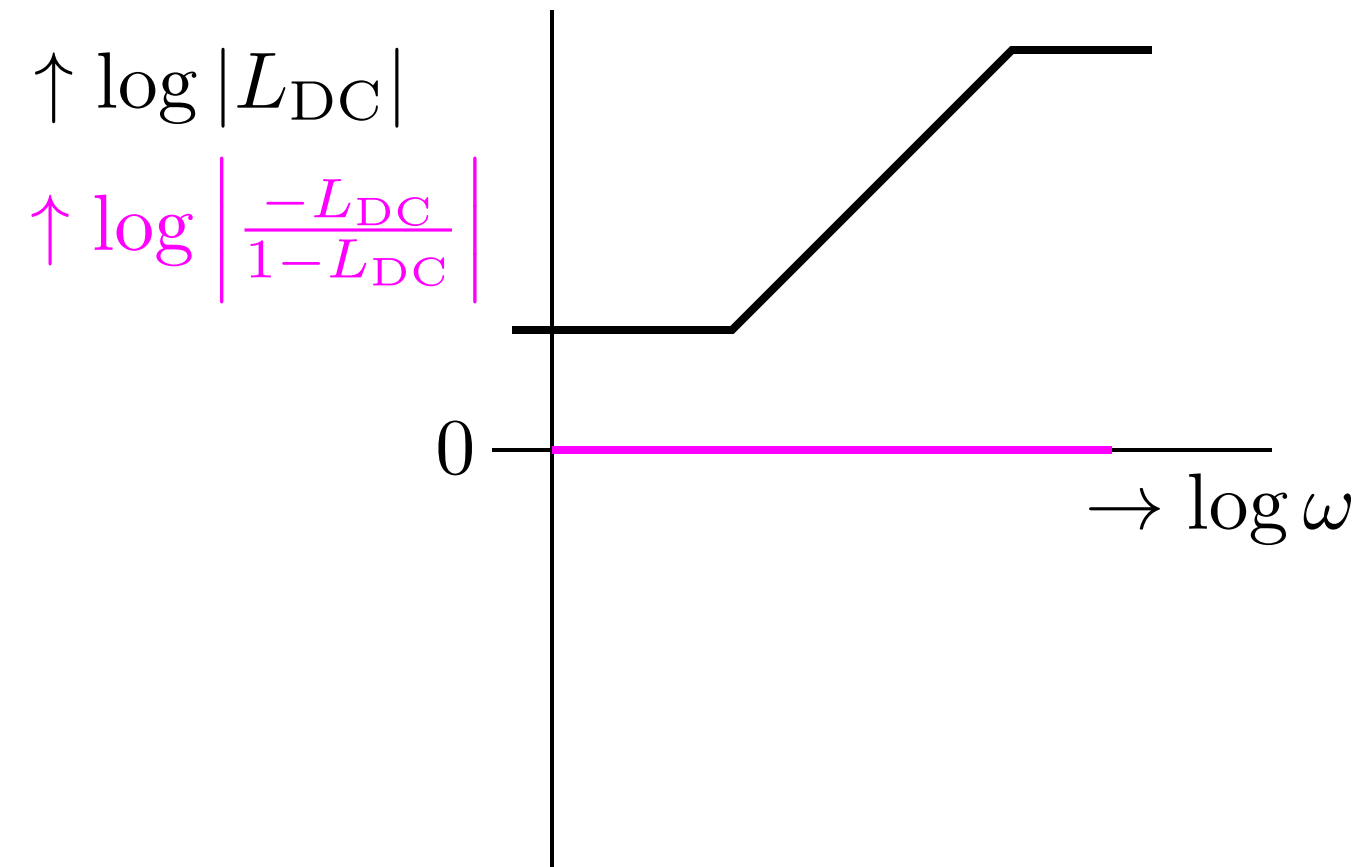
# Root locus first order with zero right



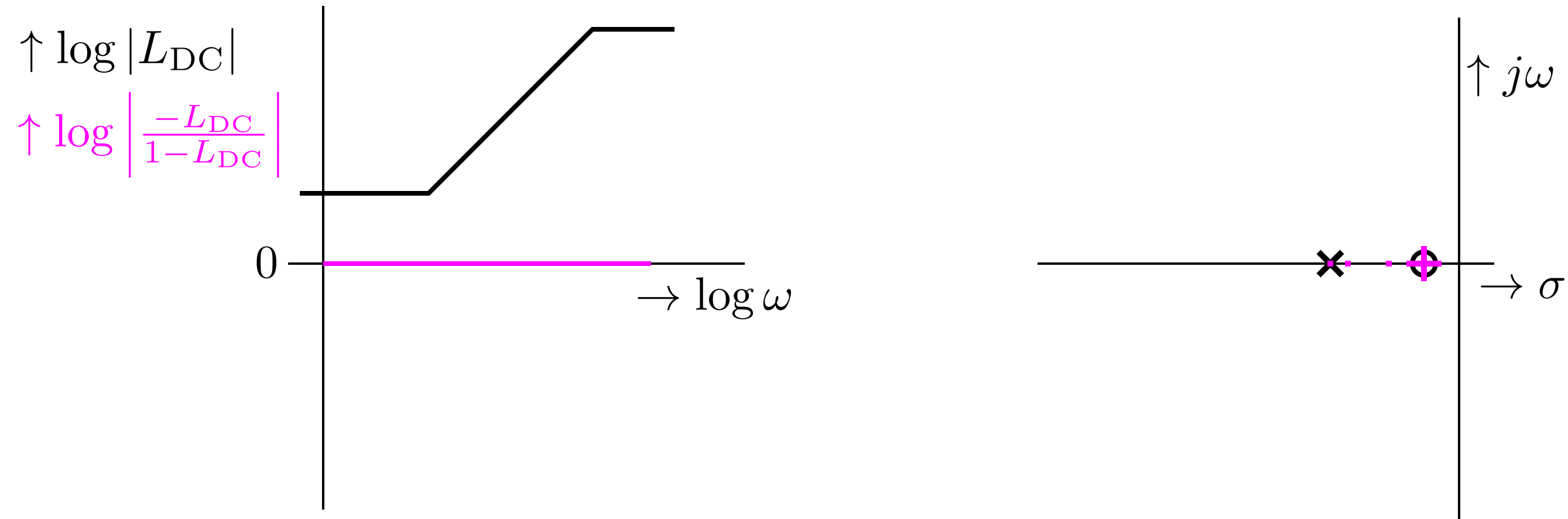
# Root locus first order with zero right



# Root locus first order with zero right

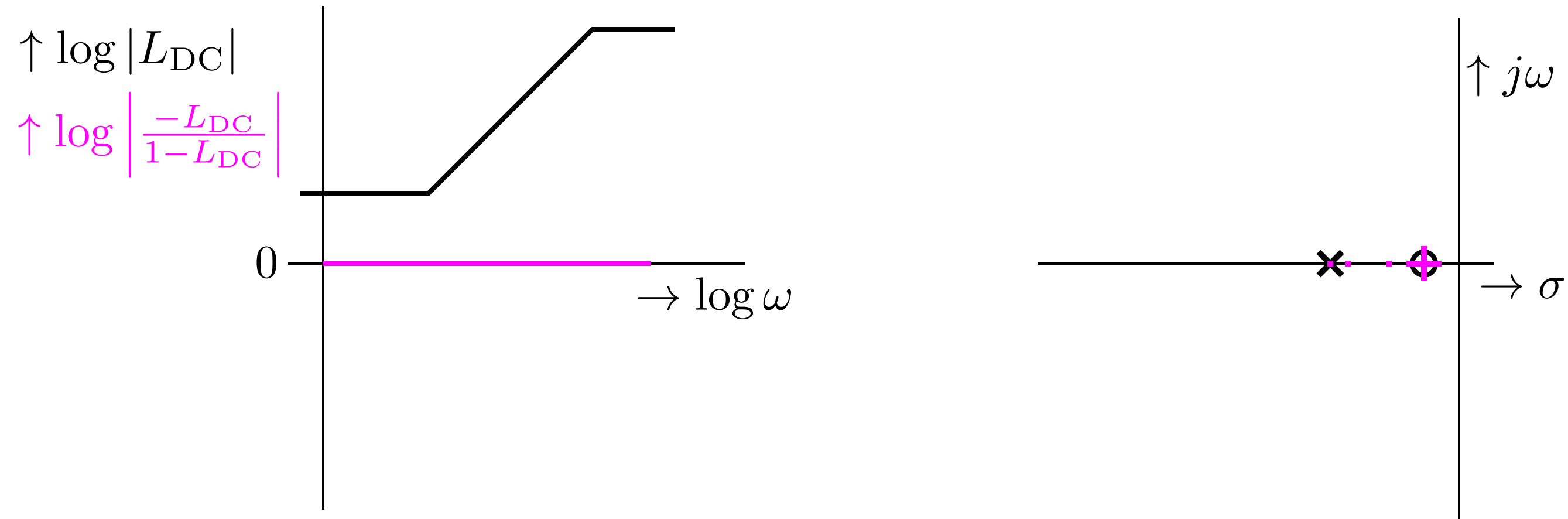


# Root locus first order with zero right



Note: pole only drops on the zero if DC loop gain is infinite!

# Root locus first order with zero right



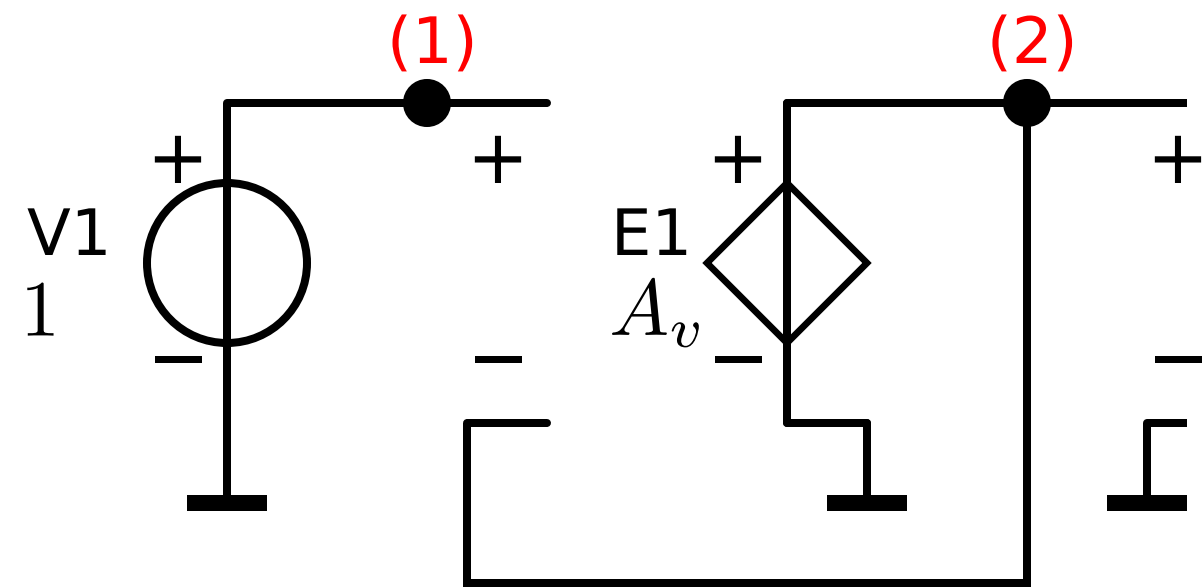
Note: pole only drops on the zero if DC loop gain is infinite!



# Root locus SLiCAP

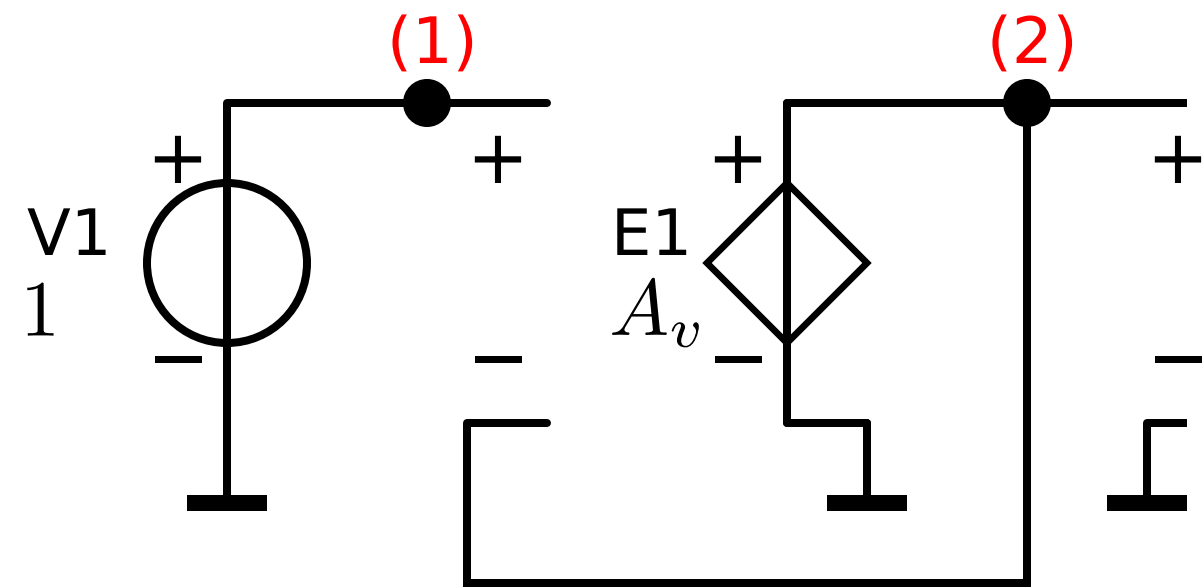
# Root locus SLiCAP

Circuit for plotting root locus



# Root locus SLiCAP

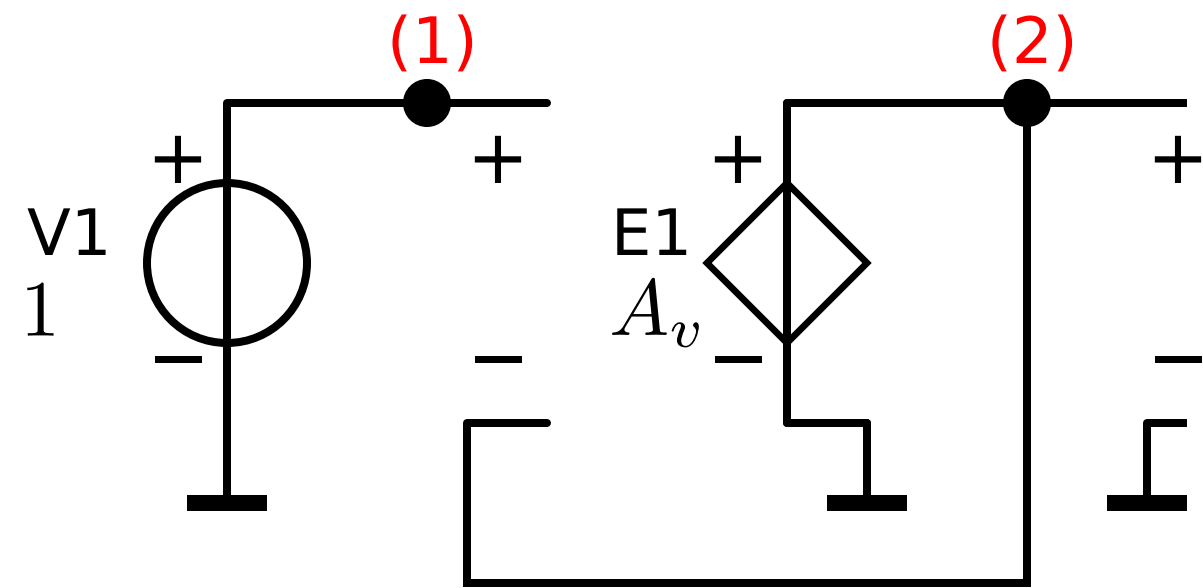
## Circuit for plotting root locus



E1 = loop gain reference

# Root locus SLiCAP

Circuit for plotting root locus

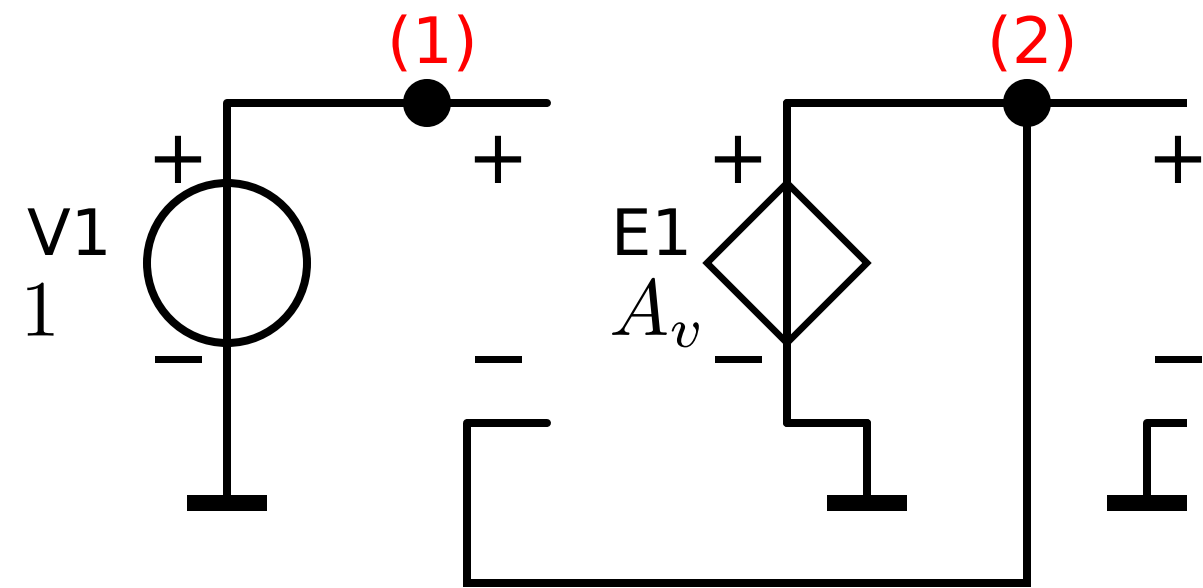


$E1$  = loop gain reference

Loop gain equals voltage gain of  $E1$

# Root locus SLiCAP

Circuit for plotting root locus



$E1$  = loop gain reference

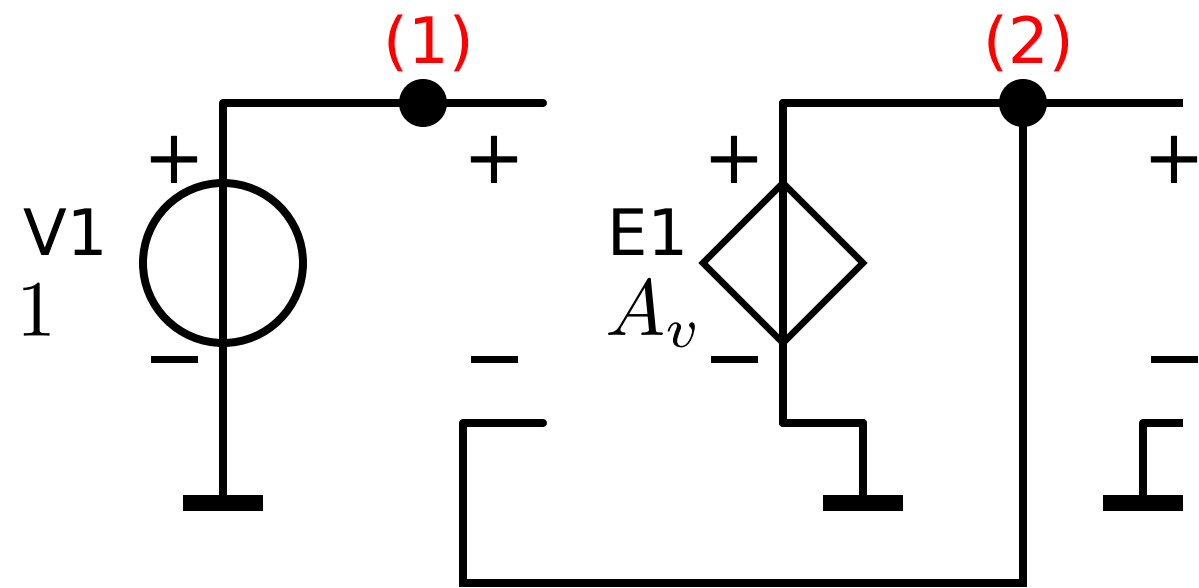
Loop gain equals voltage gain of  $E1$

Transfer of  $E1$  has DC gain, poles and zeros

# Root locus SLiCAP

Circuit for plotting root locus

Root locus plot:



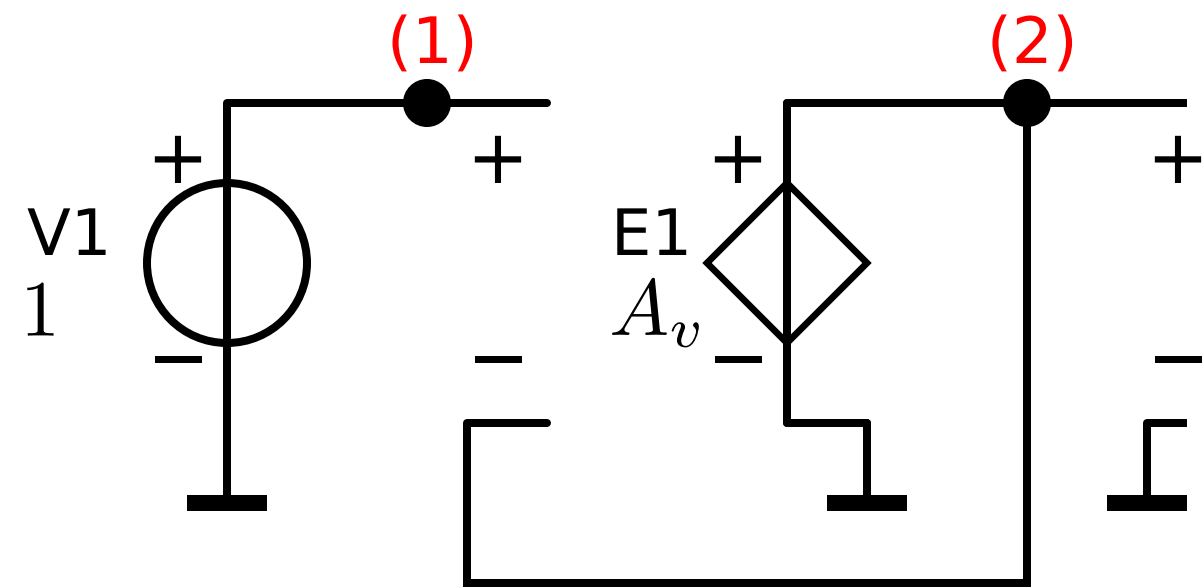
$E1$  = loop gain reference

Loop gain equals voltage gain of  $E1$

Transfer of  $E1$  has DC gain, poles and zeros

# Root locus SLiCAP

Circuit for plotting root locus



Root locus plot:

1. Poles of the loop gain

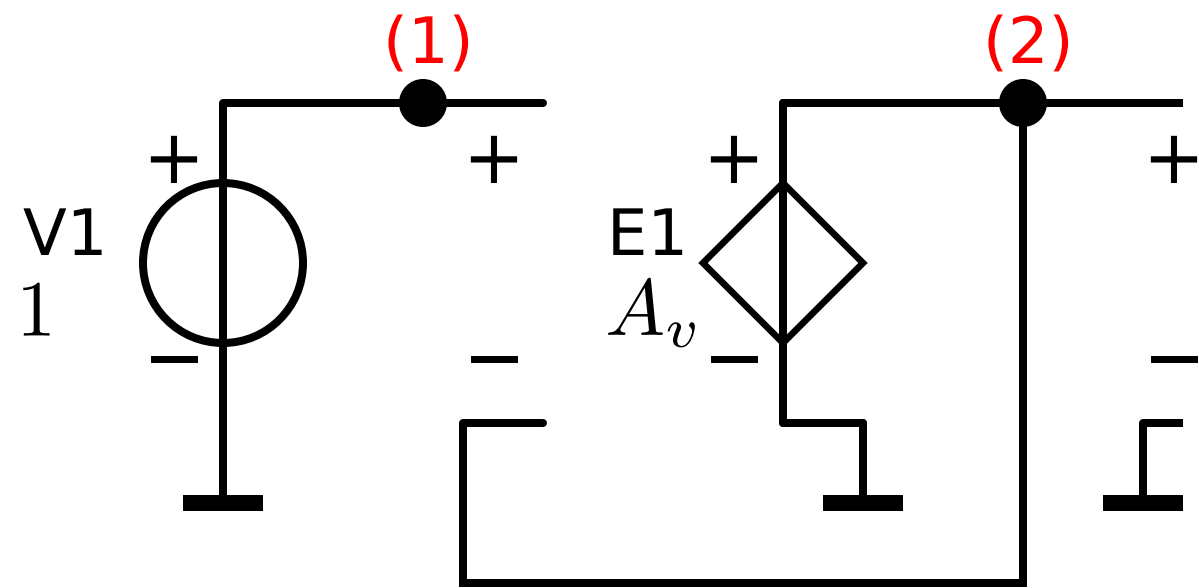
$E1$  = loop gain reference

Loop gain equals voltage gain of  $E1$

Transfer of  $E1$  has DC gain, poles and zeros

# Root locus SLiCAP

Circuit for plotting root locus



Root locus plot:

1. Poles of the loop gain
2. Zeros of the loop gain

$E1$  = loop gain reference

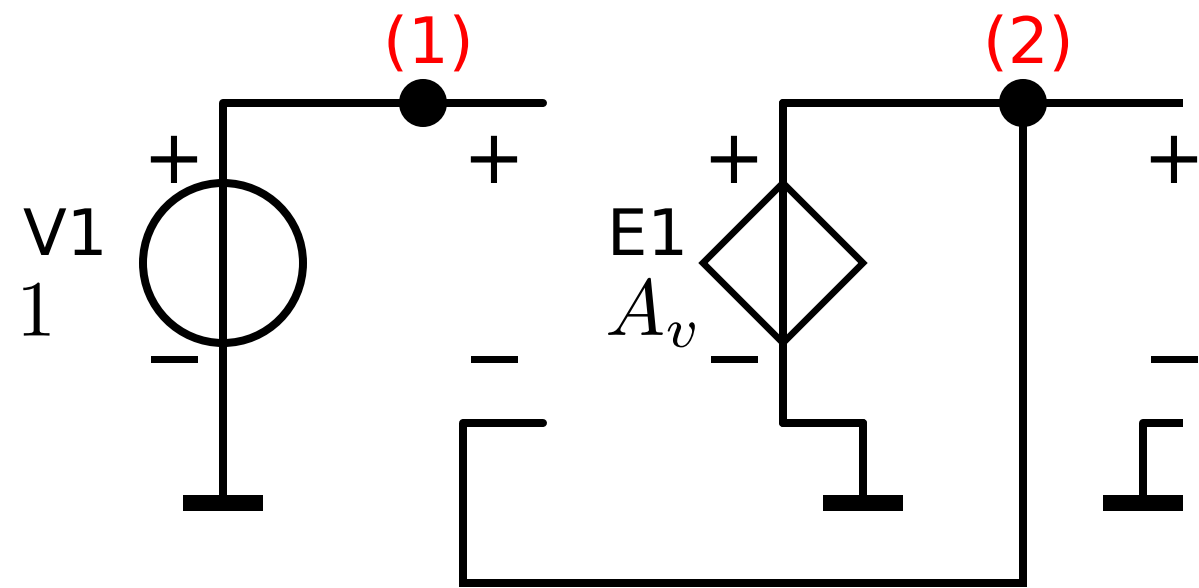
Loop gain equals voltage gain of  $E1$

Transfer of  $E1$  has DC gain, poles and zeros



# Root locus SLiCAP

Circuit for plotting root locus



E1 = loop gain reference

Loop gain equals voltage gain of E1

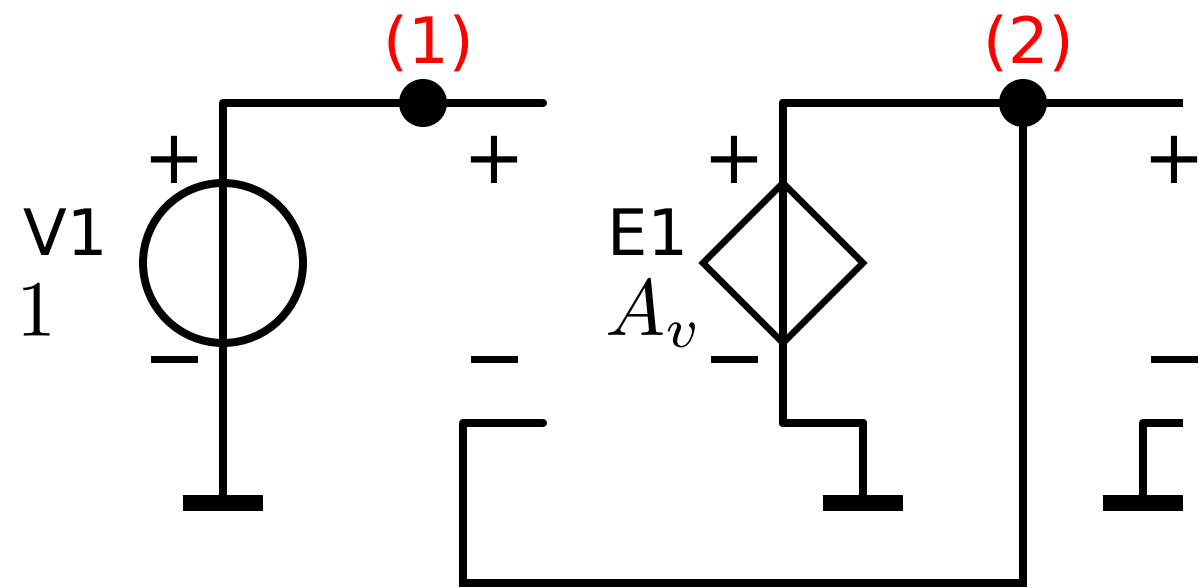
Transfer of E1 has DC gain, poles and zeros

Root locus plot:

1. Poles of the loop gain
2. Zeros of the loop gain
3. Poles of the servo function while stepping the DC gain of E1

# Root locus SLiCAP

Circuit for plotting root locus



$E1$  = loop gain reference

Loop gain equals voltage gain of  $E1$

Transfer of  $E1$  has DC gain, poles and zeros

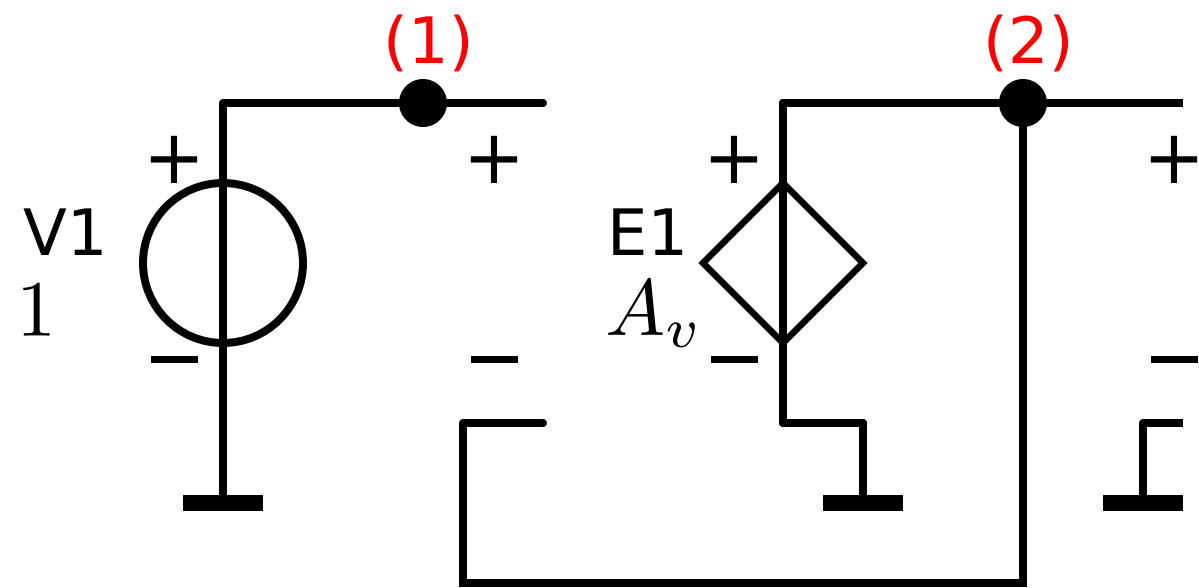
Root locus plot:

1. Poles of the loop gain
2. Zeros of the loop gain
3. Poles of the servo function while stepping the DC gain of  $E1$

[See section 11.5.3](#)

# Root locus SLiCAP

Circuit for plotting root locus



$E1$  = loop gain reference

Loop gain equals voltage gain of  $E1$

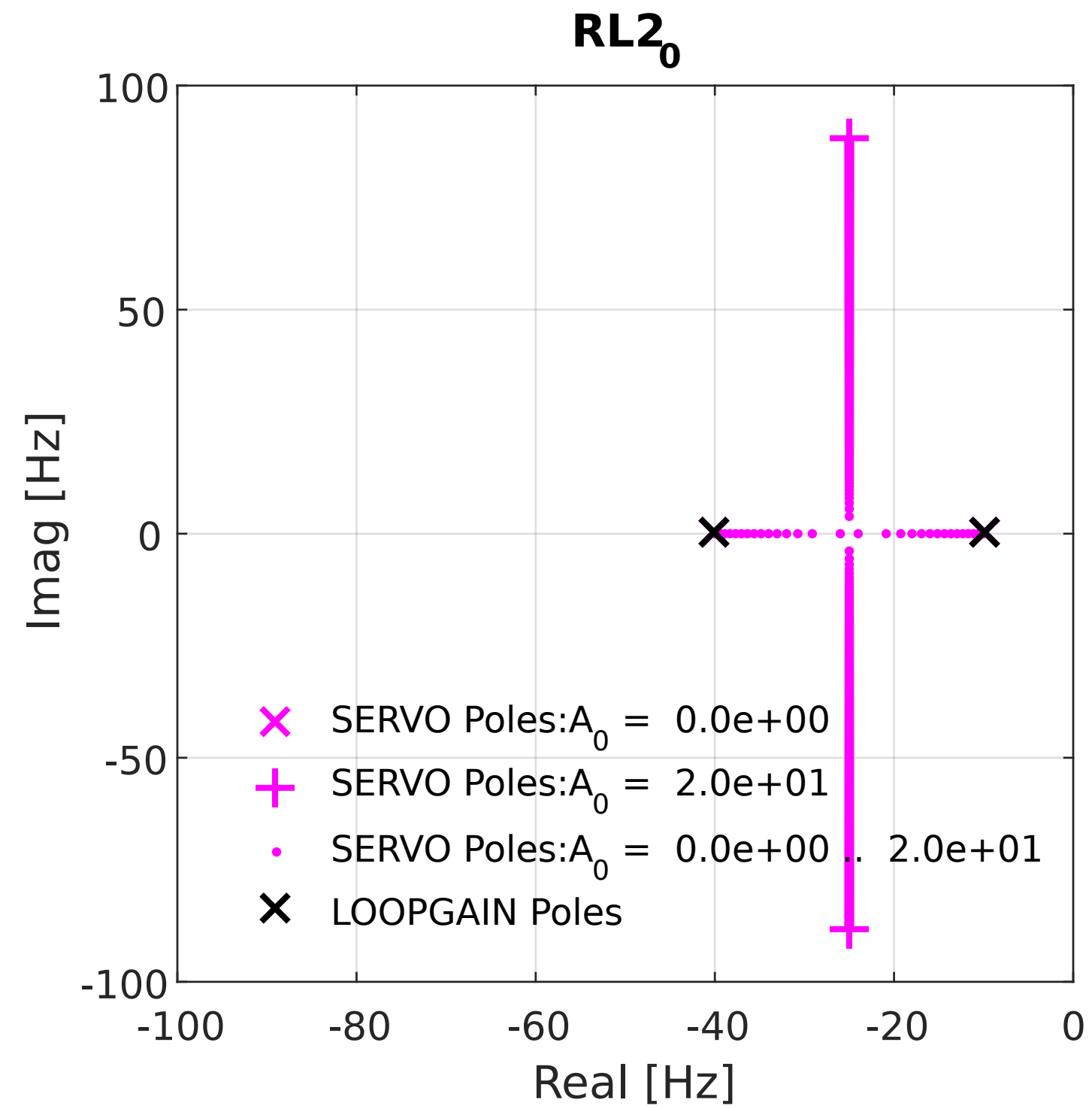
Transfer of  $E1$  has DC gain, poles and zeros

Root locus plot:

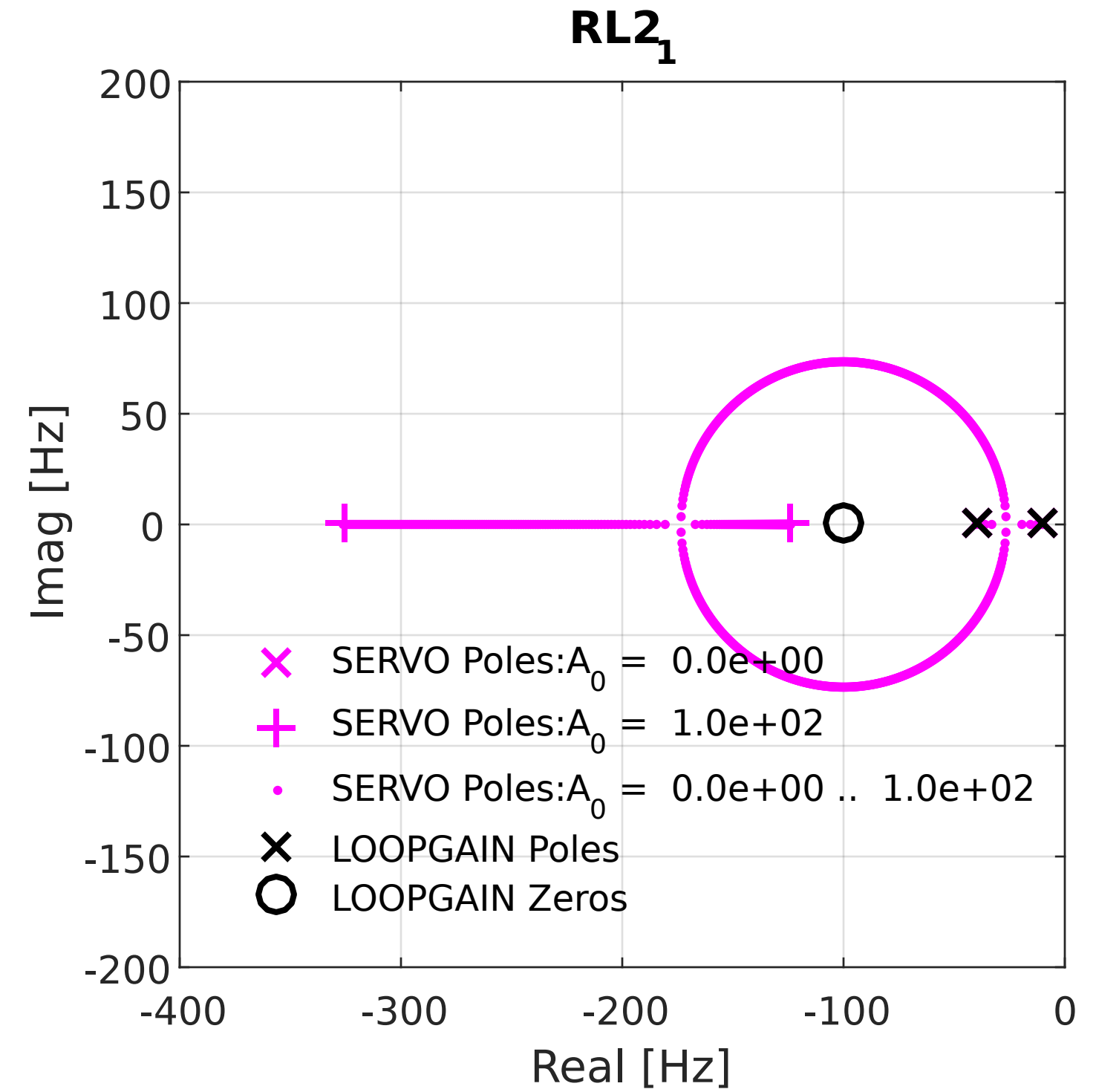
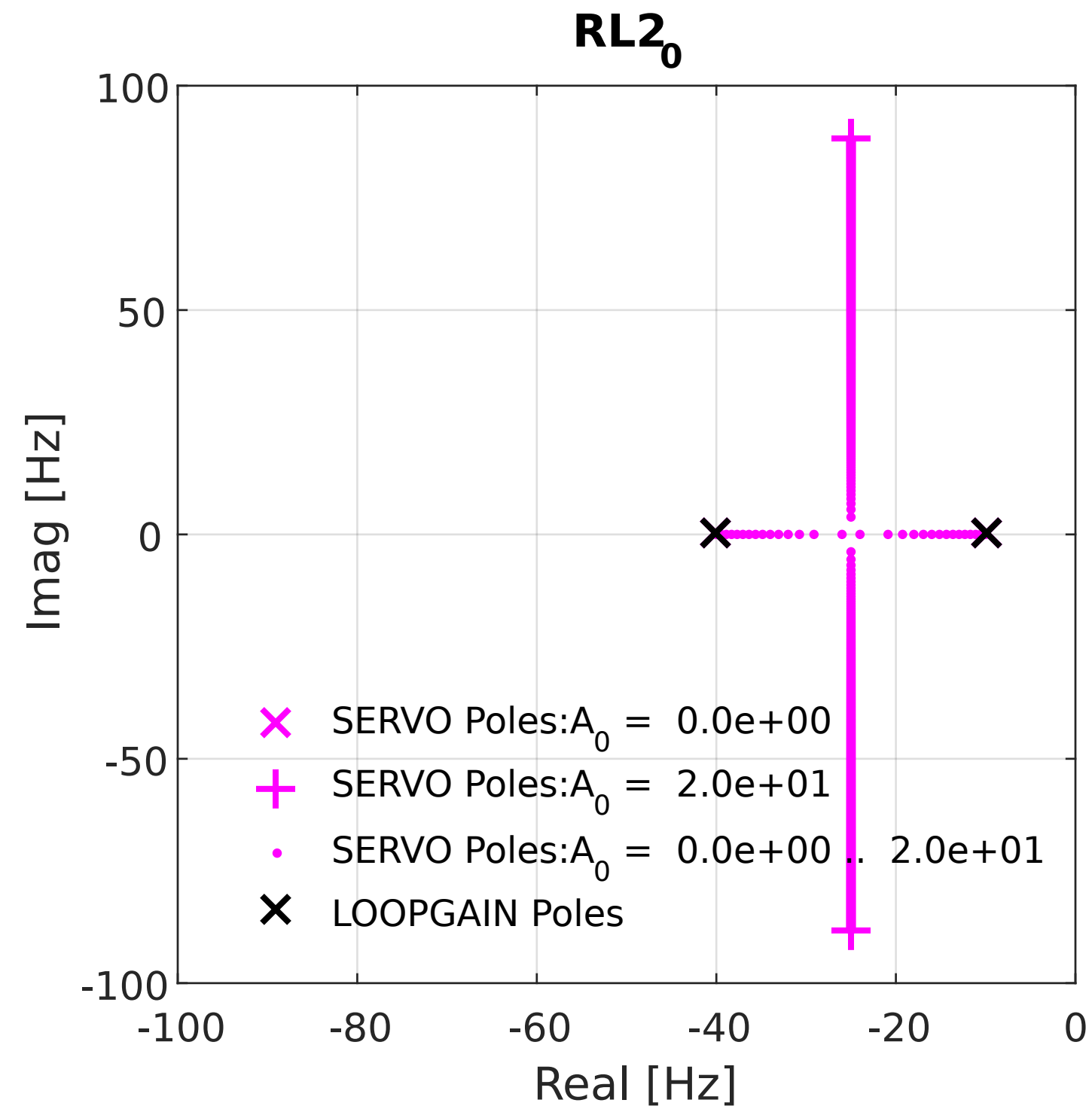
1. Poles of the loop gain
2. Zeros of the loop gain
3. Poles of the servo function while stepping the DC gain of  $E1$

See section 11.5.3

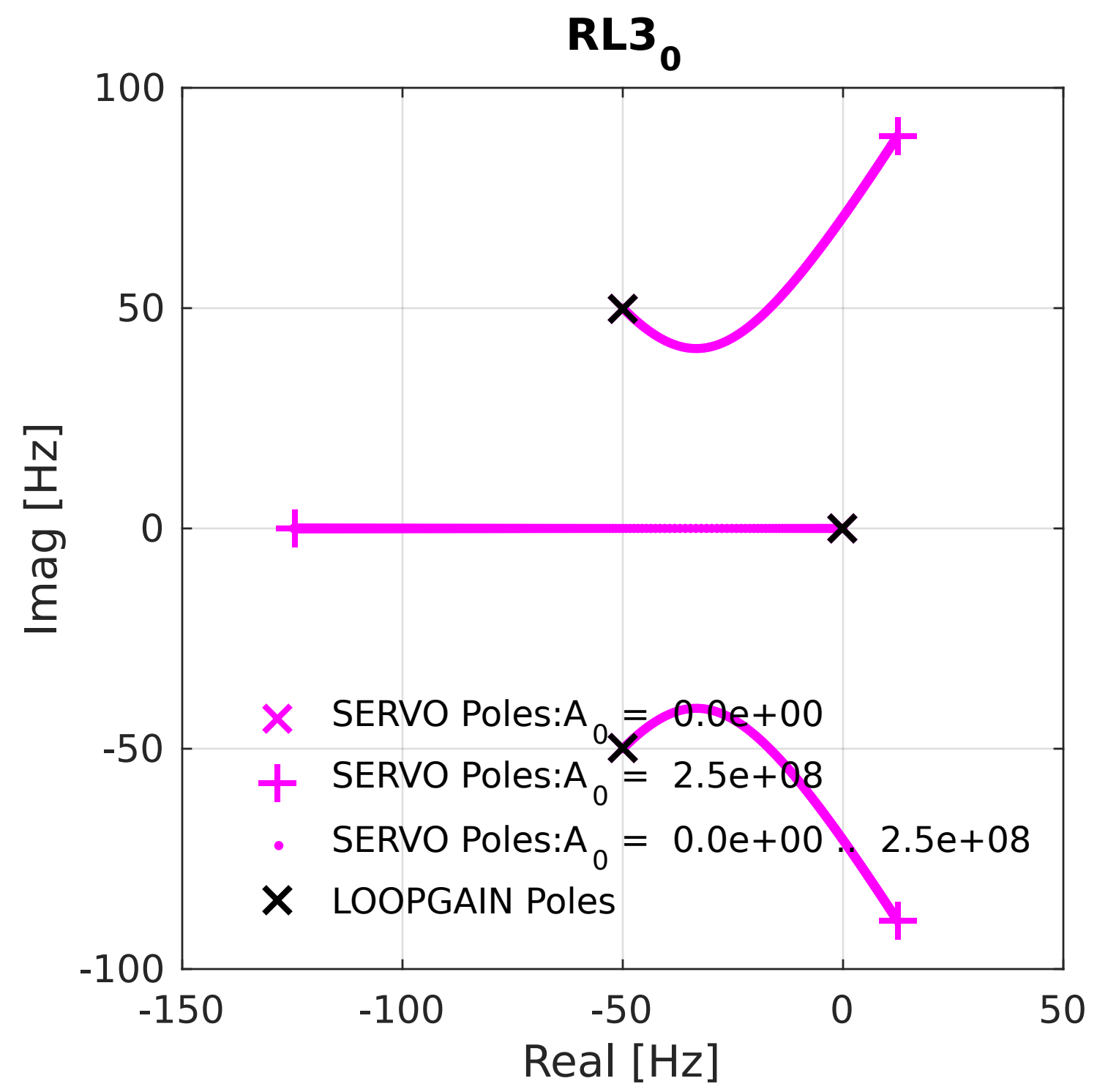
# Root locus SLiCAP second order



# Root locus SLiCAP second order



# Root locus SLiCAP third order



# Root locus SLiCAP third order

